

Dual wideband high output current operational amplifier

Features

- Wideband of 200 MHz/gain = 4
- Very high slew rate
- High output current
- Specified on 25 and 100- Ω loads
- Thermal shut-down set at $T_j = 150^\circ C$
- Large protection on outputs
- Power-down mode with high Z-out
- Very low noise
- Tested at 5 and 12-V supplies
- Single or dual supply operation
- Minimum and maximum limits are tested in full production

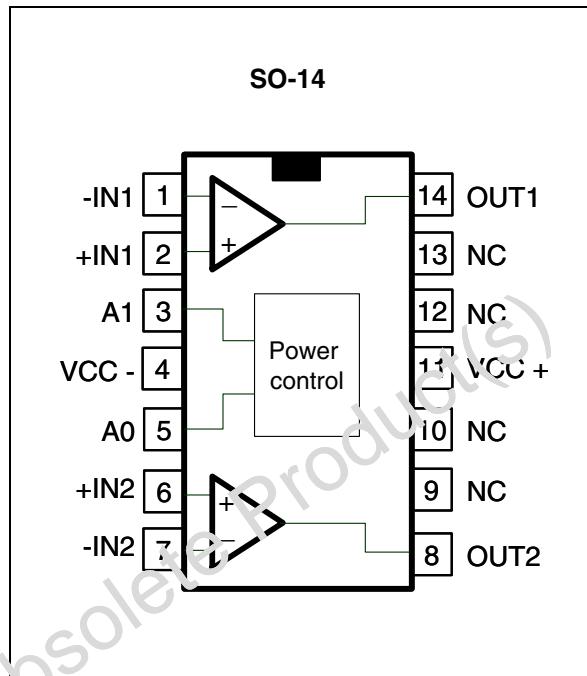
Applications

- Power line equipment
- Drivers for xDSL modems
- Video drivers
- Video distribution

Description

The TS617 is a wideband high output current dual operational amplifier, optimized for driving high data rates on power lines and twisted pair telephone lines.

The TS617 features a linearity on low resistive loads, large output swing and high output current, very low noise, low consumption and I-supply control and power-down mode with high Z-out and short settling times.



1 Absolute maximum ratings and operating conditions

Table 1. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{CC}	Supply voltage ⁽¹⁾	14	V_{DC}
V_{id}	Differential input voltage ⁽²⁾	± 2	V
T_{oper}	Operating free air temperature range	-40 to +85	°C
T_{stg}	Storage temperature	-65 to +150	°C
T_j	Maximum junction temperature	150	°C
Θ_{JA}	SO14 thermal resistance junction to ambient	103	°C/W
Θ_{JC}	SO14 thermal resistance junction to case	31	°C/W
P_{max}	SO14 maximum power dissipation ⁽³⁾ (at $T_a = 25^\circ C$) for $T_j = 150^\circ C$	1.2	W
ESD	HBM: human body model ⁽⁴⁾	2	kV
	MM: machine model ⁽⁵⁾	200	V
	CDM: charged device model	1	kV

1. All voltage values are measured with respect to the ground pin.
2. Differential voltage is between the non-inverting input terminal and the inverting input terminal.
3. Short-circuits can cause excessive heating. Destructive dissipation can result from short circuits on amplifiers. An internal thermal shut-down set the circuit in power down mode automatically when the silicon temperature reaches 150°C.
4. Human body model: 100 pF discharged through a 1.5 kΩ resistor into P_{min} of device.
5. This is a minimum value. Machine model ESD: a 200 pF capacitor is charged to the specified voltage, then discharged directly into the IC with no external series resistor.

Table 2. Operating conditions

Symbol	Parameter	Value	Unit
V_{CC}	Power supply voltage ⁽¹⁾	4.5 to 13	V
V_{icm}	Common mode input voltage	± 6	V

1. Tested in full production at $\pm 2.5 V$ and $\pm 6 V$ supply voltages.

2 Electrical characteristics

Table 3. $V_{CC} = \pm 2.5$ V, $T_{amb} = 25^\circ$ C, full power mode (unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
DC performance						
V_{io}	Input offset voltage	T_{amb}	-10	2.5	10	mV
		$T_{min.} < T_{amb} < T_{max.}$		2.7		
ΔV_{io}	V_{io} drift vs. temperature	$T_{min.} < T_{amb} < T_{max.}$		5		$\mu\text{V}/^\circ\text{C}$
		T_{amb}	-50	-15	+50	
I_{ib+}	Non-inverting input bias current	$T_{min.} < T_{amb} < T_{max.}$		-20		μA
		T_{amb}	-30	-11	-30	
I_{ib-}	Inverting input bias current	$T_{min.} < T_{amb} < T_{max.}$		-12		μA
		T_{amb}	-30	-11	-30	
CMR	Common mode rejection ratio 20 log ($\Delta V_{ic}/\Delta V_{io}$)	$\Delta V_{ic} = \pm 1$ V	45	51		dB
		$T_{min.} < T_{amb} < T_{max.}$		50		
SVR	Supply voltage rejection ratio 20 log ($\Delta V_{CC}/\Delta V_{io}$)	$\Delta V_{CC} = \pm 2$ V to ± 2.5 V	50	76		dB
		$T_{min.} < T_{amb} < T_{max.}$		55		
$PSRR$	Power supply rejection ratio 20 log ($\Delta V_{CC}/\Delta V_{out}$)	Gain = +4, Rfb = 390 Ω $\Delta V_{CC} = \pm 100$ mV at 100 kHz		50		dB
I_{CC}	Supply current per operator	Full power, no load		9	11	mA
		$T_{min.} < T_{amb} < T_{max.}$		10		
		Medium power, no load		7	8	mA
		$T_{min.} < T_{amb} < T_{max.}$		7.5		
		Low power, no load		2	3	mA
Dynamic performance and output characteristics						
Bw	-3 dB bandwidth in small signal	Small signal $V_{out} = 20$ mVp-p $R_L = 100 \Omega$ Gain = +2, Rfb = 620 Ω Gain = +4, Rfb = 390 Ω Gain = +8, Rfb = 390 Ω	70	125 105 55		MHz
		Gain = +4, medium power		90		
		Gain = +4, low power		33		
R_{OL}	Open-loop transimpedance	$R_L = 100 \Omega$ $V_{out} = 2$ Vp-p	50	110		$k\Omega$
		$T_{min.} < T_{amb} < T_{max.}$		90		
SR	Slew rate	$V_{out} = 3$ Vp-p, gain = +4, $R_L = 100 \Omega$	300	450		$V/\mu\text{s}$
		Medium power		400		
		Low power		100		

