



BGA3131

DOCSIS 3.1 upstream amplifier

Rev. 3.1 — 30 April 2021

Product data sheet

1 General description

The BGA3131 is an upstream amplifier meeting the Data Over Cable Service Interface Specifications (DOCSIS 3.1). It is designed for cable modem, CATV set top box and VoIP modem applications. The device operates from 5 MHz to 205 MHz. The BGA3131 provides 58 dB gain control range in 1 dB increments with high incremental accuracy. Its maximum gain setting delivers 37 dB voltage gain and a superior linear performance.

It supports the DOCSIS 3.1 output power levels while meeting the stringent ACLR requirements.

The BGA3131 operates at 5 V supply. The gain is controlled via a 3-wire serial interface. The current consumption can be reduced in 4 steps via the serial interface. This interface enables the user to optimize between DC power efficiency and linearity. In addition, the current is automatically reduced at lower gain settings while preserving the linearity performance. In disable mode, the device draws typical 25 mA while it can be still programmed to new gain and current settings.

The BGA3131 is housed in 20 pins 5 mm x 5 mm leadless HVQFN package.

2 Features and benefits

- 58 dB gain control range in 1 dB steps using a 3-wire serial interface
- 5 MHz to 205 MHz frequency operating range
- ± 0.4 dB incremental gain step accuracy
- Maximum voltage gain 37 dB
- Excellent IMD3 of -60 dBc at 68 dBmV total output power
- Excellent second harmonic level of -65 dBc at 68 dBmV total output power
- Excellent third harmonic level of -65 dBc at 68 dBmV total output power
- Excellent noise figure of 6.5 dB at maximum gain
- Capable of transmitting modulated carriers while meeting the DOCSIS 3.1 ACLR specification. At an output power of 65 dBmV at the F-connector (assuming 3 dB of output loss), the typical ACLR is -62 dBc
- 5 V single supply operation
- Excellent ESD protection at all pins
- Unconditionally stable
- Compliant to Directive 2002/95/EC, regarding Restriction of Hazardous Substances (RoHS)



3 Applications

- DOCSIS 3.1 and 3.0 cable modems
- VoIP modems
- Set-top boxes

4 Quick reference data

Table 1. Quick reference data

Typical values at $V_{CC} = 5\text{ V}$, decimal current setting = 3, decimal gain setting 50 to 63; $T_{amb} = 25\text{ °C}$; $Z_{i(dif)} = 200\ \Omega$; $Z_{o(se)} = 75\ \Omega$; voltage gain does include loss due to output transformer. Unless otherwise specified. All RF parameters are measured on an application board with the circuit as shown in [Figure 12](#) and components listed in [Table 17](#).

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
I_{CC}	supply current	transmit-enable mode; TX_EN = HIGH	610	660	720	mA
		transmit-disable mode; TX_EN = LOW	-	25	-	mA
G_v	voltage gain	gain code = 111111 [1] [2]	-	37	-	dB
NF	noise figure	transmit-enable mode; gain code = 111111	-	6.5	-	dB
α_{2H}	second harmonic level	transmit-enable mode; gain code = 111111; $P_i(\text{RMS}) = 31\text{ dBmV}$; $P_L(\text{RMS}) = 68\text{ dBmV}$ into $75\ \Omega$ impedance	-	-65	-	dBc
α_{3H}	third harmonic level	transmit-enable mode; gain code = 111111; $P_i(\text{RMS}) = 31\text{ dBmV}$; $P_L(\text{RMS}) = 68\text{ dBmV}$ into $75\ \Omega$ impedance	-	-65	-	dBc
IMD3	third-order intermodulation distortion	transmit-enable mode; gain code = 111111; $P_L(\text{RMS}) = 65\text{ dBmV}$ per tone into $75\ \Omega$ impedance	-	-60	-	dBc
$P_{L(1dB)}$	output power at 1 dB gain compression	CW input signal RMS value, $f = 205\text{ MHz}$	-	78	-	dBmV

[1] $P_i \leq 31\text{ dBmV}$

[2] Excluding 5.7 dB loss of resistive matching circuit, to match $75\ \Omega$ to $50\ \Omega$

Table 2. ACLR quick reference data

Typical values at $V_{CC} = 5\text{ V}$, decimal current setting = 3, decimal gain setting 60, $T_{amb} = 25\text{ °C}$; $Z_{i(dif)} = 200\ \Omega$; $Z_{o(se)} = 75\ \Omega$; channel bandwidth = 192 MHz, integration bandwidth = 9.6 MHz, $f = 5\text{ MHz to }205\text{ MHz}$. Unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
DOCSIS 3.1						
ACLR	adjacent channel leakage ratio	$P_i(\text{RMS}) = 31\text{ dBmV}$; $P_L(\text{RMS}) = 68\text{ dBmV}$, channel configuration: channel bandwidth is 192 MHz, with exclusion band at 147.5 MHz, with a bandwidth of 9.6 MHz. Input signal with a PAPR of 13 dB	-	-62	-	dBc

5 Ordering information

Table 3. Ordering information

Type number	Orderable part number	Package		
		Name	Description	Version
BGA3131	BGA3131J	HVQFN20	plastic thermal enhanced very thin quad flat package; no leads; 20 terminals; body 5 mm x 5 mm x 0.85 mm	SOT662-1

