

L6561

POWER FACTOR CORRECTOR

1 FEATURES

- VERY PRECISE ADJUSTABLE OUTPUT OVERVOLTAGE PROTECTION
- MICRO POWER START-UP CURRENT (50µA TYP.)
- VERY LOW OPERATING SUPPLY CURRENT(4mA TYP.)
- INTERNAL START-UP TIMER
- CURRENT SENSE FILTER ON CHIP
- DISABLE FUNCTION
- 1% PRECISION (@ T_j = 25°C) INTERNAL REFERENCE VOLTAGE
- TRANSITION MODE OPERATION
- TOTEM POLE OUTPUT CURRENT: ±400mA
- DIP-8/SO-8 PACKAGES

2 **DESCRIPTION**

L6561 is the improved version of the L6560 standard Power Factor Corrector. Fully compatible with the standard version, it has a superior performant multiplier making the device capable of working in wide input voltage range applications (from 85V to 265V) with an excellent THD. Furthermore the start up current has been reduced at few tens of mA and a disable function has been implemented on the ZCD pin, guaranteeing lower current consumption in stand by mode.

Figure 2. Block Diagram

Figure 1. Packages



Table 1. Order Codes

Part Number	Package
L6561	DIP-8
L6561D	SO-8
L6561D013TR	Tape & Reel

Realised in mixed BCD technology, the chip gives the following benefits:

- micro power start up current
- 1% precision internal reference voltage
- $(Tj = 25^{\circ}C)$
- Soft Output Over Voltage Protection
- no need for external low pass filter on the current sense
- very low operating quiescent current minimises power dissipation

The totem pole output stage is capable of driving a Power MOS or IGBT with source and sink currents of ± 400 mA. The device is operating in transition mode and it is optimised for Electronic Lamp Ballast application, AC-DC adaptors and SMPS.



Table 2. Absolute Maximum Ratings

Symbol	Pin	Parameter	Value	Unit	
I _{Vcc}	8	$I_{q} + I_{Z}; (I_{GD} = 0)$	30	mA	
I _{GD}	7	Output Totem Pole Peak Current (2µs)	±700	mA	
INV, COMP MULT	1, 2, 3	Analog Inputs & Outputs	-0.3 to 7	V	
CS	4	Current Sense Input	-0.3 to 7	V	
ZCD	5	Zero Current Detector	50 (source) -10 (sink)	mA mA	
P _{tot}		Power Dissipation @T _{amb} = 50 °C (DIP-8) (SO-8)		1 0.65	W W
Tj		Junction Temperature Operating Range	-40 to 150	°C	
T _{stg}		Storage Temperature	-55 to 150	°C	

Figure 3. Pin Connection (Top view)



Table 3. Thermal Data

Symbol	Parameter	SO 8	MINIDIP	Unit
R _{th j-amb}	Thermal Resistance Junction to ambient	150	100	°C/W

Table 4. Pin Description

Ν.	Name	Function
1	INV	Inverting input of the error amplifier. A resistive divider is connected between the output regulated voltage and this point, to provide voltage feedback.
2	COMP	Output of error amplifier. A feedback compensation network is placed between this pin and the INV pin.
3	MULT	Input of the multiplier stage. A resistive divider connects to this pin the rectified mains. A voltage signal, proportional to the rectified mains, appears on this pin.
4	CS	Input to the comparator of the control loop. The current is sensed by a resistor and the resulting voltage is applied to this pin.
5	ZCD	Zero current detection input. If it is connected to GND, the device is disabled.
6	GND	Current return for driver and control circuits.
7	GD	Gate driver output. A push pull output stage is able to drive the Power MOS with peak current of 400mA (source and sink).
8	V _{CC}	Supply voltage of driver and control circuits.

57

(1) Parameter guaranteed by design, not tested in production.