Table 3. $V_{CC} = \pm 2.5$ V, $T_{amb} = 25^\circ$ C, full power mode (unless otherwise specified) (continued)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{OH}	High level output voltage	$R_L = 100 \Omega$ $R_L = 25 \Omega$	1.6 1.5	1.68 1.58		V
		$T_{min.} < T_{amb} < T_{max.}$ $R_L = 100 \Omega$ $R_L = 25 \Omega$		1.6 1.5		
V_{OL}	Low level output voltage	$R_L = 100 \Omega$ $R_L = 25 \Omega$		-1.68 -1.57	-1.60 -1.50	V
		$T_{min.} < T_{amb} < T_{max.}$ $R_L = 100 \Omega$ $R_L = 25 \Omega$			-1.6 -1.5	
I_{out}	I_{sink}	$V_{out} = -1$ V	300	350		mA
		$T_{min.} < T_{amb} < T_{max.}$		330		
	I_{source}	$V_{out} = +1$ V		-370	-300	
		$T_{min.} < T_{amb} < T_{max.}$		-350		
	$I_{short\ circuit}$ (limited by thermal shut down)	Output to GND		600		
Noise and distortion						
eN	Input noise voltage	$F = 100$ kHz		3.5		nV/ $\sqrt{\text{Hz}}$
iNp	Positive input noise current	$F = 100$ kHz		39		pA/ $\sqrt{\text{Hz}}$
iNn	Positive input noise current	$F = 100$ kHz		20		pA/ $\sqrt{\text{Hz}}$
H2/H3	2nd and 3rd harmonics	$V_{out} = 2$ Vp-p, gain = +4, $R_{fb}=390 \Omega$, $F = 6$ MHz $R_L = 25 \Omega$ H2 H3			-58 -68	dBc
Power-down function						
Ipdw	Consumption in power-down mode	A1 = 0, A0 = 0		65	200	μA
		$T_{min.} < T_{amb} < T_{max.}$		80		
ton	Time for power-down mode to operating mode	A1 = A0 = 0 to A1 = A0 = 1 $V_{out} = 1$ Vdc, gain = +4		80	200	ns
toff	Time for operating mode to power-down mode	A1 = A0 = 1 to A1 = A0 = 0 $V_{out} = 1$ Vdc, gain = +4		450	1000	ns

Table 4. $V_{CC} = \pm 6$ V, $T_{amb} = 25^\circ$ C, full power mode (unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
DC performance						
V_{io}	Input offset voltage	T_{amb}	-12	4.5	12	mV
		$T_{min.} < T_{amb} < T_{max.}$		5		
ΔV_{io}	V_{io} drift vs. temperature	$T_{min.} < T_{amb} < T_{max.}$		5		$\mu\text{V}/^\circ\text{C}$
I_{ib+}	Non-inverting input bias current	T_{amb}	-80	-25	80	μA
		$T_{min.} < T_{amb} < T_{max.}$		-30		
I_{ib-}	Inverting input bias current	T_{amb}	-35	-7	35	μA
		$T_{min.} < T_{amb} < T_{max.}$		-9		
CMR	Common mode rejection ratio 20 log ($\Delta V_{ic}/\Delta V_{io}$)	$\Delta V_{ic} = \pm 3$ V	45	51		dB
		$T_{min.} < T_{amb} < T_{max.}$		50		
SVR	Supply voltage rejection ratio 20 log ($\Delta V_{CC}/\Delta V_{io}$)	$\Delta V_{CC} = \pm 2.5$ V to ± 6 V	50	72		dB
		$T_{min.} < T_{amb} < T_{max.}$		65		
PSRR	Power supply rejection ratio 20 log ($\Delta V_{CC}/\Delta V_{out}$)	Gain = +4, $R_{fb} = 390$ Ω $\Delta V_{CC} = \pm 100$ mV at 100 kHz		50		dB
I_{CC}	Supply current per operator	Full power, no load		13	16	mA
		$T_{min.} < T_{amb} < T_{max.}$		14		
		Medium power, no load		9.7	12	mA
		$T_{min.} < T_{amb} < T_{max.}$		10.5		
		Low power, no load		2.6	4	mA
		$T_{min.} < T_{amb} < T_{max.}$		2.9		
Dynamic performance and output characteristics						
Bw	-3 dB bandwidth in small signal	Small signal $V_{out} = 20$ mVp-p $R_L = 100$ Ω Gain = +2, $R_{fb} = 620$ Ω Gain = +4, $R_{fb} = 390$ Ω Gain = +8, $R_{fb} = 390$ Ω	140	300 210 110		MHz
		Gain = +4, medium power		160		
		Gain = +4, low power		60		
ROL	Open-loop transimpedance	$R_L = 100$ Ω , $V_{out} = 7$ Vp-p	30	70		$k\Omega$
		$T_{min.} < T_{amb} < T_{max.}$		50		
SR	Slew rate	Gain = +4, $R_L = 100$ Ω $V_{out} = 5$ Vp-p $V_{out} = 10$ Vp-p	400	650 1000		$V/\mu\text{s}$
		Medium power, $V_{out} = 5$ Vp-p		500		
		Low power, $V_{out} = 5$ Vp-p		200		
V_{OH}	High level output voltage	$R_L = 100$ Ω	5	5.1		V
		$T_{min.} < T_{amb} < T_{max.}$		5		

Table 4. $V_{CC} = \pm 6 \text{ V}$, $T_{amb} = 25^\circ \text{ C}$, full power mode (unless otherwise specified) (continued)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit	
V_{OL}	Low level output voltage	$R_L = 100 \Omega$		-5.1	-5	V	
		$T_{min.} < T_{amb} < T_{max.}$		-5			
I_{out}	I_{sink}	$V_{out} = -4 \text{ V}$	450	540		mA	
		$T_{min.} < T_{amb} < T_{max.}$		510			
	I_{source}	$V_{out} = +4 \text{ V}$		-550	-450		
		$T_{min.} < T_{amb} < T_{max.}$		-490			
$I_{short \ circuit}$ (limited by thermal shut down)		Output to GND (limited by internal thermal shut-down)		800			
Noise and distortion							
eN	Equivalent input noise voltage	$F = 100 \text{ kHz}$		3.5		$\text{nV}/\sqrt{\text{Hz}}$	
iNp	Positive input noise current	$F = 100 \text{ kHz}$		39		$\text{pA}/\sqrt{\text{Hz}}$	
iNn	Positive input noise current	$F = 100 \text{ kHz}$		20		$\text{pA}/\sqrt{\text{Hz}}$	
H2/H3	2nd and 3rd harmonics	$V_{out} = 2 \text{ Vp-p}$, gain = +4, $F = 6 \text{ MHz}$ $R_L = 100 \Omega$ H2 H3		-66 -73		dBc	
Power-down function							
I_{pdw}	Consumption in power-down mode	$A1 = 0, A0 = 0$		90	300	μA	
		$T_{min.} < T_{amb} < T_{max.}$		110			
t_{on}	Time for power-down mode to operating mode	$A1 = A0 = 0$ to $A1 = A0 = 1$ $V_{out} = 4 \text{ Vdc}$, gain = +4		60	200	ns	
t_{off}	Time for operating mode to power-down mode	$A1 = A0 = 1$ to $A1 = A0 = 0$ $V_{out} = 4 \text{ Vdc}$, gain = +4		550	1000	ns	