6 Marking code

Table 4. Marking

Type number	Marking code
BGA3131	3131

7 Functional diagram

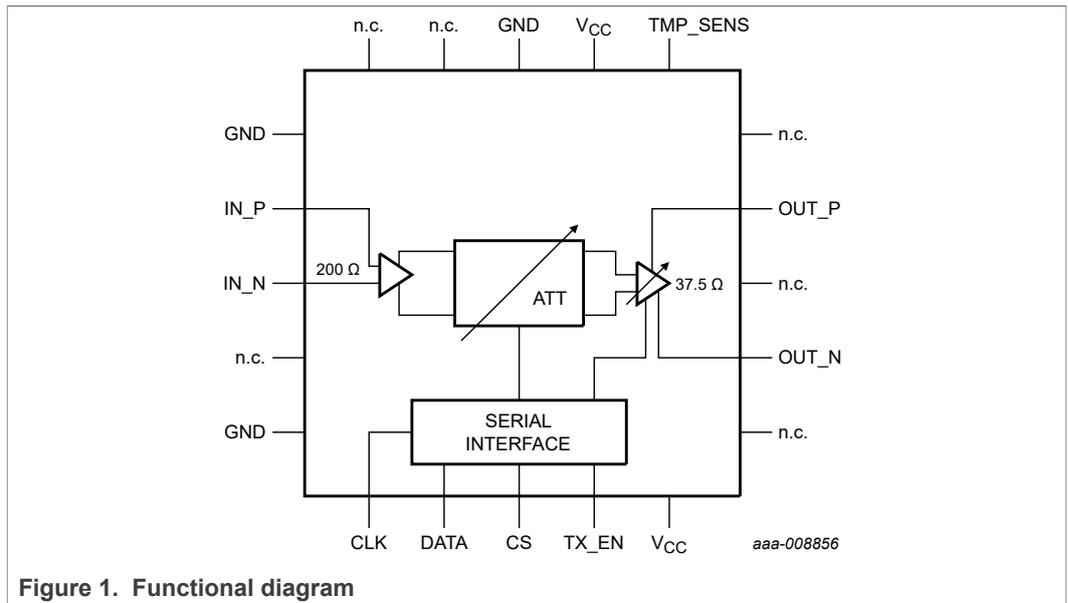
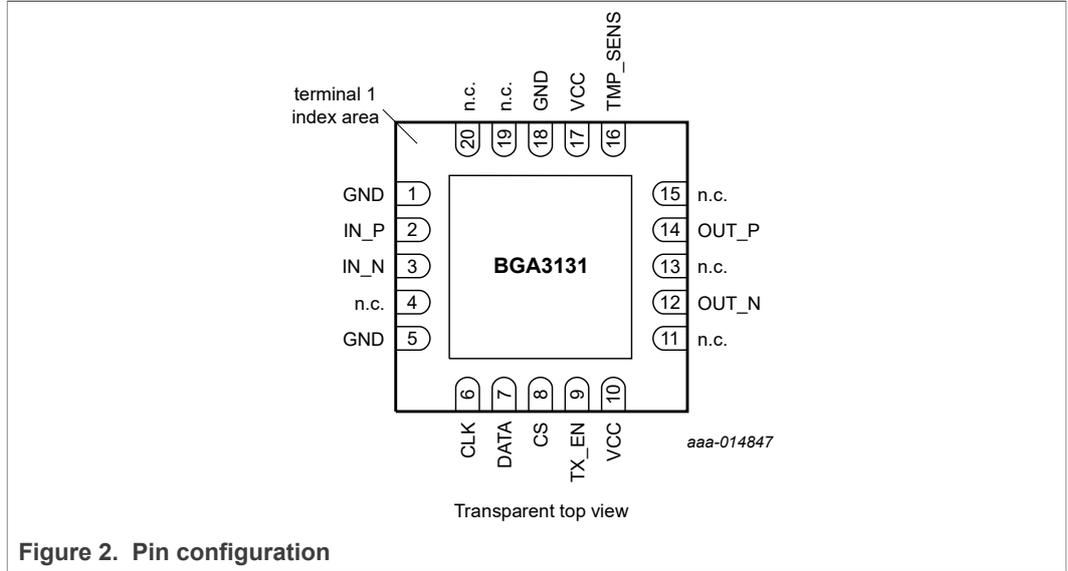


Figure 1. Functional diagram

8 Pinning information

8.1 Pinning



8.2 Pin description

Table 5. Pin description

Symbol	Pin	Description
GND	1	ground
IN_P	2	amplifier input +
IN_N	3	amplifier input –
n.c.	[1] 4	not connected
GND	5	ground
CLK	6	clock
DATA	7	data
CS	8	chip select
TX_EN	9	transmit enable active HIGH
V _{CC}	10	supply voltage
n.c.	[1] 11	not connected
OUT_N	12	amplifier output –
n.c.	[1] 13	not connected, pin can be left open, grounded, or connected to the center tap voltage in the application
OUT_P	14	amplifier output +
n.c.	[1] 15	not connected
TMP_SENS	16	temperature sense

Table 5. Pin description...continued

Symbol	Pin	Description
V _{CC}	17	supply voltage
GND	18	ground
n.c.	^[1] 19	not connected
n.c.	^[1] 20	not connected
GND	die paddle	ground

[1] not connected pins can either be left open or grounded in the application

9 Functional description

9.1 Logic programming

The programming word is set through a shift register. It uses the data of the SPI bus (pin name DATA), clock (pin name CLK), and enable (pin name TX_EN) lines. By default, the data is entered in order with the most significant bit (MSB) first and the least significant bit (LSB) last. The Chip Select line (CS) must be low during the data entry, then set high to sample the shift register. The rising edge of the clock pulse shifts each data value into the shift register. When the register is programmed, the new settings take effect:

- on the rising edge, of <CS> if <TX_EN> was HIGH
- on the rising edge, of <TX_EN> if <TX_EN> was LOW

Table 6. Programming register

Data bit	11	10	9	8	7	6	5	4	3	2	1	0
Function	Register address				Current setting ^[1]		Attenuation (gain) setting ^[2]					
Settings	0	0	0	0	C[1]	C[0]	G[5]	G[4]	G[3]	G[2]	G[1]	G[0]
Initialize	0	0	0	1	Soft reset (mirror)	LSB first (mirror)	ASC address (mirror)	16b mode (mirror)	16b mode	ASC address	LSB first	Soft reset
Reserved	0	0	1	0	0	0	0	0	0	0	0	0
Reserved	0	0	1	1	0	0	0	0	0	0	0	0

[1] For current bit settings see [Table 8](#)

[2] For gain bit settings see [Table 7](#)

