

AC/DC Convertor IC

PWM Type DC/DC Converter IC Built-in Switching MOSFET for Non-Isolated Type

BM2P104Q-Z BM2P107QK-Z

General Description

The PWM type DC/DC converter for AC/DC provides an optimum system for all products that include an electrical outlet. It enables simpler design of a high effective converter specializing in non-isolated devices. This series has a built-in starter circuit that tolerates 730 V / 800 V, and it contributes to low power consumption. With a current detection resistor for switching as internal device, it can be designed as small power supply. Since current mode control is utilized, current is restricted in each cycle and excellent performance is demonstrated in bandwidth and transient response. The oscillation frequency is fixed to 100 kHz A frequency hopping function is also on chip, and it contributes to low EMI. In addition, a built-in super junction MOSFET which tolerates 730 V / 800 V makes the design easy.

Features

- PWM Current Mode Method
- Frequency Hopping Function
- Burst Operation at Light Load
- Built-in 730 V / 800 V Starter Circuit that Tolerates
- Built-in 730 V / 800 V Super Junction MOSFET
- VCC Pin Under Voltage Protection
- VCC Pin Over Voltage Protection
- Over Current Limiter Function per Cycle
- Soft Start Function

Key Specifications

Power Supply Voltage Operation Range VCC: 8.00 V to 10.81 V DRAIN: 730 V / 800 V(Max)
 Pulse Operation Current 1.20 mA(Typ)
 Burst Operation Current 0.45 mA(Typ)

Burst Operation Current
 Oscillation Frequency
 100 kHz(Typ)

■ Operation Temperature Range -40 °C to +105 °C

MOSFET ON Resistor
BM2P104Q-Z:
BM2P107QK-Z:
4.0 Ω(Typ)
7.5 Ω(Typ)

Package DIP7K

W(Typ) x D(Typ) x H(Max)

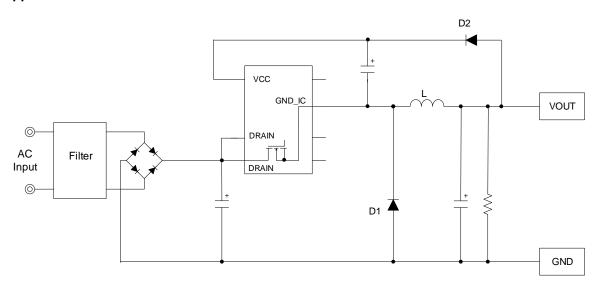
9.27 mm x 6.35 mm x 8.63 mm pitch 2.54 mm



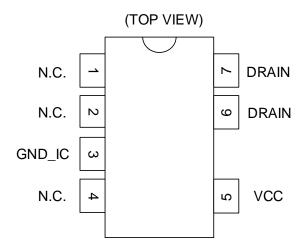
Applications

LED Lights, Air Conditioners, Cleaners etc.

Typical Application Circuit



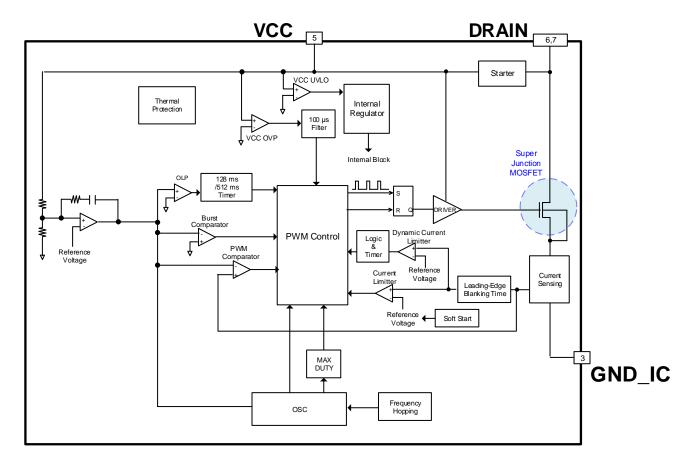
Pin Configuration



Pin Descriptions

diolio									
Din No Din Nome	1/0	Function	ESD Diode						
PIII NO.	Pin No. Pin Name I/C	I/O	Function	VCC	GND_IC				
1	N.C.	-	Non Connection	-	-				
2	N.C.	-	Non Connection	-	-				
3	GND_IC	I/O	GND pin	0	-				
4	N.C.	-	Non Connection	-	-				
5	VCC	I	Power supply input pin	•	0				
6	DRAIN	I/O	MOSFET DRAIN pin	-	0				
7	DRAIN	I/O	MOSFET DRAIN pin	-	0				

Block Diagram



Description of Blocks

1 Back Converter

This is the IC for exclusive use of non-isolated type back converter.