Table 5. Electrical Characteristics

(V_{CC} = 14.5V; T_{amb} = -25°C to 125°C;unless otherwise specified)

Symbol	Pin	Parameter	Test Condition	Min.	Тур.	Max.	Unit
SUPPLY	VOLTA	GE SECTION	1		-		
V _{CC}	8	Operating Range	after turn-on	11		18	V
V _{CC ON}	8	Turn-on Threshold		11	12	13	V
VCC OFF	8	Turn-off Threshold		8.7	9.5	10.3	V
Hys	8	Hysteresis		2.2	2.5	2.8	V
SUPPLY	CURR	ENT SECTION					
ISTART-U	8	Start-up Current	before turn-on (V _{CC} =11V)	20	50	90	μΑ
lq	8	Quiescent Current			2.6	4	mA
Icc	8	Operating Supply Current	C _L = 1nF @ 70KHz		4	5.5	mA
			in OVP condition V _{pin1} = 2.7V		1.4	2.1	mA
lq	8	Quiescent Current	V _{PIN5} ≤150mV, V _{CC} > V _{CC off}		1.4	2.1	mA
	8		$V_{PIN5} \le 150 mV$, $V_{CC} < V_{CC off}$	20	50	90	μA
Vz	8	Zener Voltage	$I_{CC} = 25 \text{mA}$	18	20	22	·
	MPLI				1		
VINV	1	Voltage Feedback Input	T _{amb} = 25°C	2.465	2.5	2.535	V
		Threshold	12V < V _{CC} < 18V	2.44		2.56	V
		Line Regulation	V _{CC} = 12 to 18V		2	5	mV
I _{INV}	1	Input Bias Current			-0.1	-1	μA
G _V		Voltage Gain	Open loop	60	80		dB
GB		Gain Bandwidth			1		MHz
I _{COMP} 2		Source Current	$V_{COMP} = 4V, V_{INV} = 2.4V$	-2	-4	-8	mA
		Sink Current	$V_{COMP} = 4V, V_{INV} = 2.6V$	2.5	4.5		mA
VCOMP	2	Upper Clamp Voltage	I _{SOURCE} = 0.5mA	-	5.8		V
00111		Lower Clamp Voltage	$I_{Sink} = 0.5 mA$		2.25		V
MULTIPL	IFR SP	· · ·	-Onk				
VMULT	3	Linear Operating Voltage		0 to 3	0 to 3.5		V
	0	Output Max. Slope	V _{MULT} = from 0V to 0.5V	1.65	1.9		
$\frac{\Delta V_{CS}}{\Delta V_{mult}}$			V_{COMP} = Upper Clamp Voltage	1.00	1.0		
К		Gain	$V_{MULT} = 1V V_{COMP} = 4V$	0.45	0.6	0.75	1/V
	1	SE COMPARATOR			<u> </u>		
V _{CS}	4	Current Sense Reference Clamp	$V_{MULT} = 2.5V$ $V_{COMP} = Upper Clamp Voltage$	1.6	1.7	1.8	V
I _{CS}	4	Input Bias Current	$V_{OS} = 0$		-0.05	-1	μΑ
t _{d (H-L)}	4	Delay to Output			200	450	ns
	4	Current Sense Offset			0	15	mV
ZERO CU	RREN	T DETECTOR					
V _{ZCD}	5	Input Threshold Voltage Rising Edge	(1)		2.1		V
		Hysteresis	(1)	0.3	0.5	0.7	V
V _{ZCD}	5	Upper Clamp Voltage	$I_{ZCD} = 20\mu A$	4.5	5.1	5.9	V
V _{ZCD}	5	Upper Clamp Voltage	I _{ZCD} = 3mA	4.7	5.2	6.1	V

<u> </u>	

Table 5. Electrical Characteristics (continued)

(V_{CC} = 14.5V; T_{amb} = -25°C to 125°C;unless otherwise specified)