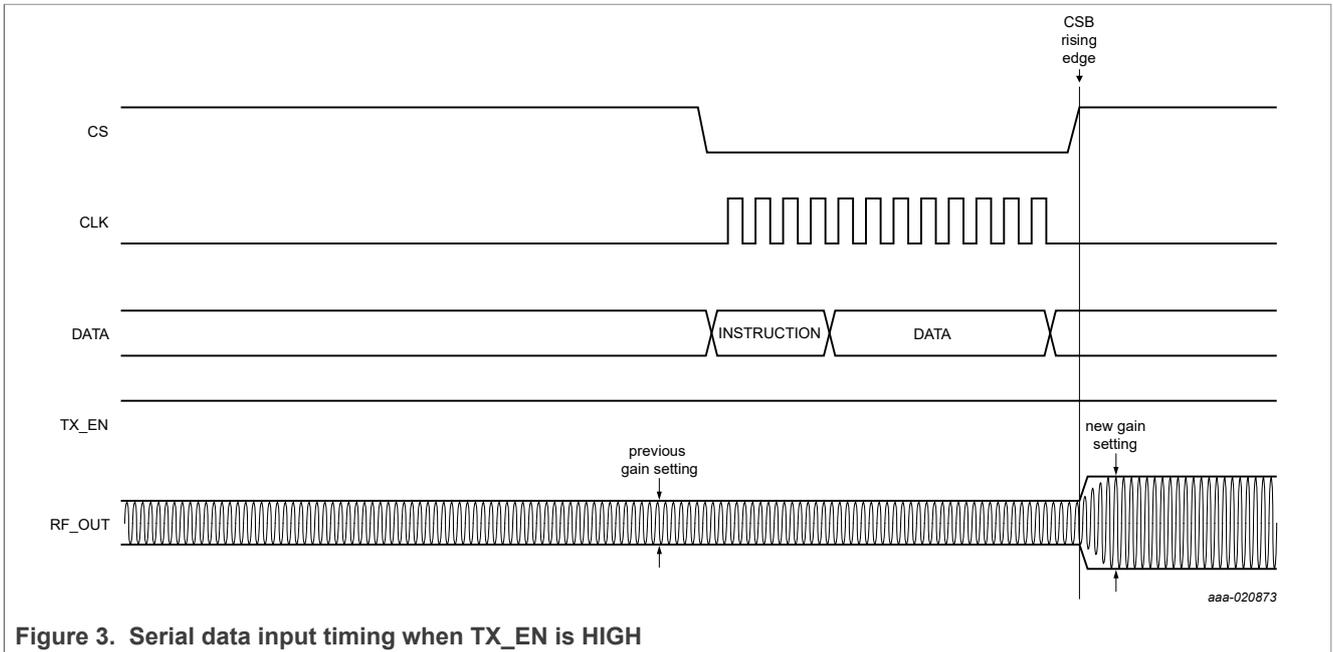


Figure 3. Serial data input timing when TX_EN is HIGH

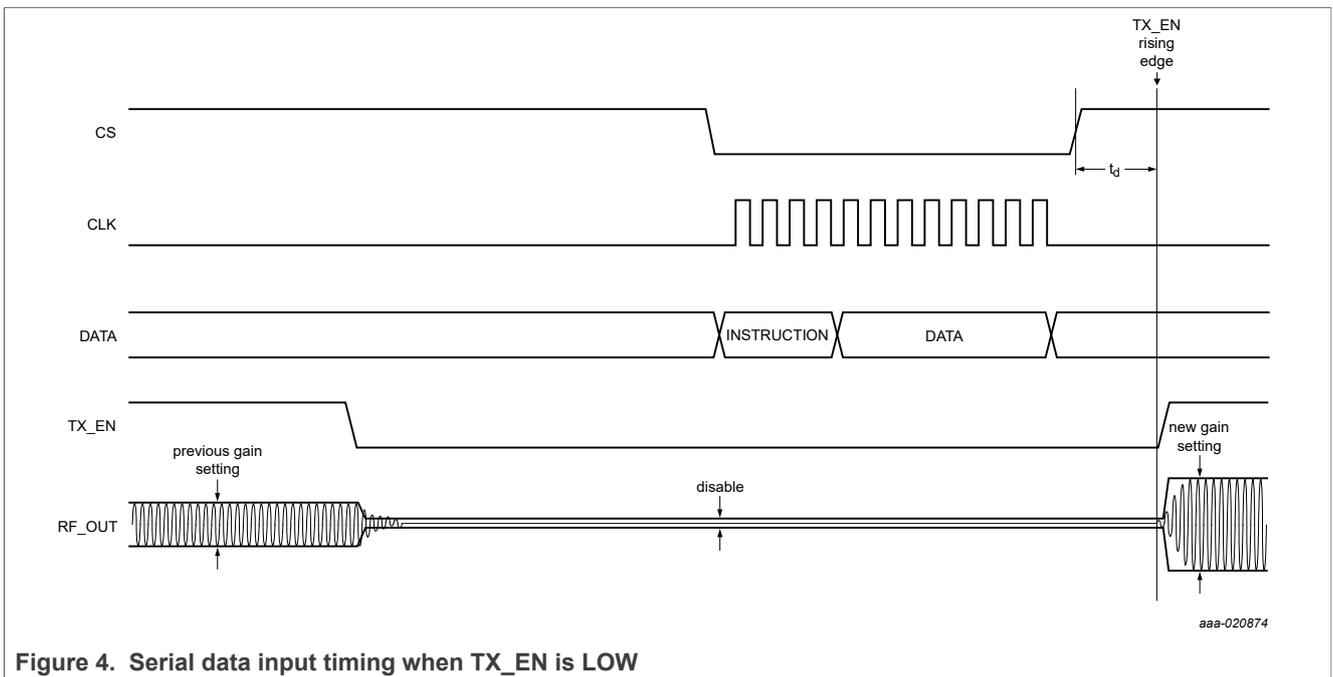


Figure 4. Serial data input timing when TX_EN is LOW

9.2 Register settings

9.2.1 Register address

Only addresses 0000 to 0011 are used. Other addresses do not affect the VGA.
 Address 0000 is used to configure attenuation and current parameters of the device.
 Address 0001 is used to configure the SPI interface specifically.
 Addresses 0010 and 0011 are reserved, and must be kept at value 0.

9.2.2 Gain/attenuator setting

The gain shall be controlled via the SPI bus. Data bits D0 through D5 set the gain/attenuator level, with 111111 being the min attenuation setting, and 000101 being the maximum attenuation setting. A new gain/attenuator setting can be loaded while the VGA is on (transmit-enable).

Table 7. Gain settings

Gain setting G[5:0] ^[1]			Typical gain
binary notation	decimal notation		(dB)
000000 to 000101	0 to 5		-21
000110	6		-20
111110	62		36
111111	63		37

[1] With every increment of the gain setting between 000101 (5) and 111111 (63), the typical gain increases accordingly

9.2.3 Output stage current setting

The current (of the output stage) shall be controlled via the 3-wire bus. Data bits D6 and D7 set the current. Setting 11 sets the maximum current for maximum linearity. The current can be lowered for improved efficiency at lower output power levels, or lower linearity requirements. Setting 00 sets the minimum current. A new current setting can be loaded while the VGA is on (transmit-enable).

Table 8. Supply current settings

At decimal gain setting 63.

Current setting C[1:0]		Typical supply current
binary notation	decimal notation	(mA)
00	0	350
01	1	410
10	2	480
11	3	660

The current is automatically reduced at lower gain settings while preserving the linearity performance.

Table 9. Device current settings versus gain settings

Gain setting		Typical current (mA)				Comments
Attenuation bit [5.0]	H value	Current setting C[1:0] = 00	Current setting C[1:0] = 01	Current setting C[1:0] = 10	Current setting C[1:0] = 11	
111111	0x3F	350	410	480	660	Max gain (code = 63)
110001	0x31	280	315	345	370	Gain code = 49
101011	0x2B	290	320	350	375	Gain code = 43
100101	0x25	240	260	260	330	Gain code = 37
011001	0x19	220	235	235	250	Gain code = 25

9.2.4 SPI Initialize register

The SPI receiver may be configured in several communication modes. By default, the device is waiting for a 12-bit, MSB first SPI frame. In that case, the address field is 4 bits wide, and data field is 8 bits wide. Using the Initialize register at address 0x01 allows switching the device to different SPI modes. Register 0x01 contains four effective bits, but programmed with the mirror value of the 4 LSBs in the 4 MSBs.

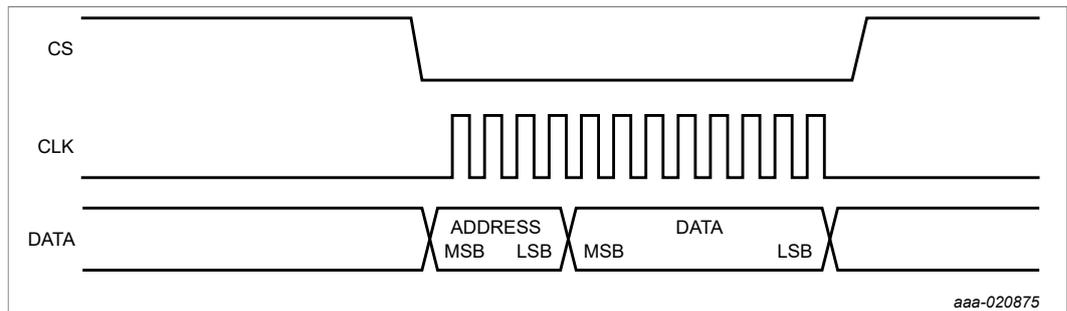


Figure 5. Default SPI frame, 12-bit, MSB first, descending address

9.2.4.1 SPI Soft Reset

By setting bits *Soft_reset* AND *Soft_reset (mirror)* at address 0x01, the device is put back in its default state. (maximum gain)