<Basic operation of back converter>

1.1 When the MOSFET for Switching is ON

When the MOSFET turns ON, current I_L flows to coil L and energy is stored. At this moment, the voltage of the GND_IC pin becomes the voltage near the DRAIN pin, and the diode D1 is OFF.

$$I_L = \frac{(V_{IN} - V_{OUT})}{L} \times ton$$

Where:

 $egin{array}{ll} V_{IN} & \mbox{is the DRAIN Voltage} \\ V_{OUT} & \mbox{is the Output Voltage} \\ I_L & \mbox{is the Inductor Current} \\ ton & \mbox{is ON-Time of MOSFET} \\ \end{array}$

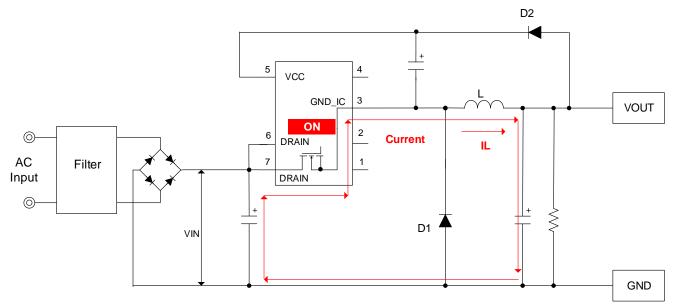


Figure 1. Back Converter Operation (MOSFET=ON)

1 Back Converter - continued

1.2 When the MOSFET for Switching is OFF

When the MOSFET turns OFF, the energy stored in coil is output via diode. At the moment, the MOSFET is OFF.

$$I_L = \frac{V_{OUT}}{L} \times toff$$

Where:

 $egin{array}{ll} V_{OUT} & ext{is the Output Voltage} \\ I_L & ext{is the Inductor Current} \\ toff & ext{is OFF-Time of MOSFET} \\ \end{array}$

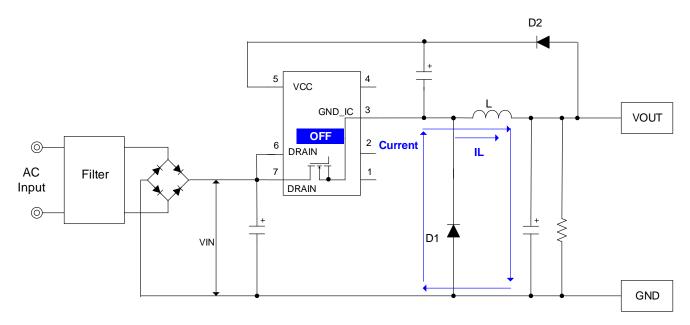


Figure 2. Back Converter Operation (MOSFET=OFF)

2 Start Sequences

Start sequences are shown in Figure 3. See the sections below for detailed descriptions.

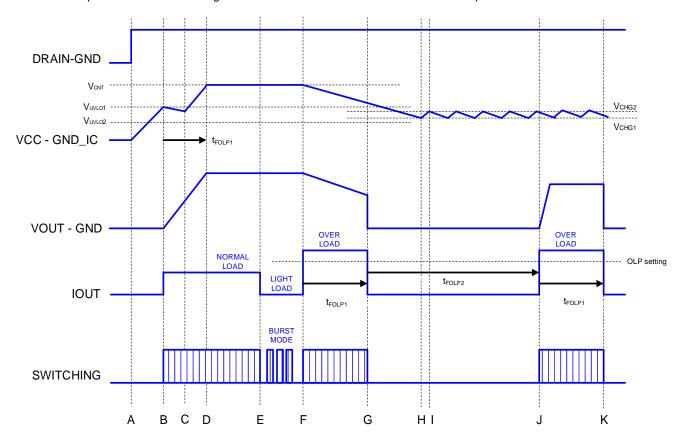


Figure 3. Start Sequences Timing Chart

- A: The input voltage is applied to the DRAIN pin and the VCC pin voltage rises.
- B: If the VCC pin voltage exceeds V_{UVLO1}, the IC starts to operate. And if the IC judges the other protection functions as normal condition, it starts switching operation.
 - The soft start function limits the over current limiter value to prevent any excessive voltage or current rising. When the switching operation starts, the VOUT rises.
- C: Until the VOUT becomes constant value from starting-up, the VCC pin voltage drops by the VCC pin consumption
- D: After switching starts, it is necessary that the output voltage is set to rating voltage within tFOLP1 (128 ms Typ).
- E: At light load, the IC starts burst operation to restrict the consumption power.
- F: When the load exceeds a certain electric power, the IC starts over load operation.
- G: If the setting over load status lasts for t FOLP1 (128 ms Typ), switching is turned OFF.
- H: When the VCC pin voltage becomes less than V_{CHG1}, recharge operation is started.
- I: When the VCC pin voltage becomes more than V_{CHG2}, recharge operation is stopped.
- J: After tFOLP2 (512 ms Typ), the over load protection circuit starts switching.
- K: Same as G.

3 Stop Sequences

Stop sequences are shown in Figure 4.