Table 5. A1 and A0 thresholds

Symbol	Parameter	Test condition	Min.	Max.
A1	A1 and A0 threshold	High level	$-V_{CC} + 2\text{ V}$	$+V_{CC}$
A0		Low level	$-V_{CC}$	$-V_{CC} + 0.8\text{ V}$

Table 6. A1 and A0 states

A1	A0	State of the TS617
0	0	Standby
0	1	Low power
1	0	Medium power
1	1	High power

Table 7. Feedback resistor

Gain (V_2/V_1)	R_{fb} (Ω)
2	620
4	390
8	390

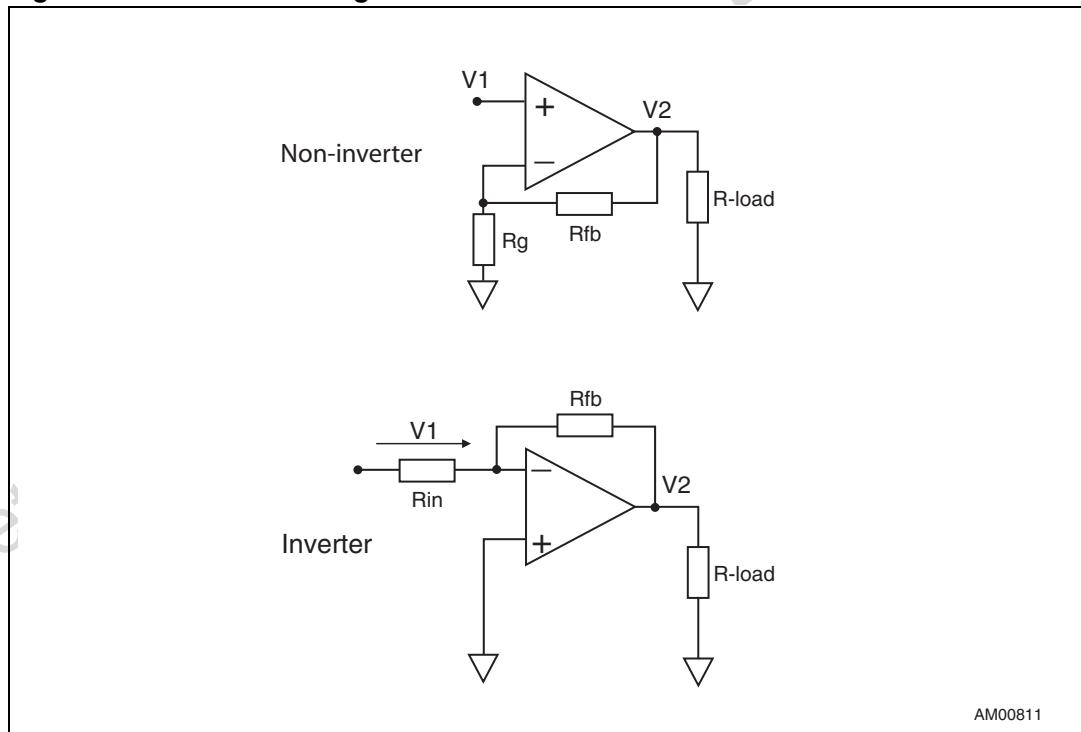
Figure 1. Feedback configuration

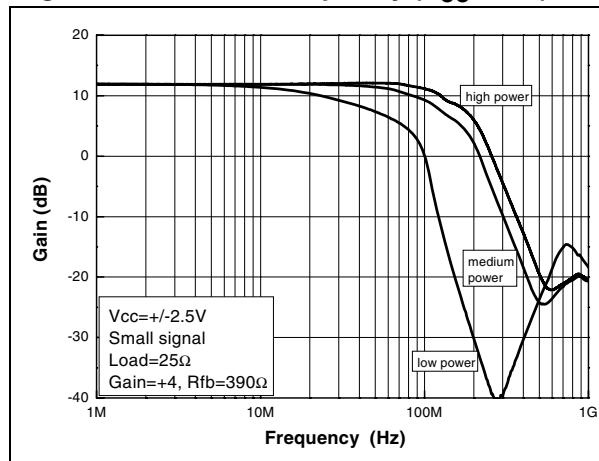
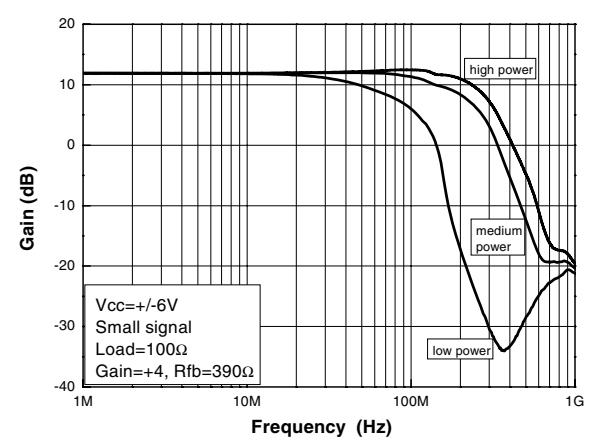
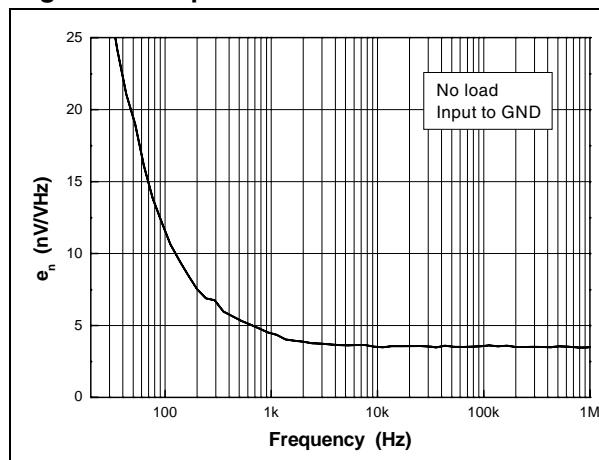
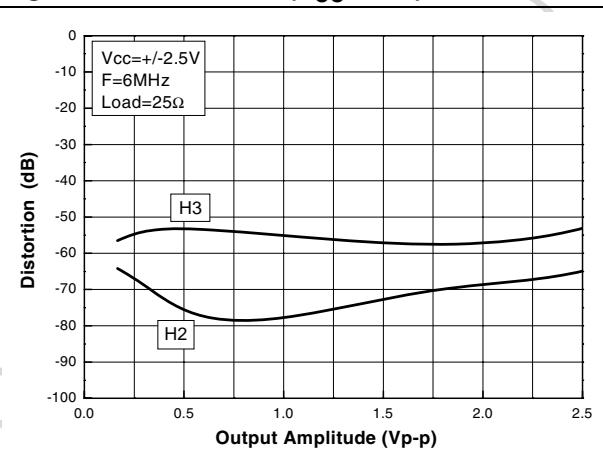
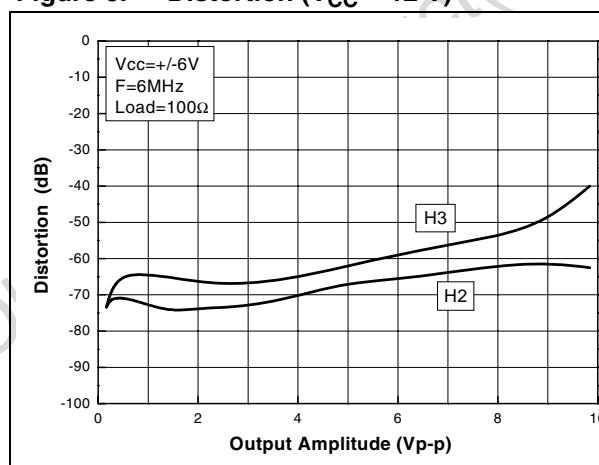
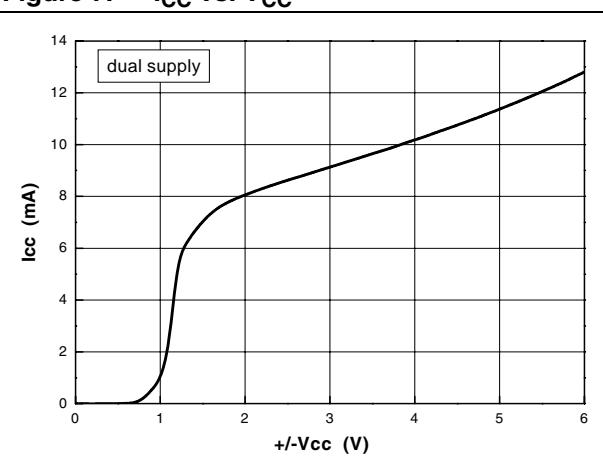
Figure 2. Gain vs. frequency ($V_{CC} = 5$ V)**Figure 3. Gain vs. frequency ($V_{CC} = 12$ V)****Figure 4. Input noise****Figure 5. Distortion ($V_{CC} = 5$ V)****Figure 6. Distortion ($V_{CC} = 12$ V)****Figure 7. I_{CC} vs. V_{CC}** 