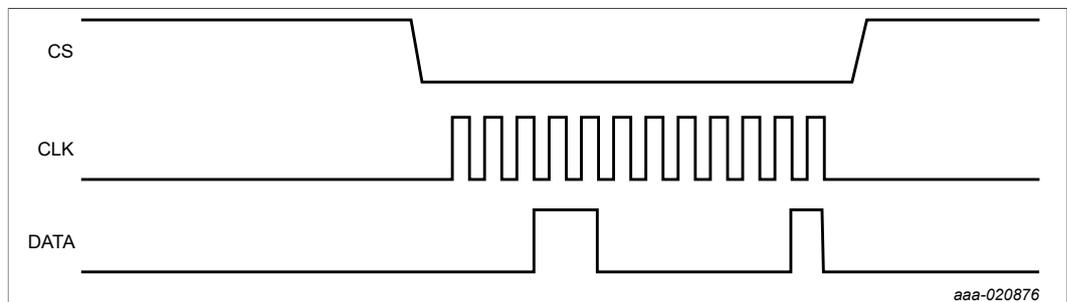


Figure 6. SPI frame for Soft Reset, 12-bit, MSB first, descending address

9.2.4.2 SPI 16-bit mode

By default, the SPI frame is made of 4 bits for address and a multiple of 8 bits for data. By setting bits *16b_mode* AND *16b_mode (mirror)* at address 0x01, the device is configured such that the next SPI command will be a 16-bit command. Address is sent on 8 bits, whereas data is a multiple of 8 bits.

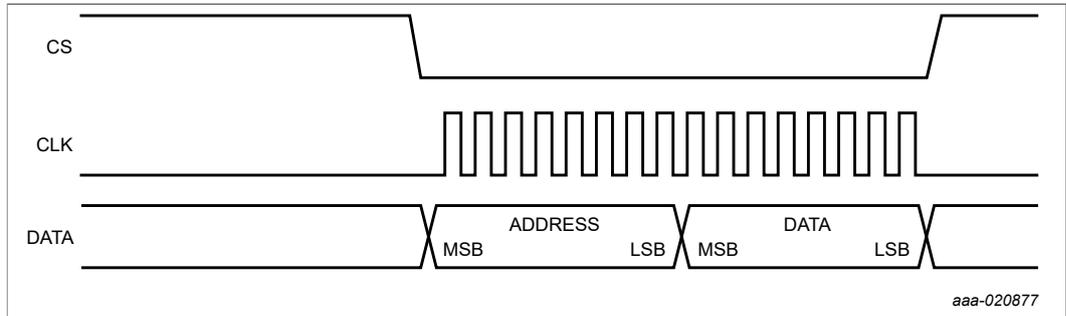


Figure 7. Default SPI frame, 16-bit, MSB first, descending address

9.2.4.3 SPI ascending address

By default, the SPI slave can be programmed with a single SPI frame. The SPI contains a start address and several data bytes to be written at start address. The SPI has an auto decrementing mechanism to store each data in corresponding register. By setting bits *asc_addr* AND *asc_addr (mirror)* at address 0x01, the device is configured such that the internal addresses are auto-increment instead of auto-decrement.

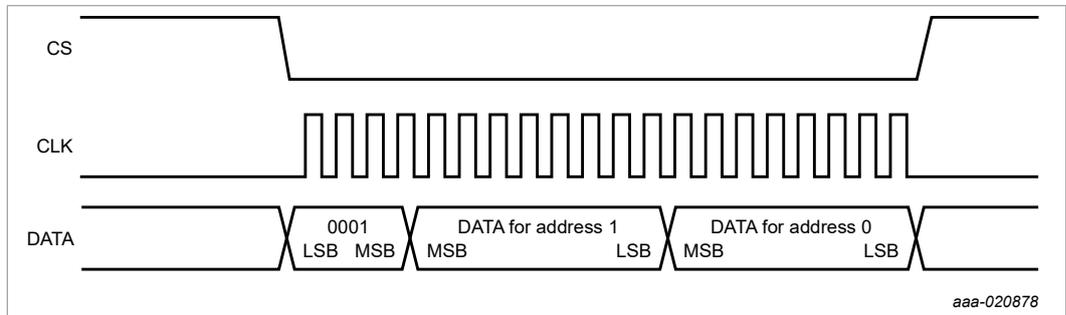


Figure 8. Burst SPI frame, 12-bit, MSB first, descending address

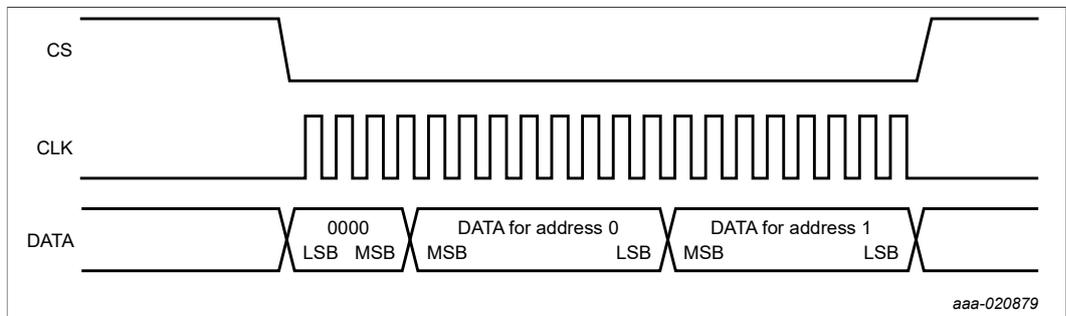


Figure 9. Burst SPI frame, 12-bit, MSB first, ascending address

9.2.4.4 SPI LSB first

By default, the SPI slave waits for the MSB data first. By setting bits *lsb_first* AND *lsb_first (mirror)* at address 0x01, the device is configured. The first bit received is considered as the LSB of each field (address and data).

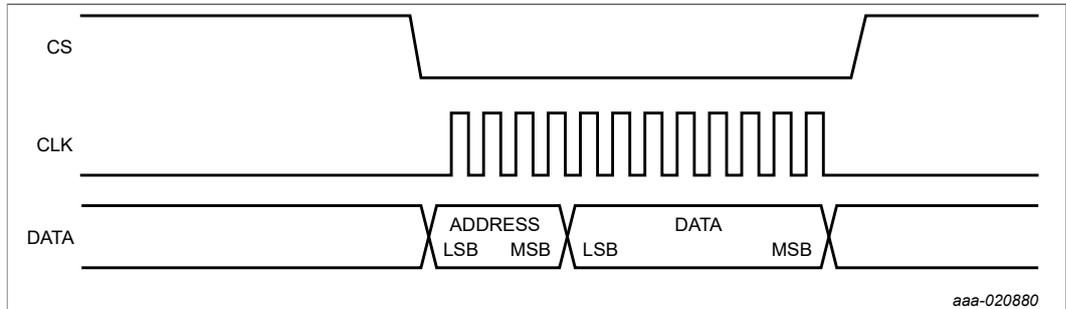


Figure 10. Burst SPI frame, 12-bit, MSB first, descending address

9.3 TX enable / TX disable

The amplifier can be disabled or enabled by making TX_EN (pin 9) LOW or HIGH. A LOW to HIGH TX enable transition enables new programmed settings. If no new settings are programmed, the last programmed setting is reactivated.

10 Limiting values

Table 10. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Absolute Maximum Ratings are given as Limiting Values of stress conditions during operation, that must not be exceeded under the worst probable conditions.