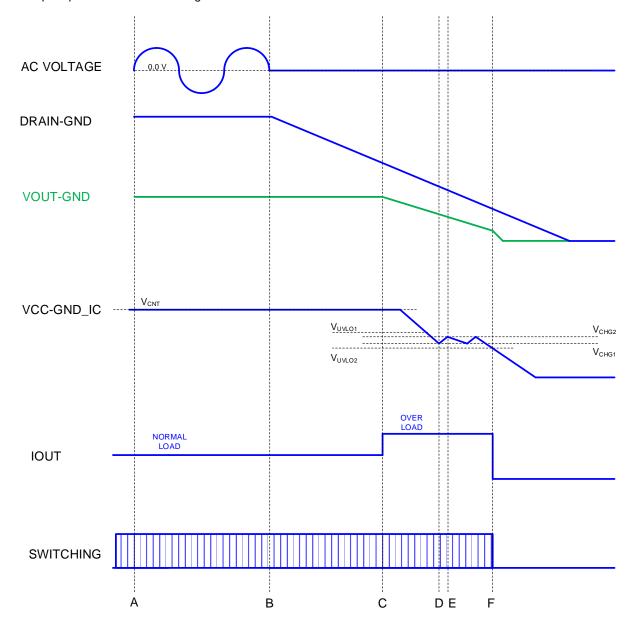


Figure 4. Stop Sequences Timing Chart

- A: Normal operation
- B: The input AC voltage is stopped. The DRAIN voltage starts to drop.
- C: If the DRAIN voltage drops below a certain voltage, it becomes maximum duty and over load protection operates.
- D: If the output voltage drops, the VCC pin voltage drops too. And recharge operation is started.
- E: The recharge operation is stopped.
- F: If the DRAIN voltage drops below a certain voltage, the VCC pin voltage lowers UVLO or less in order to stop recharge operation.

4 Start Circuit

This IC enables low standby electric power and high-speed startup because it has a built-in start circuit. The consumption current after startup is only idling current I_{START3} (Typ=10 µA). The startup current flows from the DRAIN pin.

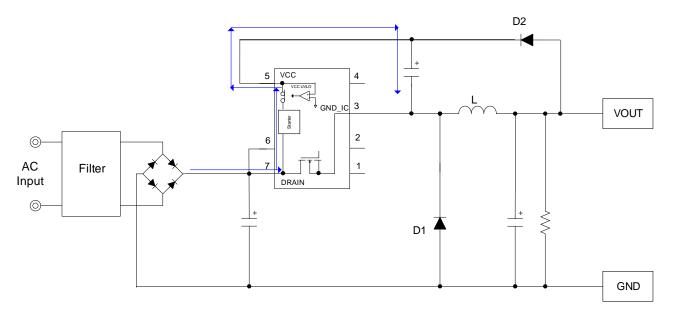


Figure 5. Start Circuit

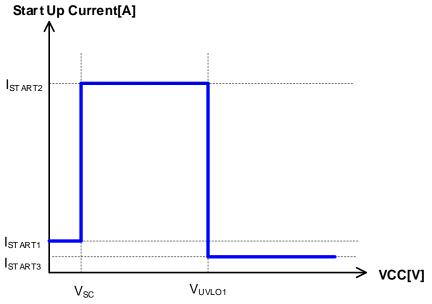


Figure 6. Startup Current vs VCC Voltage

5 The VCC Pin Protection Function

This IC has the internal protection function at the VCC pin shown in below.

- 1) Under voltage protection function UVLO
- 2) Over voltage protection function VCC OVP
- 3) VCC recharge function

5.1 VCC UVLO / VCC OVP Function

VCC UVLO function and VCC OVP function are auto recovery type comparators that have voltage hysteresis. VCC OVP has an internal mask time. If the condition that the VCC pin voltage is higher than VovP1 lasts for tcomp (100 µs Typ), it performs detection. The recovery requirements are that the VCC pin voltage is lower than VovP2.

5.2 VCC Recharge Function

If the VCC $p\bar{l}n$ drops to V_{CHG1} after once the VCC pin becomes more than V_{UVLO1} and the IC starts to operate, the VCC charge function operates. At that time, the VCC pin is charged from the DRAIN pin through start circuit. When the VCC pin voltage is more than V_{CHG2} , charge is stopped.