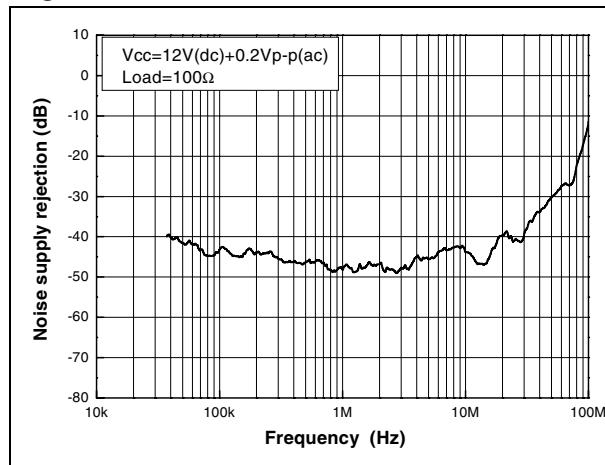
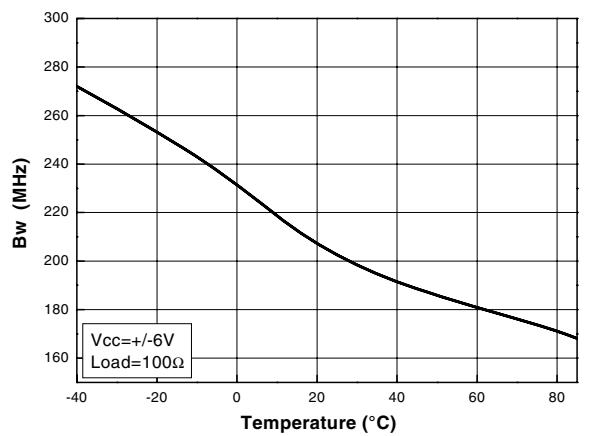
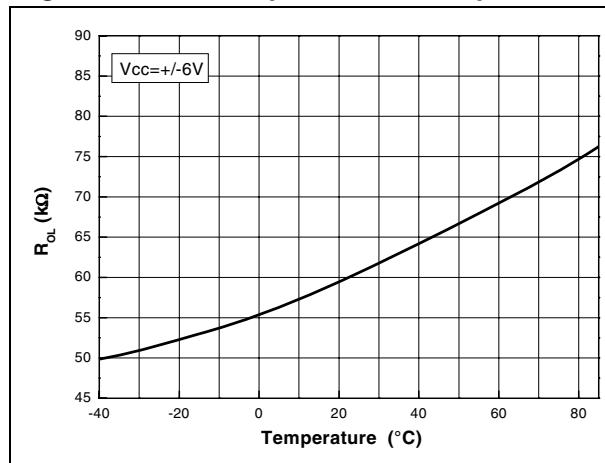
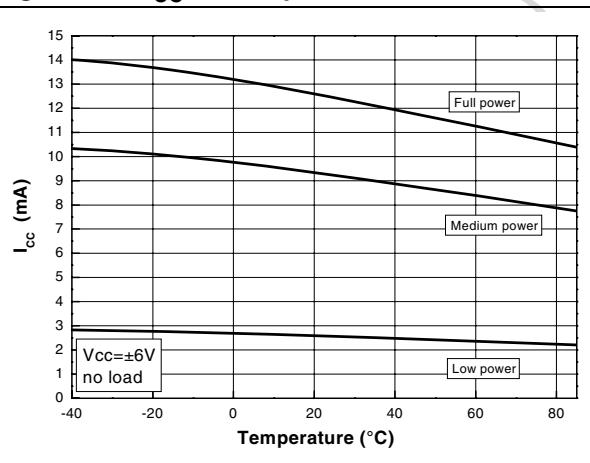
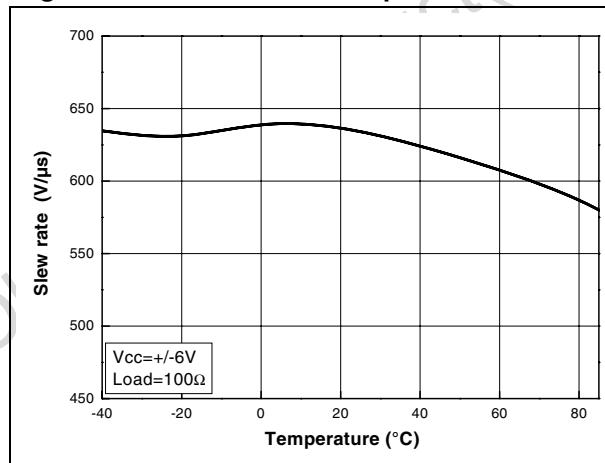
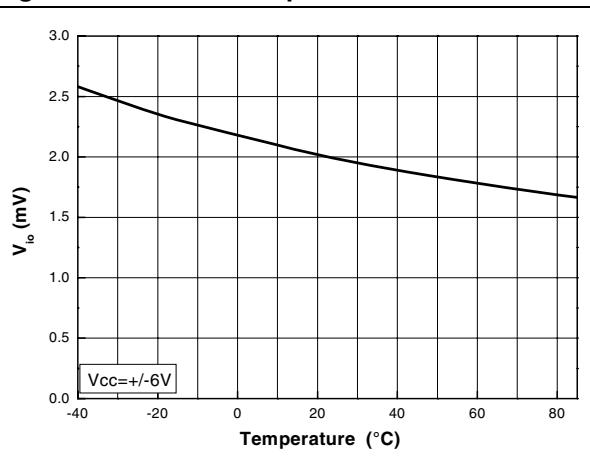
Figure 8. PSRR**Figure 9. Bandwidth vs. temperature****Figure 10. Transimpedance vs. temperature****Figure 11. I_{CC} vs. temperature****Figure 12. Slew rate vs. temperature****Figure 13. V_{io} vs. temperature**