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CC}	supply voltage		-	6	V
I_i	input current	on pin TMP_SENS	-	1	mA
V_i	input voltage	on pin IN_P	-0.5	6	V
		on pin IN_N	-0.5	6	V
		on pin CLK ^[1]	-0.5	6	V
		on pin DATA ^[1]	-0.5	6	V
		on pin CS ^[1]	-0.5	6	V
		on pin TX_EN ^[1]	-0.5	6	V
		on pin OUT_N	-0.5	6	V
		on pin OUT_P	-0.5	6	V
$P_{I(max)}$	maximum input power		-	60	dBmV
T_{stg}	storage temperature		-55	150	°C
T_j	junction temperature		-	150	°C
f_{SPI}	SPI frequency	Master writes to slave; load on DATA line, 30 pF maximum, under nominal V_{IL} and V_{IH} levels	-	25	MHz
V_{ESD}	electrostatic discharge voltage	Human Body Model (HBM); According to JEDEC standard 22-A114E ^[2]	-	+/-4	kV
		Charged Device Model (CDM); According to JEDEC standard 22-C101B ^[2]	-	+/-2	kV

[1] All digital pins may not exceed V_{CC} as the internal ESD circuit can be damaged. To prevent this damage, it is recommended that control pins are limited to a maximum of 5 mA.

[2] Stressed with pulses of 200 ms in duration.

11 Thermal characteristics

Table 11. Thermal characteristics

Symbol	Parameter	Conditions	Typ	Unit
$R_{th(j-bot)}$	thermal resistance from junction to bottom of package	Still air, natural convection ^[1]	6.1	K/W
$R_{th(j-a)}$	thermal resistance from junction to ambient	Still air, natural convection ^[1]	29.3	K/W
$\Psi_{(j-top)}$	thermal characterization parameter from junction to top of package	Still air, natural convection ^[1]	9.9	K/W

[1] Simulated using final element method model resembling the device mounted on the application board. See [Figure 13](#)

Note:

For more thermal details, refer to the BGA3131 Thermal management guidelines AN11753 at www.nxp.com.

11.1 Thermal compact model

Compact thermal model parameters (Delphi compact model), definitions according to [Figure 11](#)

Table 12. Delphi model parameters ^[1]

R _{th} (K/W)	junction	top inner	top outer	bottom inner	bottom outer	sides	leads	surface areas [mm ²]
junction		201		6.17				
top inner			1207	543	1330			4.27
top outer				114	60.4		222	20.7
bottom inner					53.8	315	311	9.53
bottom outer							37.8	12.6
sides							101	17
leads								2.85

[1] Cells are intentionally left empty

Table 13. Delphi compact model definition

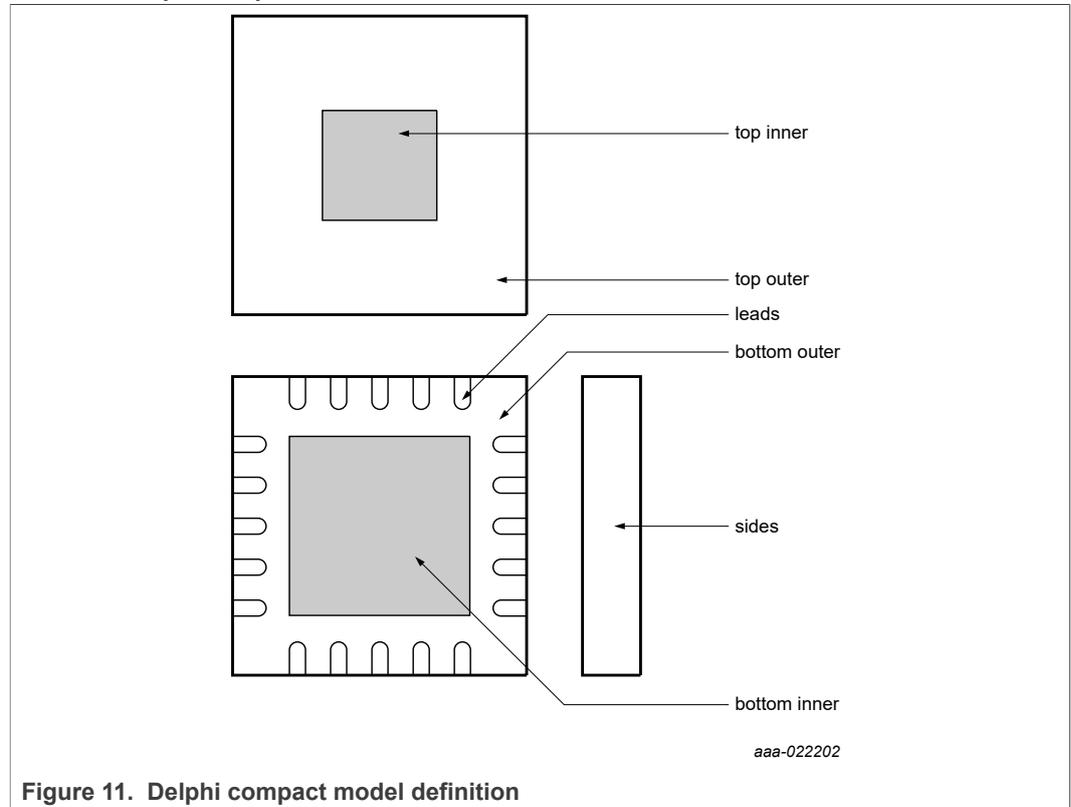


Figure 11. Delphi compact model definition

12 Static characteristics

Table 14. Characteristics

Typical values at $V_{CC} = 5\text{ V}$, decimal current setting = 3, decimal gain settings 50 to 63, $T_{amb} = 25\text{ °C}$; $Z_{i(dif)} = 200\ \Omega$; $Z_{o(se)} = 75\ \Omega$. Unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{CC}	supply voltage		4.75	5.0	5.25	V
I_{CC}	supply current	transmit-enable mode; TX_EN = HIGH	610	660	720	mA
		transmit-disable mode; TX_EN = LOW	-	25	-	mA
V_{IH}	HIGH-level input voltage	[1]	1.8	-	$V_{CC} + 0.6$	V
V_{IL}	LOW-level input voltage	[1]	0	-	0.8	V
P	power dissipation		-	3.3		W

[1] Voltage on the control pins.

13 Dynamic characteristics

Table 15. Characteristics

Typical values at $V_{CC} = 5\text{ V}$, decimal current setting = 3, decimal gain setting 50 to 63; $T_{amb} = 25\text{ °C}$; $Z_{i(dif)} = 200\ \Omega$; $Z_{o(se)} = 75\ \Omega$; voltage gain does include loss due to output transformer. Unless otherwise specified. All RF parameters are measured on an application board with the circuit as shown in [Figure 12](#) and components listed in [Table 17](#).