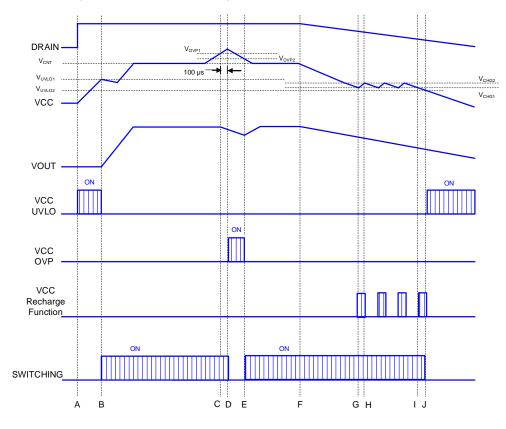


Figure 7. VCC UVLO / VCC OVP / VCC Recharge Function Timing Chart

- A: The input voltage is applied to the DRAIN pin and the VCC pin voltage rises.
- B: When the VCC pin voltage becomes higher than V_{UVLO1}, the IC starts operating. And if the IC judges the other protection functions as normal condition, it starts switching operation. The soft start function limits the over current limiter value to prevent any excessive voltage or current rising. When the switching operation starts, the VOUT rises.
- C: When the VCC pin voltage becomes higher than V_{OVP1}, VCC OVP timer operates.
- D: When the condition that the VCC pin voltage is higher than V_{OVP1} lasts for t_{COMP} (100 µs Typ), the IC detects VCC OVP and stops switching.
- E: When the VCC pin voltage becomes lower than V_{OVP2}, VCC OVP is released.
- F: When the input power supply is turned OFF, the DRAIN pin voltage drops.
- G: When the VCC pin voltage becomes less than V_{CHG1}, recharge function is started.
- H: When the VCC pin voltage becomes higher than V_{CHG2}, recharge function is stopped.
- When the VCC pin voltage becomes lower than V_{CHG1}, recharge function is started. However, the supply to the VCC pin decrease and the VCC pin voltage drops because of low DRAIN voltage.
- J: When the VCC pin voltage becomes lower than V_{UVLO2}, VCC UVLO function starts operating.

6 DC/DC Driver

This performs current mode PWM control. An internal oscillator sets a fixed oscillation frequency f_{SW} (100 kHz Typ). This IC has a built-in oscillation frequency hopping function. The maximum duty is D_{MAX} (75 % Typ). To achieve the low consumption power at light load, it also has an internal burst mode circuit.

6.1 Setting of the Output Voltage

Adopting the non-isolated type without photo coupler, the VCC voltage should be set to rating value. VCC Voltage means the voltage between the VCC pin and the GND_IC pin. The output voltage VOUT is defined by the formula below. The voltage when the MOSFET is OFF is shown in Figure 8.

$$V_{OUT} = V_{CNT} + V_{FD2} - V_{FD1}$$

Where

 V_{FD1} is the forward voltage of diode D1.

 V_{FD2} is the forward voltage of diode D2.

 V_{CNT} is the VCC Control Voltage

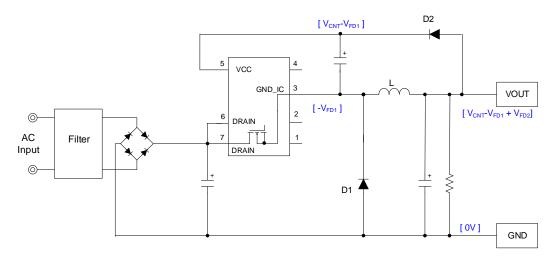


Figure 8. Back Converter Circuit (At MOSFET Turned OFF)

At light load, the output voltage may rise because the VCC voltage is difference from the output voltage. In this case, it is necessary that the output pin is connected to resistor and the voltage should be lowered. The circuit diagram is shown in Figure 9.

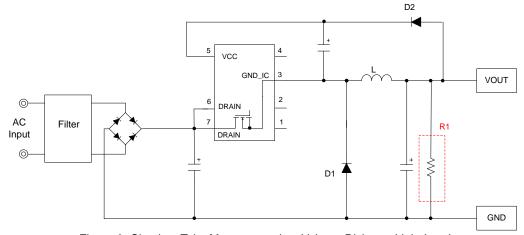


Figure 9. Circuit to Take Measure against Voltage Rising at Light Load

6.1 Setting of the Output Voltage - continued

This IC has a few external parts by fixing the VCC voltage and it enables simpler design. If you adjust the output voltage, it can become the variable voltage by adding zener diodes. However, it is necessary to consider the dispersion of the diodes.

The output voltage VOUT is defined by the formula below. The voltage when the MOSFET is OFF is shown in Figure 10.

$$V_{OUT} = V_{CNT} + V_{FD2} - V_{FD1} + V_{ZD1}$$

Where

 V_{FD1} is the forward voltage of diode D1. V_{FD2} is the forward voltage of diode D2.

 V_{ZD1} is the zener diode ZD1 voltage.

 V_{CNT} is the VCC Control Voltage

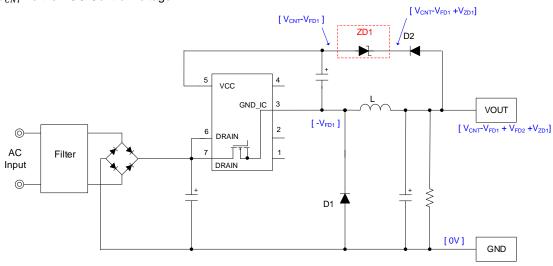


Figure 10. Back Converter Output Dispersion Circuit (At MOSFET Turned OFF)

6.2 Frequency Circuit

mode 1: burst operation

mode 2: fixed frequency operation (It operates in maximum frequency.)

mode 3: over load operation (pulse operation is stopped and burst operation is started.)