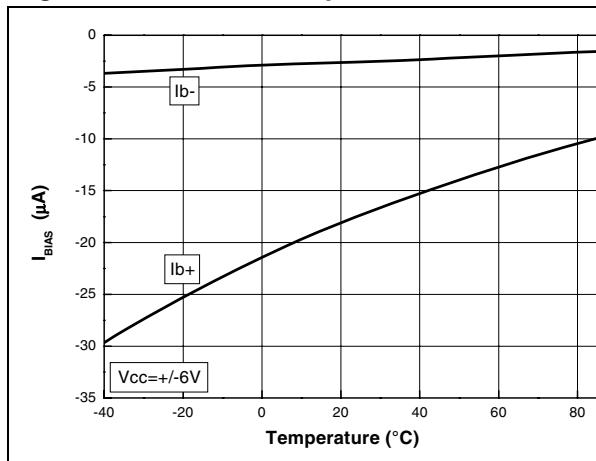
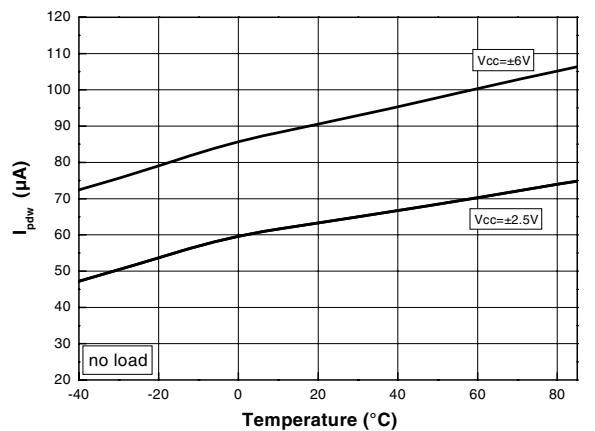
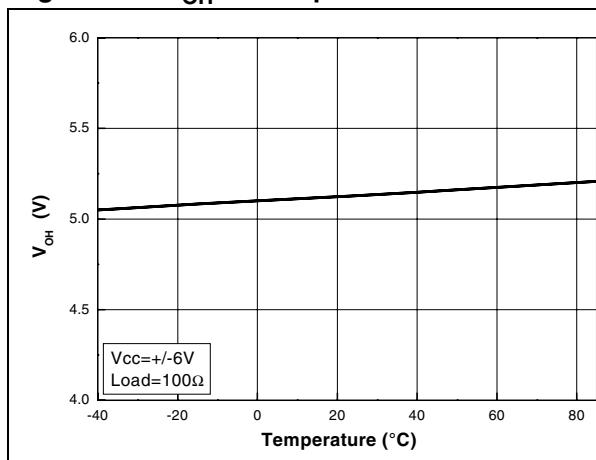
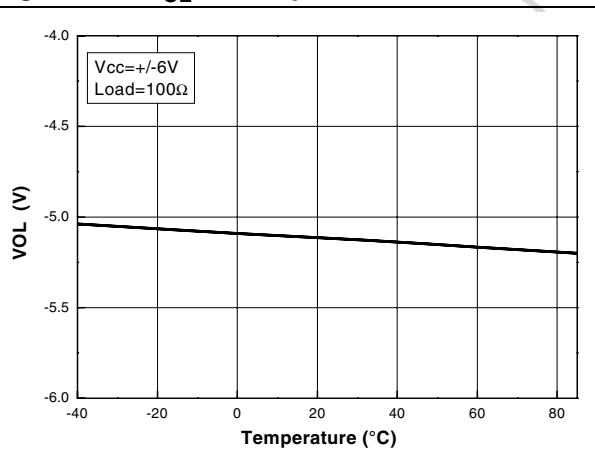
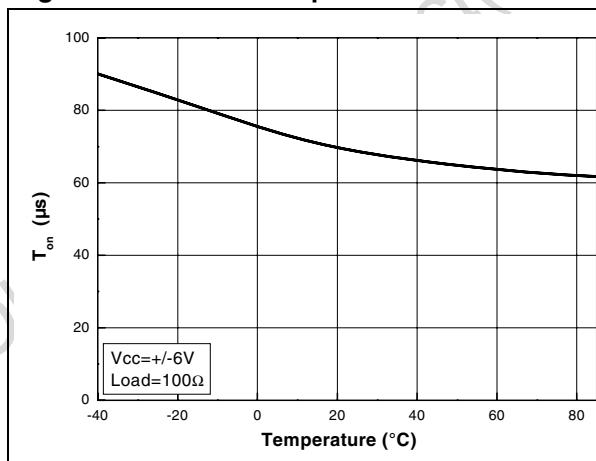
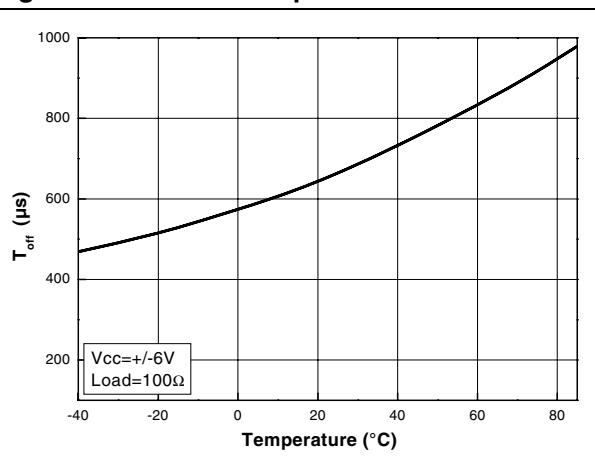
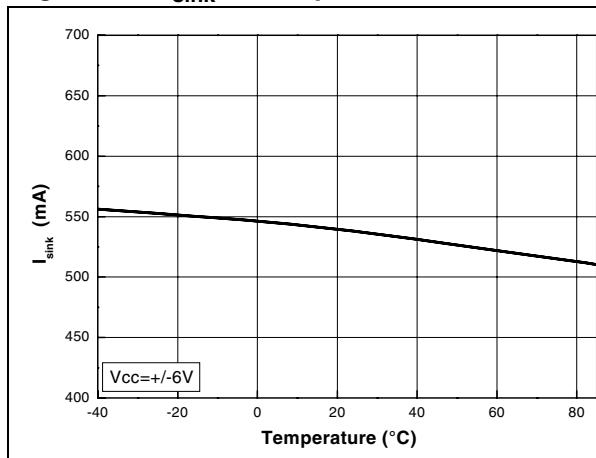
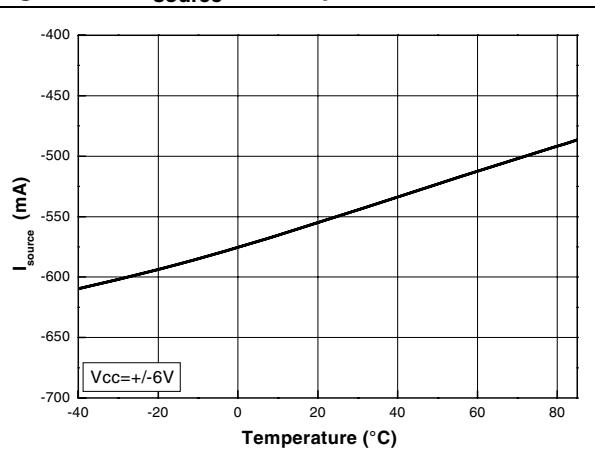