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
G_v	voltage gain	gain code = 111111 [1] [2]	-	37	-	dB
		gain code = 001111 [1] [2]	-	-11	-	dB
G_{flat}	gain flatness	f = 5 MHz to 205 MHz [1]	-	+/-0.5	-	dB
RL_o	output return loss	transmit-mode enabled overall gain settings, measured in 75 Ω system	-	14	-	dB
		transmit-mode disabled overall gain settings, measured in 75 Ω system	-	12	-	dB
RL_i	input return loss	transmit-mode enabled overall gain settings, measured in 200 Ω system	-	20	-	dB
		transmit-mode disabled overall gain settings, measured in 200 Ω system	-	20	-	dB
G_{step}	gain step	[1]	-	1.0	-	dB
$E_{G(dif)}$	differential gain error	[1]	-	+/-0.4	-	dB
$R_{i(dif)}$	differential input resistance		-	200	-	Ω
$R_{o(dif)}$	differential output resistance		-	37.5	-	Ω
f_{range}	frequency range		5	-	205	MHz
α_{isol}	isolation	transmit-disable mode; TX_EN = LOW; f = 205 MHz	-	60	-	dB
NF	noise figure	transmit-mode; gain code = 111111	-	6.5	-	dB
		transmit-mode; gain code = 100101	-	15	-	dB

Table 15. Characteristics...continued

Typical values at $V_{CC} = 5\text{ V}$, decimal current setting = 3, decimal gain setting 50 to 63; $T_{amb} = 25\text{ °C}$; $Z_{i(dif)} = 200\ \Omega$; $Z_{o(se)} = 75\ \Omega$; voltage gain does include loss due to output transformer. Unless otherwise specified. All RF parameters are measured on an application board with the circuit as shown in [Figure 12](#) and components listed in [Table 17](#).

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{sw(G)}$	gain switch time	transmit-disable/transmit-enable transient duration	-	3	-	μs
		transmit-enable/transmit-disable transient duration	-	0.5	-	μs
V_{trt}	transient voltage	transmit-disable/transmit-enable transient step size; peak value				
		$\geq 55\text{ dBmV}$ output power	[3] [4]	45	-	mV
		49 dBmV output power	[3] [4]	15	-	mV
		43 dBmV output power	[3] [4]	10	-	mV
		37 dBmV output power	[3] [4]	5	-	mV
		$\leq 34\text{ dBmV}$ output power	[3] [4]	3	-	mV
α_{2H}	second harmonic level	transmit-enable mode; gain code = 111111; $P_i(\text{RMS}) = 31\text{ dBmV}$; $P_L(\text{RMS}) = 68\text{ dBmV}$ into $75\ \Omega$ impedance	-	-65	-	dBc
α_{3H}	third harmonic level	transmit-enable mode; gain code = 111111; $P_i(\text{RMS}) = 31\text{ dBmV}$; $P_L(\text{RMS}) = 68\text{ dBmV}$ into $75\ \Omega$ impedance	-	-65	-	dBc
IMD3	third-order intermodulation distortion	transmit-enable mode; gain code = 111111; $P_L = 65\text{ dBmV}(\text{rms})$ per tone into $75\ \Omega$ impedance	-	-60	-	dBc
$P_{L(1dB)}$	output power at 1 dB gain compression	CW input signal RMS value, $f = 205\text{ MHz}$	-	78	-	dBmV (rms)

- [1] $P_i = 30\text{ dBmV}$
- [2] Excluding loss of resistive matching circuit, to match $75\ \Omega$ to $50\ \Omega$
- [3] Measured at the output of the output balun
- [4] Assume 3 dB loss between by output of the balun and F-connector in the final application

Table 16. ACLR quick reference data

Typical values at $V_{CC} = 5\text{ V}$, decimal current setting = 3, decimal gain setting 60, $T_{amb} = 25\text{ °C}$; $Z_{i(dif)} = 200\ \Omega$; $Z_{o(se)} = 75\ \Omega$; channel bandwidth = 196 MHz, integration bandwidth = 9.6 MHz, $f = 5\text{ MHz}$ to 205 MHz. Unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
DOCSIS 3.1						
ACLR	adjacent channel leakage ratio	$P_i(\text{RMS}) = 31\text{ dBmV}$; $P_L(\text{RMS}) = 68\text{ dBmV}$, channel configuration: channel bandwidth is 192 MHz, with exclusion band at 147.5 MHz, with a bandwidth of 9.6 MHz. Input signal with a PAPR of 13 dB	-	-62	-	dBc

14 Application information

14.1 External components

Matching the balanced output of the chip to a single-ended 75 Ω load is accomplished using a 1: 2 ratio transformer. For measurements in a 50 Ω system, R5 and R6 are added for impedance transformation from 75 Ω to 50 Ω. R5 and R6 are not required in the final application.

The transformer also cancels even mode distortion products and common mode signals, such as the voltage transients that occur while enabling and disabling the amplifiers.

External capacitors are needed for the functionality of the circuit, the pins are internal nodes in the output amplifier. The measured voltage on the temperature sense pin 16 at an input current of 1 mA, is related to the die temperature.