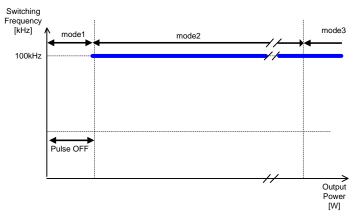


Figure 11. State Transition of Oscillation Frequency

6.3 Frequency Hopping Function

Frequency hopping function achieves low EMI by change the frequency at random. The wave width of frequency's upper limit is ± 6 % for basic frequency.

6 DC/DC Driver - continued

6.4 PWM Error Amp and PWM Comparator

The internal error Amp achieves the reduction of external parts. In addition, this IC adopts current mode method. It makes the design easy.

6.5 Over Current Limiter

This IC has an internal over current limiter per switching cycle. This function monitors the coil current and if it exceeds a certain current, the IC stops switching. Additionally, an internal current detection resistor contributes to reduction of parts and improvement of efficiency. The peak current by which the IC switches to the over load mode is determined by the formula below.

$$Peak \ current = I_{PEAK} + \frac{(V_{DRAIN} - V_{OUT})}{L} \times tdelay$$

Where:

 I_{PEAK} is the over current limiter internal the IC.

 $V_{\it DRAIN}$ is the DRAIN voltage.

 V_{OUT} is the output voltage.

L is the Coil value.

tdelay is the Delay time after detection of over current limiter.

6.6 Dynamic Over Current Limiter

This IC has a built-in dynamic over current limiter. In case that coil current exceeds I_{DPEAK} (1.60 A Typ) two times consecutively, it stops pulse operation for t_{DPEAK} (128 µs Typ).

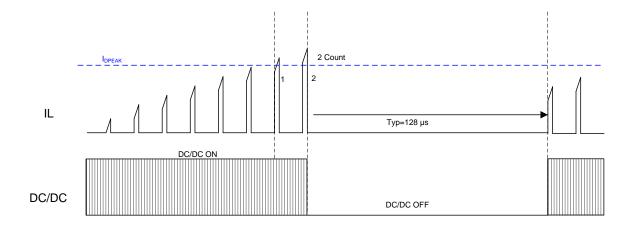


Figure 12. Dynamic Over Current Limiter

6 DC/DC Driver - continued

6.7 Soft Start Operation

At starting up, this function controls the over current limiter value in order to prevent any excessive voltage or current rising. The details are shown in Figure 13. The IC enables the soft start operation by changing the over current limiter value with time.

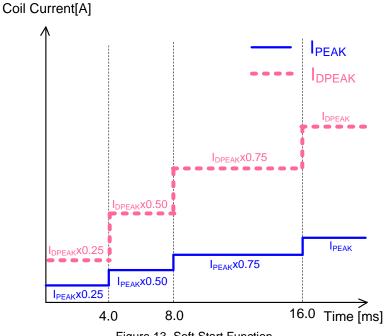


Figure 13. Soft Start Function

7 Output Over Load Protection Function (OLP comparator)

Output over load protection function monitors load status and stops switching at over load. In the over load condition, the output voltage lowers. If a state is electric power set in the IC or more continues for tFOLP1 (128 ms Typ), the IC stops switching by judging the status as over load. The recovery after detection of OLP is tFOLP2 (512 ms Typ) later.

Temperature Protection Circuit

Temperature protection circuit stops the oscillation of DC/DC if the IC becomes more than a certain temperature.

9 Operation Mode of Protection Circuits
The operation mode of protection functions is shown in Table 1.

Table 1. The operation mode of protection functions

Function	VCC Pin Under Voltage Protection	VCC Pin Over Voltage Protection	Over Temperature Protection	Over Power Protection
Detection	V _{UVLO2} (at falling voltage)	V _{OVP1} (at rising voltage)	150 °C (at rising temperature)	the current detected by over current detection or more
Release	V _{UVLO1} (at rising voltage)	V _{OVP2} (at falling voltage)	85 °C (at falling temperature)	under over current detection
Detection Timer	-	100 μs	100 μs	128 ms
Release Timer	-	-	-	512 ms
Туре	Auto Recovery	Auto Recovery	Auto Recovery	Auto Recovery
Timer Reset Condition 1	-	VCC UVLO Detection	VCC UVLO Detection	VCC UVLO Detection
Timer Reset Condition 2	-	<detection> Release Condition <release> Detection Condition</release></detection>	<detection> Release Condition <release> Detection Condition</release></detection>	<detection> Release Condition <release> Detection Condition</release></detection>

10 External Components

Each part should be designed considering the input voltage condition and output load condition. Figure 14 shows application circuit.