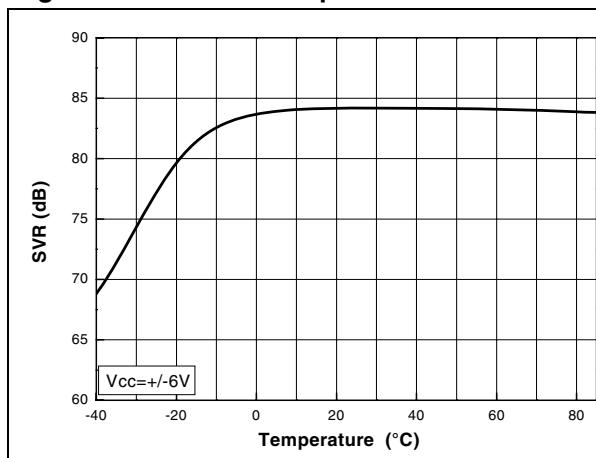
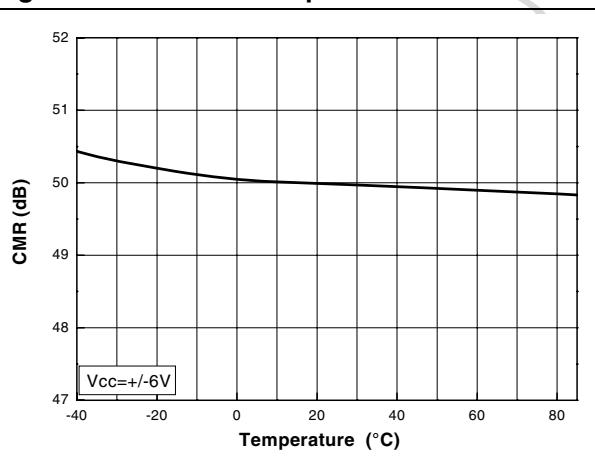
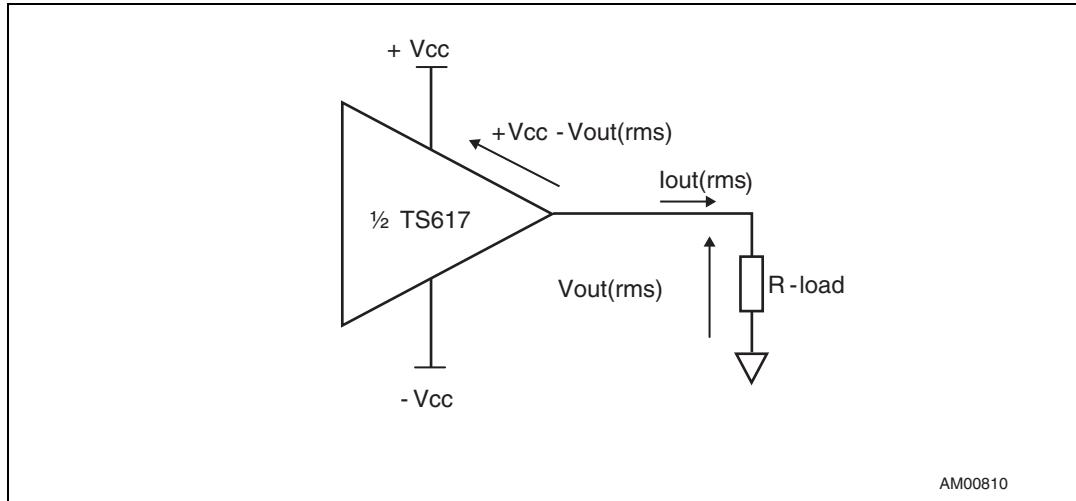
Figure 14. Ibias vs. temperature**Figure 15. Power down vs. temperature****Figure 16. V_{OH} vs. temperature****Figure 17. V_{OL} vs. temperature****Figure 18. Ton vs. temperature****Figure 19. Toff vs. temperature**