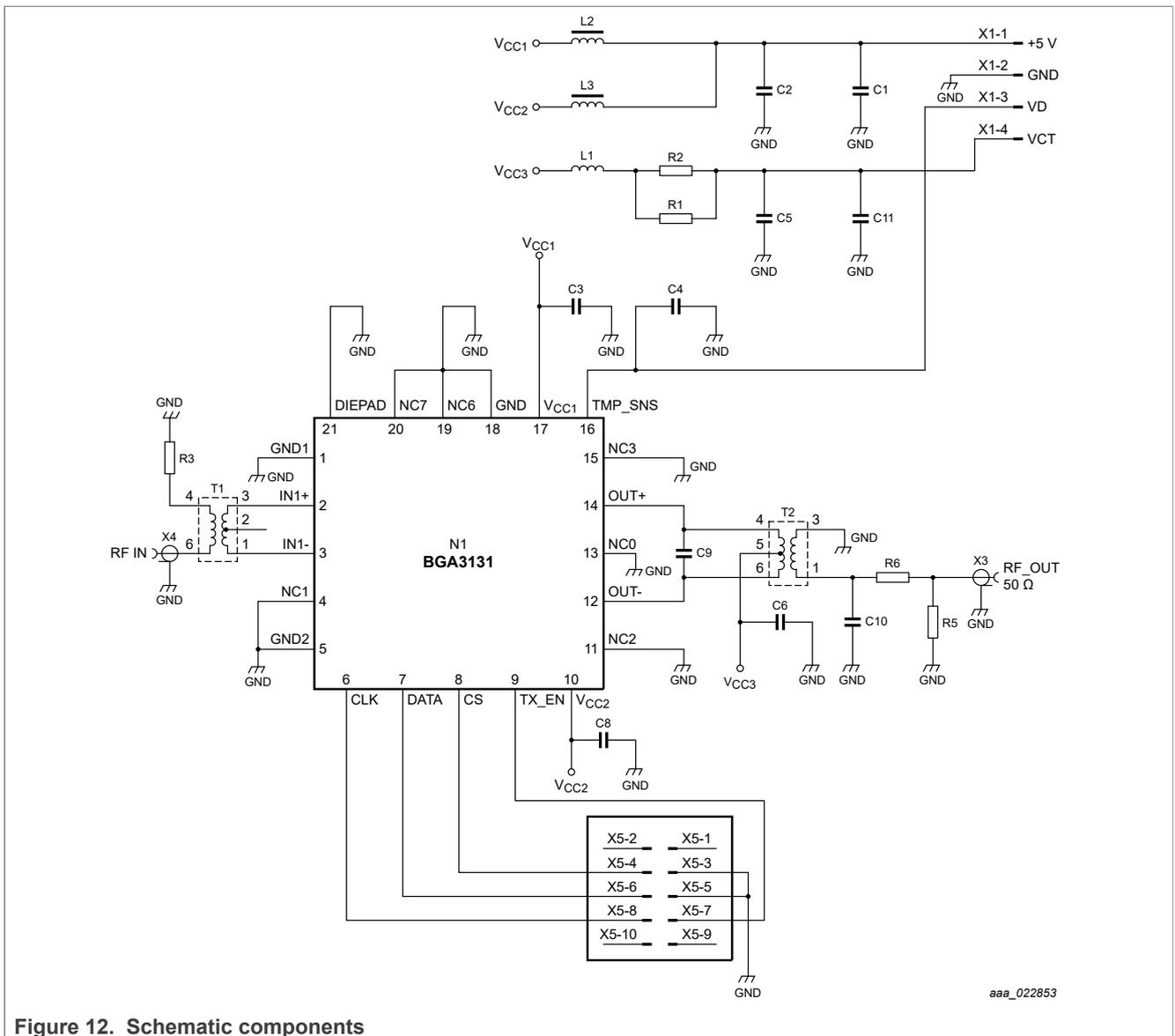


Figure 12. Schematic components

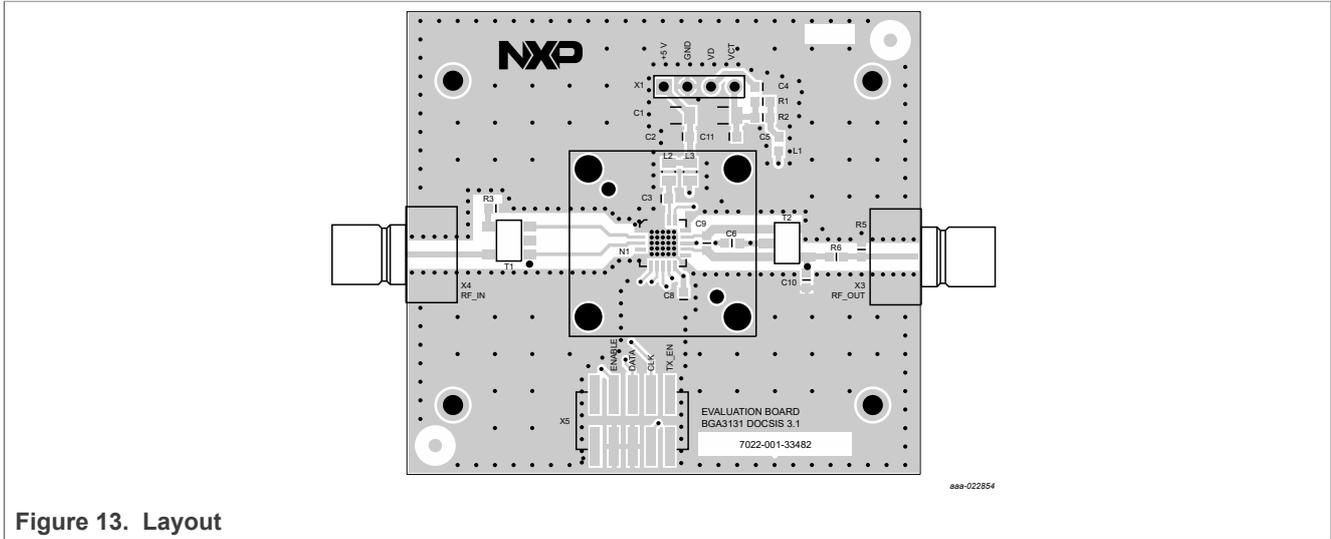


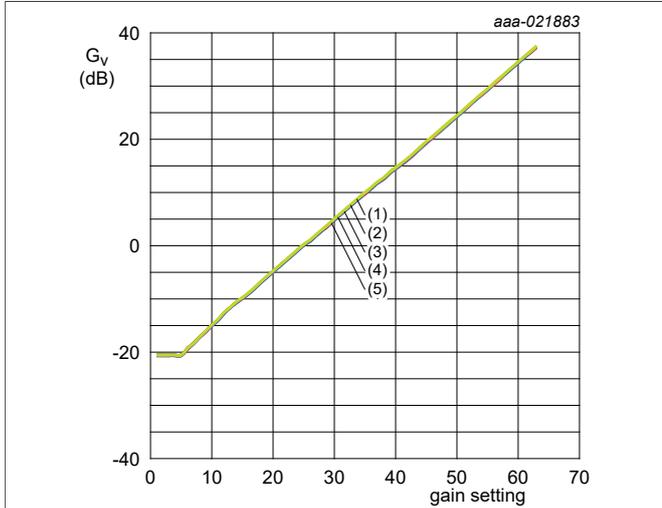
Figure 13. Layout

Table 17. List of components

For application diagram, see [Figure 12](#).

Component	Description	Value	Size	Supplier: Part No.
C1, C11	capacitor	10 μ F	SMD 1206	X7R type
C2, C5	capacitor	100 nF	SMD 0603	X7R type
C3, C4, C6,	capacitor	10 nF	SMD 0603	X7R type
C8	capacitor	56 nF	SMD 0603	X7R type
C9	capacitor	10 pF	SMD 0603	C0G type
C10	capacitor	4.7 pF	SMD 0603	C0G type
L1	place holder for optional inductor	-	-	on EVB 0 Ω mounted
L2, L3	place holder for option chokes	-	-	on EVB 0 Ω mounted
N1	amplifier	-	-	NXP: BGA3131
R1, R2, and R3	resistor	0 Ω	SMD 0603	
R5	resistor	86.6 Ω	SMD 0603	75 Ω to 50 Ω conversion for measurement purpose only
R6	resistor	43.2 Ω	SMD 0603	75 Ω to 50 Ω conversion for measurement purpose only
T1	transformer	-	-	TOKO: #617PT-1664
T2	transformer	-	-	MACOM: MABA-011056
X1	Header, 4P,	-	-	
X3, X4	SMA connector	-	-	
X5	Header, 10P	-	-	

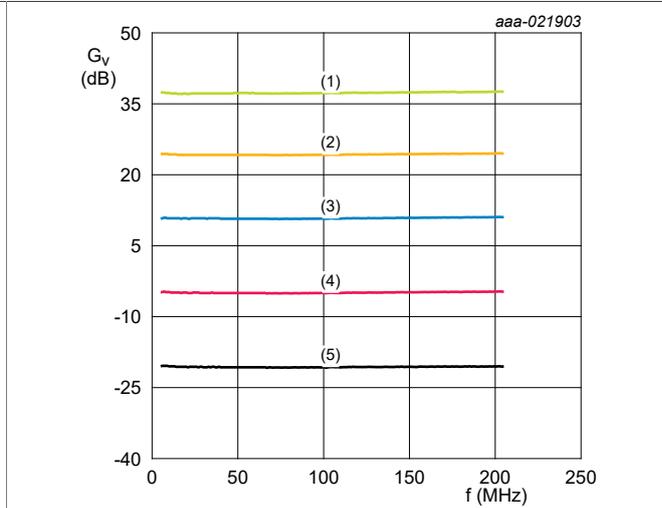
15 Graphics



$V_{CC} = 5\text{ V}$; current setting = 3; $T_{amb} = 25\text{ }^{\circ}\text{C}$

- (1) $f = 5\text{ MHz}$
- (2) $f = 50\text{ MHz}$
- (3) $f = 100\text{ MHz}$
- (4) $f = 150\text{ MHz}$
- (5) $f = 205\text{ MHz}$