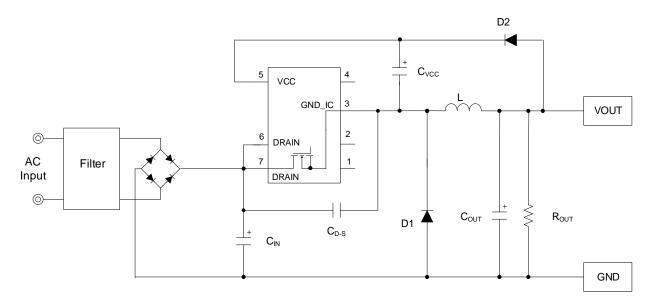


Figure 14. Application circuit

10.1 Output Capacitor Cout

Output capacitor C_{OUT} should be designed considering the spec of output ripple voltage and to startup until t_{FOLP1} (128 ms Typ). It is recommended to be 100 μ F or more.

10.2 Inductor L

The value of inductor should be designed considering the spec of output load condition and the input voltage range. If inductor value is too large, dc/dc operation becomes continuous mode and increases heat. If inductor value is too small, it is impossible that the IC controls in the Minimum ON width t_{MINON} or less, so there is a possibility of over current detection at normal operation load. It is recommended to be 270 μ H to 680 μ H.

10.3 VCC Pin Capacitor Cvcc

The VCC pin Capacitor C_{VCC} adjusts startup time and response of Error AMP. It is recommended to design less than 1/100 value of C_{OUT}.

10.4 Capacitor between the DRAIN Pin and the GND_IC Pin C_{D-S}

It is recommended to design the capacitor between the DRAIN pin and the GND_IC pin C_{D-S} less than 22 pF.

Absolute Maximum Ratings (Ta=25 °C)

<u> </u>				
Parameter	Symbol	Rating	Unit	Conditions
Maximum Applied Voltage 1	\/	-0.3 to +650	V	DRAIN(BM2P104Q-Z)
Maximum Applied Voltage 1	V _{MAX1}	730	V	DRAIN(tpulse < 10 µs) (Note 1)
Maximum Applied Voltage 1	V _{MAX1}	-0.3 to +800	V	DRAIN(BM2P107QK-Z)
Maximum Applied Voltage 2	V _{MAX2}	-0.3 to +32.0	V	VCC
DRAIN Current Pulse	I _{DD}	4.00	А	Consecutive operation (BM2P104Q-Z)
DRAIN Current Pulse	I _{DD}	2.00	А	Consecutive operation (BM2P107QK-Z)
Power Dissipation	Pd	1.00	W	(Note 2)
Maximum Junction Temperature	Tjmax	+150	°C	
Storage Temperature Range	Tstg	-55 to +150	°C	

- Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.
- Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with power dissipation taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

(Note 1) Duty is less than 1 %.

(Note 2) Derate by 4.563 mW/°C when operating Ta=25 °C or more when mounted (on 70 mm x 70 mm x 1.6 mm thick, glass epoxy on single-layer substrate).

Thermal Loss

The thermal design should set operation for the following conditions.

- 1. The ambient temperature Ta must be 105 $^{\circ}\text{C}$ or less.
- 2. The IC's loss must be within the Power Dissipation Pd.

The thermal abatement characteristics are as follows.

(PCB: 70 mm x 70 mm x 1.6 mm single layer board, the back side is copper foil)

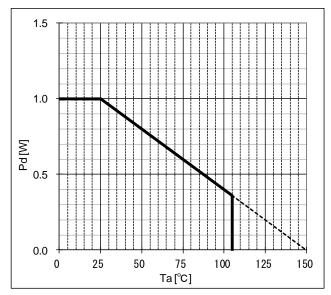


Figure 15. Thermal Abatement Characteristics

Recommended Operating Conditions

Parameter	Symbol	Specifications			Unit	Conditions	
i arameter	Cymbol	Min	Тур	Max	Offic	Conditions	
Dawer Cumby Voltage Dange 4	VDRAIN		-	650	V	DRAIN(BM2P104Q-Z)	
Power Supply Voltage Range 1	V DRAIN		-	730	V	DRAIN(tpulse < 10 µs) (Note 3)	
Power Supply Voltage Range 1	VDRAIN		-	800	V	DRAIN(BM2P107QK-Z)	
Power Supply Voltage Range 2	Vcc	8.00	-	10.81	V	VCC	
Operating Temperature	Topr	-40	-	+105	°C	Surrounding temperature	
(Note 3) Duty is less than 1 %							

Electrical Characteristics in MOSFET Part (Unless otherwise noted, Ta=25 °C)