Figure 20. I_{sink} vs. temperature**Figure 21.** I_{source} vs. temperature**Figure 22.** SVR vs. temperature**Figure 23.** CMR vs. temperature

3 Safe operating area

Figure 24. Equivalent schematic



$V_{out(rms)}$ is the rms output voltage value. $I_{out(rms)}$ is the rms output current value through the output load R -load. When supplied by $+/-V_{cc}$, the power which must be dissipated by one operator is $P = I_{out(rms)} \times (+V_{cc} - V_{out(rms)})$. Since the TS617 is used in a differential configuration using both operators, it must dissipate $2 \times P$. The power dissipated by the TS617 is derived from:

(eq1),

$$P = 2 \times [I_{out(rms)} \times (+V_{cc} - V_{out(rms)})].$$

In the following formula:

(eq2),

$$P = \frac{(T_j - T_{amb})}{R_{thja}}$$

T_j is the junction temperature, T_{amb} is the ambient temperature and R_{thja} is the junction-to-ambient thermal resistance of the package (SO-14: $R_{thja} = 103^\circ C/watt$).

Assuming that T_j must be lower than $150^\circ C$ to avoid any damage to the dice, it is derived from (eq2):

(eq3),

$$P_{max} = \frac{150 - T_{amb}}{103}, \text{ (inWatt)}$$

From (eq1) and (eq3) we can easily extract the maximum value of $V_{out(rms)}$ that the TS617 can drive without any damage (according to T_{amb} and R -load).

4 Typical application

Figure 25. Power line interface

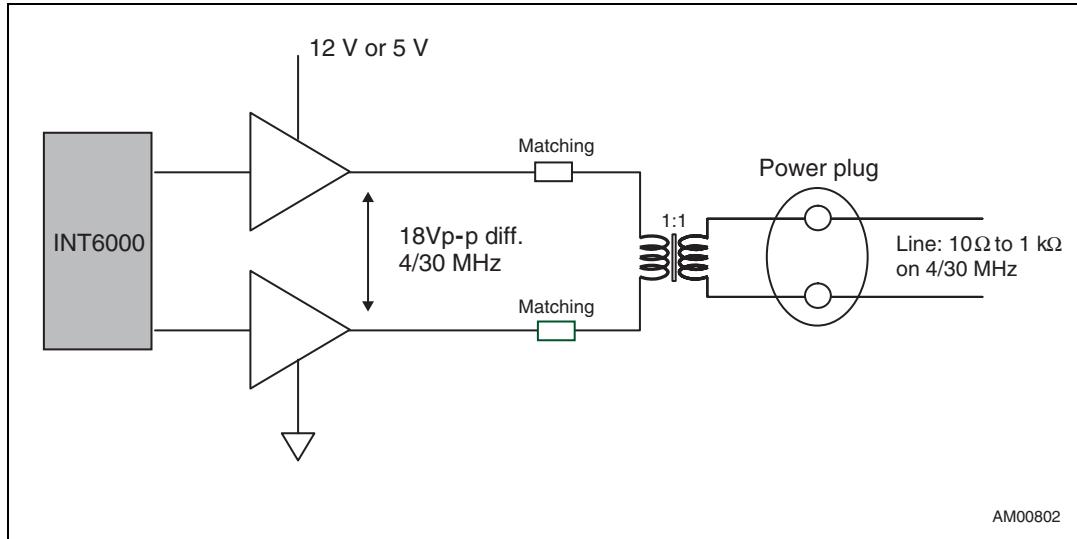
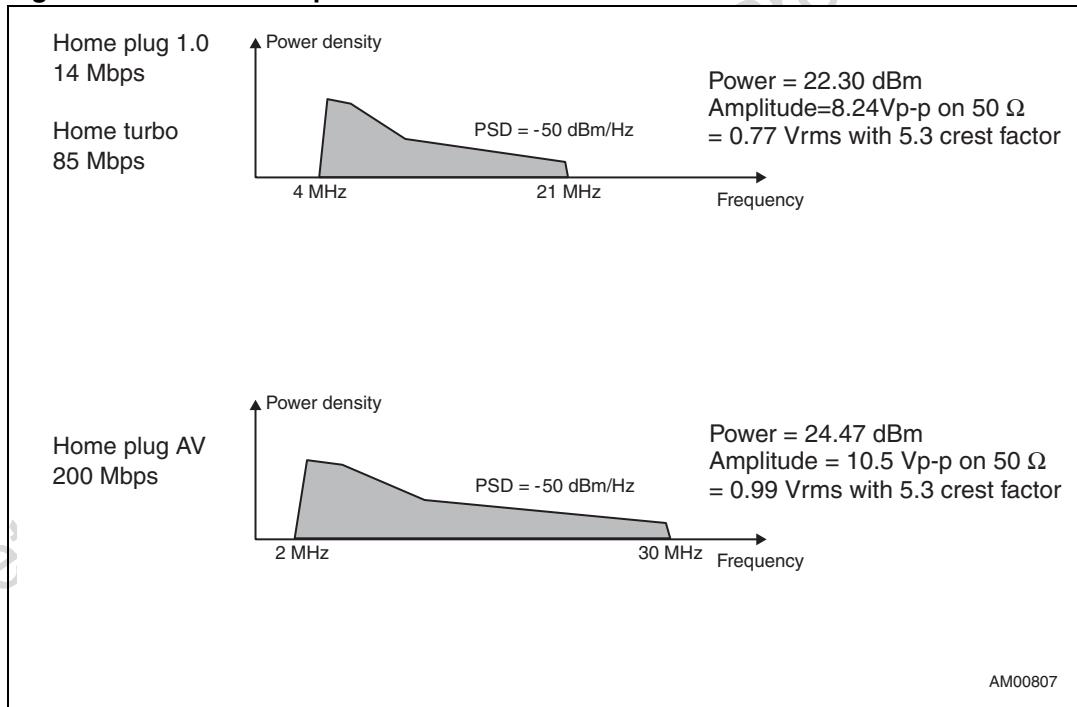
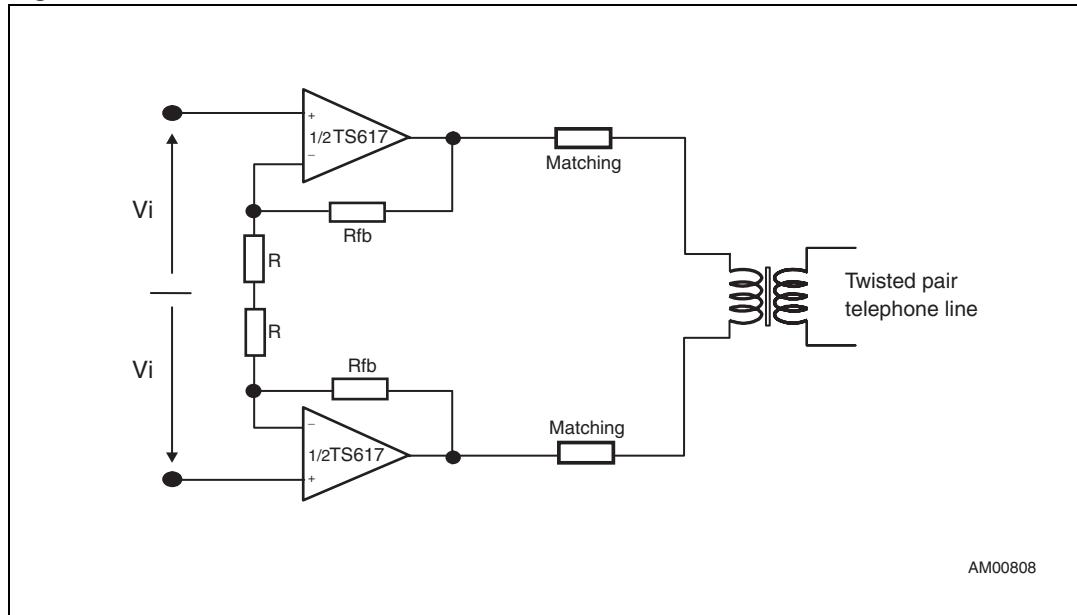
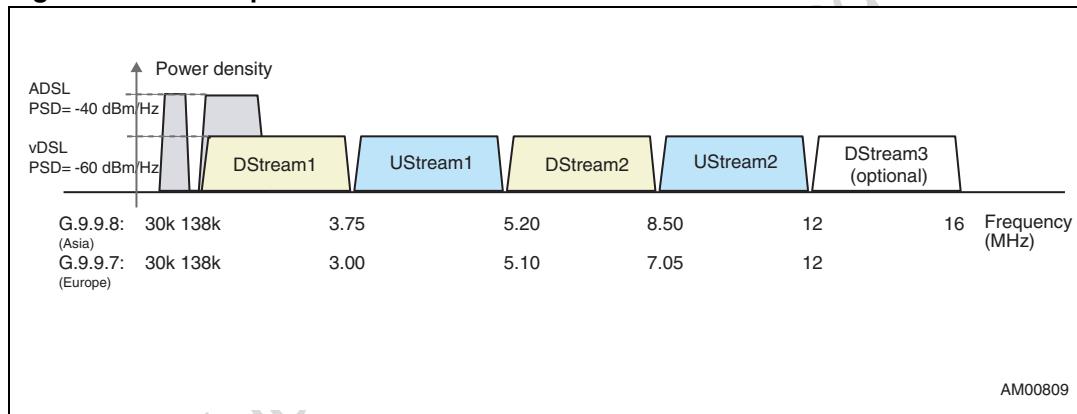