Symbol	Pin	Parameter	Test Condition	Min.	Тур.	Max.	Unit
V _{ZCD}	5	Lower Clamp Voltage	I _{ZCD} = -3mA	0.3	0.65	1	V
I _{ZCD}	5	Sink Bias Current	$1V \le V_{ZCD} \le 4.5V$		2		μA
I _{ZCD}	5	Source Current Capability		-3		-10	mA
I _{ZCD}	5	Sink Current Capability		3		10	mA
V _{DIS}	5	Disable threshold		150	200	250	mV
I _{ZCD}	5	Restart Current After Disable	V _{ZCD} < V _{dis} ; V _{CC} > V _{CCOFF}	-100	-200	-300	μA
OUTPUT	SECT	ION			•		
V_{GD}	7	Dropout Voltage	I _{GDsource} = 200mA		1.2	2	V
			I _{GDsource} = 20mA		0.7	1	V
			I _{GDsink} = 200mA			1.5	V
			I _{GDsink} = 20mA			0.3	V
tr	7	Output Voltage Rise Time	$C_L = 1nF$		40	100	ns
t _f	7	Output Voltage Fall Time	$C_L = 1nF$		40	100	ns
I _{GD off}	7	IGD Sink Current	V _{CC} =3.5V V _{GD} = 1V	5	10	-	mA
OUTPUT	OVER	VOLTAGE SECTION			•		
I _{OVP}	2	OVP Triggering Current		35	40	45	μA
		Static OVP Threshold		2.1	2.25	2.4	V
RESTART	ТІМЕ	R	-			•	
t _{START}		Start Timer		70	150	400	μs

3 OVER VOLTAGE PROTECTION OVP

The output voltage is expected to be kept by the operation of the PFC circuit close to its nominal value. This is set by the ratio of the two external resistors R1 and R2 (see fig. 5), taking into consideration that the non inverting input of the error amplifier is biased inside the L6561 at 2.5V.

In steady state conditions, the current through R1 and R2 is:

$$I_{R1sc} = \frac{V_{out} - 2.5}{R1} = I_{R2} = \frac{2.5V}{R2}$$

and, if the external compensation network is made only with a capacitor C_{comp} , the current through C_{comp} equals zero. When the output voltage increases abruptly the current through R1 becomes:

$$I_{R1} = \frac{V_{outsc} + \Delta V_{out} - 2.5}{R1} = I_{R1sc} + \Delta I_{R1}$$

Since the current through R2 does not change, ΔI_{R1} must flow through the capacitor C_{comp} and enter the error amplifier.

This current is monitored inside the L6561 and when reaches about 37μ A the output voltage of the multiplier is forced to decrease, thus reducing the energy drawn from the mains. If the current exceeds 40μ A, the OVP protection is triggered (Dynamic OVP), and the external power transistor is switched off until the current falls approximately below 10μ A.

However, if the overvoltage persists, an internal comparator (Static OVP) confirms the OVP condition keeping the external power switch turned off (see fig. 4). Finally, the overvoltage that triggers the OVP function is:

$$\Delta V_{out} = R_1 \cdot 40 \mu A.$$

Typical values for R1, R2 and C are shown in the application circuits. The overvoltage can be set indepen-



dently from the average output voltage. The precision in setting the overvoltage threshold is 7% of the overvoltage value (for instance $\Delta V = 60V \pm 4.2V$).

3.1 Disable function

The zero current detector (ZCD) pin can be used for device disabling as well. By grounding the ZCD voltage the device is disabled reducing the supply current consumption at 1.4mA typical (@ 14.5V supply voltage).

Releasing the ZCD pin the internal start-up timer will restart the device.

Figure 4.