	1			/		
Parameter	Symbol	Specifications			Unit	Conditions
1 didiffeter	Cymbol	Min	Тур	Max	Offic	Conditions
		650			٧	I _D =1 mA / V _{GS} =0 V
Voltage between DRAIN	V	650	-	-	V	(BM2P104Q-Z)
and SOURCE	$V_{(BR)DDS}$	700			\/	ID = 1 mA, VGS = 0 V
		730	-	-	V	tpulse < 10 μs
DDAIN Look Compant	IDSS	-	0	100	μA	V _{DS} =650 V / V _{GS} =0 V
DRAIN Leak Current						(BM2P104Q-Z)
ON Resistor	R _{DS(ON)}	_	4.0	4.5	Ω	I _D =0.25 A / V _{GS} =10 V
ON NESISION	NDS(ON)	_	4.0	4.5	32	(BM2P104Q-Z)
Voltage between DRAIN	V _{(BR)DDS}	800		_	V	I _D =1 mA / V _{GS} =0 V
and SOURCE	V (BR)DDS	800	•	-	v	(BM2P107QK-Z)
DDAIN Look Current	1		0	100		V _{DS} =800 V / V _{GS} =0 V
DRAIN Leak Current	IDSS	-	0	100	μA	(BM2P107QK-Z)
ON Resistor	R _{DS(ON)}	_	7.5	10.5	Ω	I _D =0.25 A / V _{GS} =10 V
014 1(03)3(0)	INDS(ON)		7.5	10.5	3.2	(BM2P107QK)

Electrical Characteristics in Start Circuits Part (Unless otherwise noted, Ta=25 °C)

Parameter	Svmbol	5	Specification	ns	Unit	Conditions	
1 diameter	Cymbol	Min	Тур	Max	Offic	Conditions	
Start Current 1	I _{START1}	0.150	0.300	0.600	mΑ	VCC=0 V	
Start Current 2	ISTART2	1.200	3.000	6.000	mΑ	VCC=7 V	
OFF Current	ISTART3	-	10	20	μA	After UVLO is released	
Start Current Switching Voltage	V_{SC}	0.500	0.800	1.200	V		

Electrical Characteristics in Control IC Part (Unless otherwise noted, Ta=25 °C)

Parameter	Symbol	mbol			Unit	Conditions
	Cymbol	Min	Тур	Max	0	Conditions
[Circuit Current]			1	1		
Circuit Current (ON) 1	I _{ON1}	-	1200	1650	μA	At pulse operation Drain=open
Circuit Current (ON) 2	I _{ON2}	300	450	550	μΑ	At burst operation
[VCC Pin Protection Function]						
VCC UVLO Voltage 1	V _{UVLO1}	8.10	8.80	9.50	V	VCC rising
VCC UVLO Voltage 2	V _{UVLO2}	6.60	7.30	8.00	V	VCC dropping
VCC UVLO Hysteresis	V _{UVLO3}	-	1.50	-	V	Vuvlo3=Vuvlo1-Vuvlo2
VCC Recharge Start Voltage	V _{CHG1}	7.00	7.70	8.40	V	
VCC Recharge Stop Voltage	V _{CHG2}	7.40	8.10	8.80	V	
VCC Recharge Hysteresis	V _{CHG3}	0.20	0.40	0.70	V	
VCC Control Voltage	V _{CNT}	9.90	10.00	10.10	V	
VCC OVP Voltage 1	V _{OVP1}	10.81	11.50	12.19	V	VCC sweep up
VCC OVP Voltage 2	V _{OVP2}	-	11.00	-	V	VCC sweep down
VCC OVP Hysteresis	V _{OVP3}	0.21	-	0.63	V	- Professional Control of the Contro
VCC OVP Timer	tcomp	50	100	150	μs	
						Control IC part
Over Temperature Protection 1	T _{SD1}	120	150	180	°C	At temperature rising ^(Note 4)
Over Temperature Protection 2	T _{SD2}	-	85	-	°C	Control IC part At temperature dropping (Note 4)
Over Temperature Protection Hysteresis	T _{SD3}	-	65	-	°C	(Note 4)
[PWM Type DC/DC Driver Block]					1	
Oscillation Frequency	fsw	94	100	106	kHz	
Frequency Hopping Width	f _{DEL}	-	6.0	-	kHz	
Maximum Duty	DMAX	66	75	84	%	
FB OLP ON Detection Timer	t _{FOLP1}	80	128	176	ms	
FB OLP OFF Detection Timer	tFOLP1	332	512	692	ms	
Soft Start Time 1		2.8	4.0	5.2	ms	
Soft Start Time 2	tss1	5.6	8.0	10.4		
Soft Start Time 3	t _{SS2}	11.2	16.0	20.8	ms ms	
[Over Current Detection Block]	tss3	11.2	10.0	20.0	1115	<u> </u>
Over Current Detection	IPEAK	0.720	0.800	0.880	Α	
Over Current Detection in SS1	IPEAK1	-	0.200	-	Α	(Note 4)
Over Current Detection in SS2	I _{PEAK2}	-	0.400	-	Α	(Note 4)
Over Current Detection in SS3	I _{PEAK3}	-	0.600	-	Α	(Note 4)
Dynamic Over Current Detection	IDPEAK	1.440	1.600	1.740	Α	
Dynamic Over Current Detection in SS1	I _{DPEAK1}	-	0.400	-	Α	(Note 4)
Dynamic Over Current Detection in SS2	I _{DPEAK2}	-	0.800	-	Α	(Note 4)
Dynamic Over Current Detection in SS3	Ідреакз	-	1.200	-	А	(Note 4)
Dynamic Over Current Enforced OFF Time	t _{DPEAK}	64	128	170	μs	
Leading Edge Blanking Time	t _{LEB}	-	150	-	ns	(Note 4)
Minimum ON Width	t _{MINON}	-	300	550	ns	(Note 4)