Figure 26. Power line spectrums



- Power on the line: $P(\text{dBm}) = 10 \log(\Sigma \text{Bw} \times 10^{\text{PSD}/10})$
- PSD: power spectrum density
- ΣBw : total bandwidth summation
- Power line: PSD = -50 dBm/Hz on 50 Ω

Figure 27. vDSL interface**Figure 28.** vDSL spectrum**Table 8.** vDSL spectrum

G. 9.9.8 Standard		G. 9.9.7 Standard	
Power (on 100- Ω line)	V _{p-p} (on 100- Ω line)	Power (on 100- Ω line)	V _{p-p} (on 100- Ω line)
CPE: 6.95 dBm	7.9 V _{p-p}	CPE: 8.48 dBm	9.4 V _{p-p}
CO: 8.4 dBm	9.3 V _{p-p}	CO: 6.82 dBm	7.8 V _{p-p}
CO DS3: 10.4 dBm	11.7 V _{p-p}	CO DS3: 9.45 dBm	10.5 V _{p-p}

5 Package information

In order to meet environmental requirements, STMicroelectronics offers these devices in ECOPACK® packages. These packages have a lead-free second level interconnect. The category of second level interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label. ECOPACK is an STMicroelectronics trademark. ECOPACK specifications are available at: www.st.com.

Obsolete Product(s) - Obsolete Product(s)

Figure 29. SO-14 package mechanical drawing

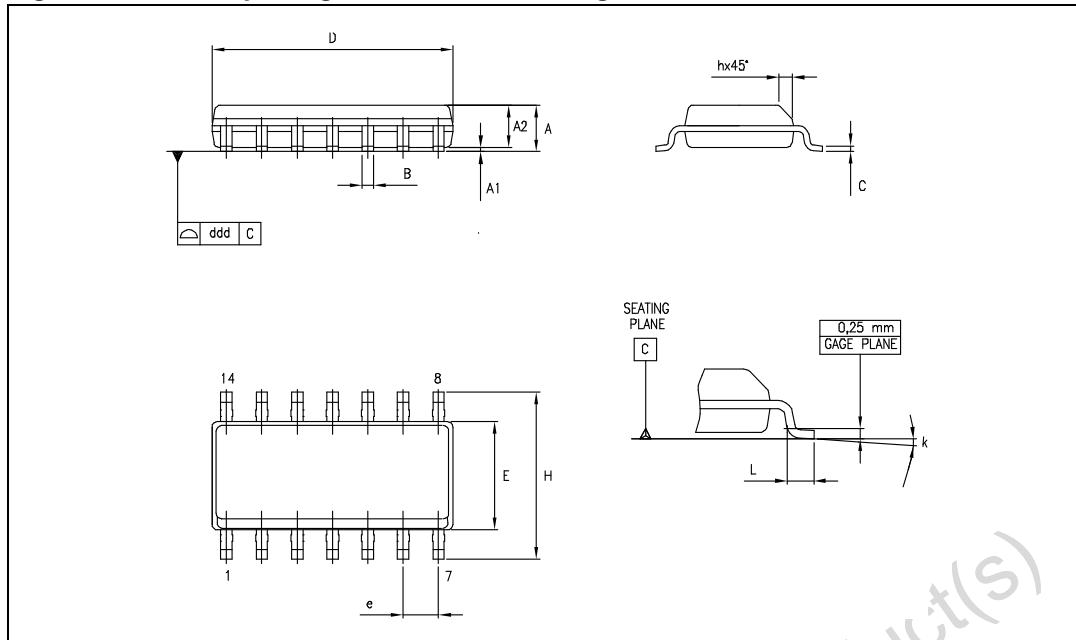


Table 9. SO-14 package mechanical data

Ref.	Dimensions			Inches		
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	1.35		1.75	0.05		0.068
A1	0.10		0.25	0.004		0.009
A2	1.10		1.65	0.04		0.06
B	0.33		0.51	0.01		0.02
C	0.19		0.25	0.007		0.009
D	8.55		8.75	0.33		0.34
E	3.80		4.0	0.15		0.15
e		1.27			0.05	
H	5.80		6.20	0.22		0.24
h	0.25		0.50	0.009		0.02
L	0.40		1.27	0.015		0.05
k	8° (max.)					
ddd			0.10			0.004

Note: D and F dimensions do not include mold flash or protrusions. Mold flash or protrusions must not exceed 0.15 mm.

6 Ordering information

Table 10. Order codes

Order code	Temperature range	Package	Packaging	Marking
TS617ID	-40°C to +85°C	SO-14	Tube	TS617I
TS617IDT			Tape & reel	

Obsolete Product(s) - Obsolete Product(s)

7 Revision history

Table 11. Document revision history

Date	Revision	Changes
03-Nov-2008	1	Initial release.

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