Figure 5. Overvoltage Protection Circuit





Figure 6. Typical Application Circuit (80W, 110VAC)





Figure 8. Typical Application Circuit (80W, Wide-range Mains)





Figure 9. Demo Board (EVAL6561-80) Electrical Schematic

Figure 10. EVAL6561-80: PCB and Component Layout (Top view, real size 57x108mm)



V _{in} (Vac)	Din (14/)	V (Vdo)		Bo (141) m (9(1)	w/o THD	reducer	with THE) reducer	
	Pin (W)	V _o (Vdc)	∆Vo (Vdc)	Po (W)	Po (W) ղ (%)	PF	THD (%)	PF	THD (%)
85	87.2	400.1	14	80.7	92.8	0.999	3.7	0.999	2.9
110	85.2	400.1	14	80.7	94.7	0.996	5.0	0.996	3.2
135	84.2	400.1	14	80.7	95.8	0.989	6.2	0.989	3.7
175	83.5	400.1	14	80.7	96.6	0.976	8.3	0.976	4.3
220	83.1	400.1	14	80.7	97.1	0.940	10.7	0.941	5.6
265	82.9	400.1	14	80.7	97.3	0.890	13.7	0.893	8.1



Figure 11. OVP Current Threshold vs. Temperature





Figure 12. Undervoltage Lockout Threshold vs. Temperature



Figure 14. Voltage Feedback Input Threshold vs. Temperature





Figure 15. Output Saturation Voltage vs. Sink Current





Figure 16. Output Saturation Voltage vs. Source Current



L6561



57

Figure 18. DIP-8 Mechanical Data & Package Dimensions



Figure 19. SO-8 Mechanical Data & Package Dimensions

L6561

Table 7. Revision History

Date	Revision	Description of Changes
January 2004	15	First Issue
June 2004	16	Modified the Style-look in compliance with the "Corporate Technical Publications Design Guide". Changed input of the power amplifier connected to Multiplier (Fig. 2).



Information furnished is believed to be accurate and reliable. However, STMicroelectronics assumes no responsibility for the consequences of use of such information nor for any infringement of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of STMicroelectronics. Specifications mentioned in this publication are subject to change without notice. This publication supersedes and replaces all information previously supplied. STMicroelectronics products are not authorized for use as critical components in life support devices or systems without express written approval of STMicroelectronics.

The ST logo is a registered trademark of STMicroelectronics. All other names are the property of their respective owners

© 2004 STMicroelectronics - All rights reserved

STMicroelectronics GROUP OF COMPANIES

Australia - Belgium - Brazil - Canada - China - Czech Republic - Finland - France - Germany - Hong Kong - India - Israel - Italy - Japan -Malaysia - Malta - Morocco - Singapore - Spain - Sweden - Switzerland - United Kingdom - United States www.st.com

X-ON Electronics

Largest Supplier of Electrical and Electronic Components

Click to view similar products for Power Factor Correction - PFC category:

Click to view products by STMicroelectronics manufacturer:

Other Similar products are found below :

L6564TD FAN7930MX L4984D NCP1612A1DR2G NCP1618ADR2G TDA4863GXUMA2 ICE2PCS06GXUMA1 PFS7624C-TL TDA48632GXUMA2 UCC29910APWR DCRG8 UCC2818AQDRQ1 L6561D013TR L6566B ICE3PCS03G AL1788W6-7 LT1249CN8#PBF NCP1616A1DR2G LT1509CN#PBF BR6000-R12 ICE3PCS02GXUMA1 DCRL5 B44066R6012E230 ICE2PCS01GXUMA1 ADP1048ARQZ-R7 LT1249IS8#PBF LT1249CS8#PBF AIF06ZPFC-01L FAN4800AUM FAN6920MRMY FAN4801SMY ICE1PCS01G NCP1652DR2G ICE2PCS05GXUMA1 ICE2PCS01G ICE2PCS06G ICE3PCS01G ICE3PCS01GXUMA1 TDA4863 TDA4863-2 TDA4863-2G XDPL8210XUMA1 ISL6730AFUZ ISL6731BFBZ NCP1615A1DR2G NCP1605DR2G AP1682EMTR-G1 AL6562AS-13 AL6562S-13 LT1509CSW#PBF