(Note 4) Not 100% tested.

Application Examples

Show a flyback circuitry example in Figure 15. Be careful with the DRAIN voltage because high voltage is produced by ringing in turn OFF. With this IC, It become able to work to 730V.

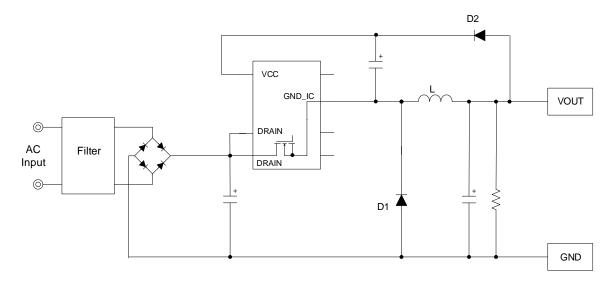


Figure 16. Flyback Application Ciucit

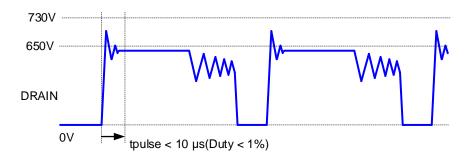


Figure 17. Drain Pin Ringing Waveform

I/O Equivalence Circuit

7	DRAIN	6	DRAIN	-	_	5	VCC	
 - E -	DRAIN Internal MOSFET GND_IC	 	DRAIN Internal MOSFET GND_IC		-	VCC		
1	N.C.	2	N.C.	3	GND_IC	4	N.C.	
No	on Connection			3 GND_IC GND_IC		4 N.C. Non Connection		

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

Operational Notes - continued

10. Regarding the Input Pin of the IC

This IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

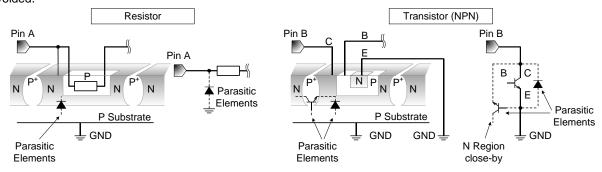


Figure 18. Example of IC Structure

11. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

12. Thermal Shutdown Circuit (TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's maximum junction temperature rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF power output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

13. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

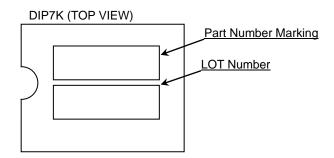
Ordering Information



Lineup

Orderable Part Number	I _{DD} (A)	V _{DRAIN} (Max) (V)	R _{DS(ON)} (Typ) (Ω)	Package	Part Number Marking
BM2P104Q-Z	4.00	730	4.0	DIP7K	BM2P104Q
BM2P107QK-Z	2.00	800	7.5	DIP/K	BM2P107QK

Making Diagram

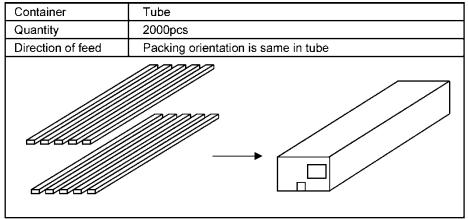


Physical Dimension and Packing Information Package Name DIP7K က 35 ± 0 . 9. 9. $27^{+0.89}_{-0.25}$ 7.62 . 0 ro . က MAX . 9 \mathfrak{S} ಣ MIN0. ი ი 0° \sim 15° 0.46 2.549. 02 ± 0 . 51 1.5°

> (UNIT:mm) PKG:DIP7K

 ${\tt Drawing} \ {\tt No.} \ {\tt EX001-0076}$

< Container Information >



Revision History

Date	Revision	Changes
16.Jul.2019	001	New release
22.Dec.2020	P15 Change the Absolute Maximum Ratings	

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JAPAN	USA	EU	CHINA	
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CLASSIV	CLASSIII	CLASSⅢ	CLASSIII	

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 - [g] Use of our Products without cleaning residue of flux (Exclude cases where no-clean type fluxes is used. However, recommend sufficiently about the residue.); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
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