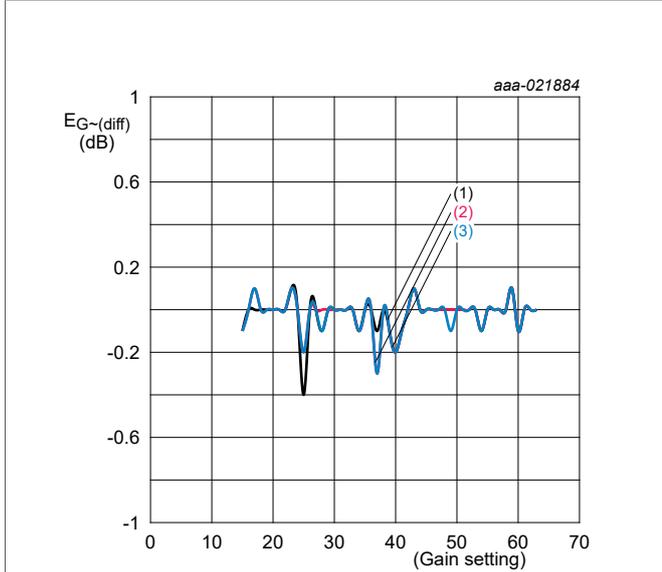
Figure 14. G_v as function of gain setting, typical values



$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^{\circ}\text{C}$

- (1) decimal gain setting = 63
- (2) decimal gain setting = 50
- (3) decimal gain setting = 36
- (4) decimal gain setting = 20
- (5) decimal gain setting = 5

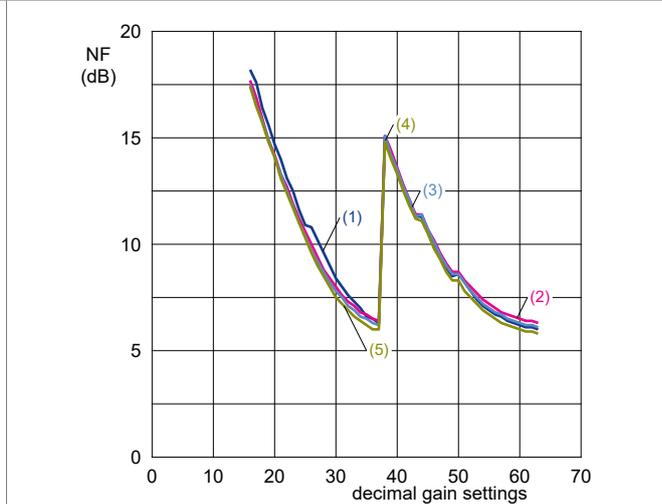
Figure 15. G_v as function of frequency, typical values



$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^{\circ}\text{C}$

- (1) $f = 5\text{ MHz}$
- (2) $f = 100\text{ MHz}$
- (3) $f = 205\text{ MHz}$

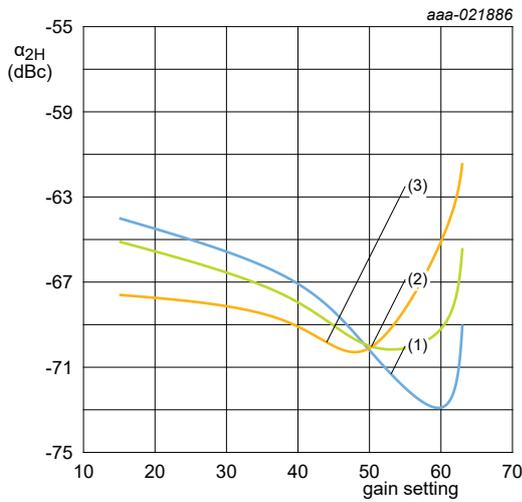
Figure 16. Differential gain error as function of gain setting; typical values



$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^{\circ}\text{C}$

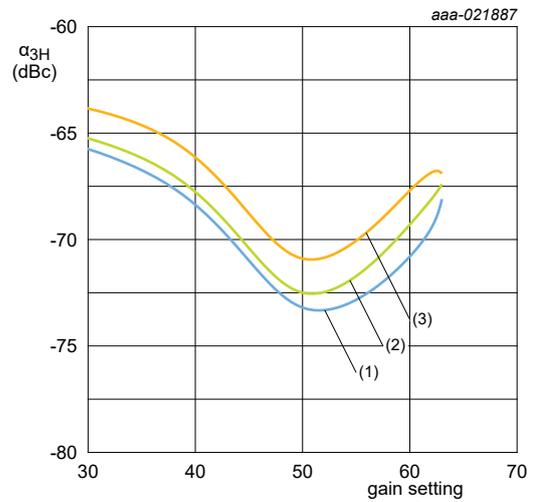
- (1) $f = 5\text{ MHz}$
- (2) $f = 50\text{ MHz}$
- (3) $f = 100\text{ MHz}$
- (4) $f = 150\text{ MHz}$
- (5) $f = 205\text{ MHz}$

Figure 17. Noise figure as function of gain setting; typical values



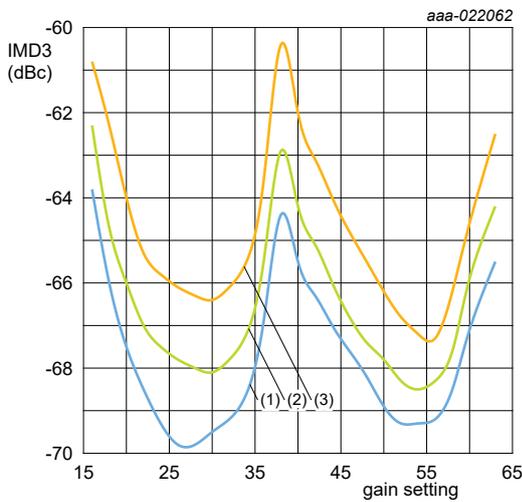
$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^\circ\text{C}$;
 $P_L = 68\text{ dBmV}$; $P_i = 31\text{ dBmV}$, Fix P_{in} at 68 dBmV,
 $P_o = >$ lower gain setting
 (1) $T_{amb} = -40\text{ }^\circ\text{C}$
 (2) $T_{amb} = 25\text{ }^\circ\text{C}$
 (3) $T_{amb} = 85\text{ }^\circ\text{C}$

Figure 18. Second order harmonic level as a function of gain setting; typical values



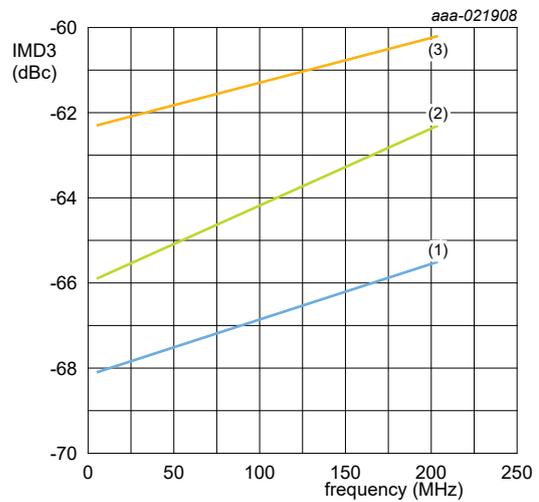
$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^\circ\text{C}$;
 $P_L = 68\text{ dBmV}$; $P_i = 31\text{ dBmV}$, Fix P_i at 68 dBmV,
 $P_o = >$ lower gain setting
 (1) $T_{amb} = -40\text{ }^\circ\text{C}$
 (2) $T_{amb} = 25\text{ }^\circ\text{C}$
 (3) $T_{amb} = 85\text{ }^\circ\text{C}$

Figure 19. Third order harmonic level as a function of gain setting; typical values



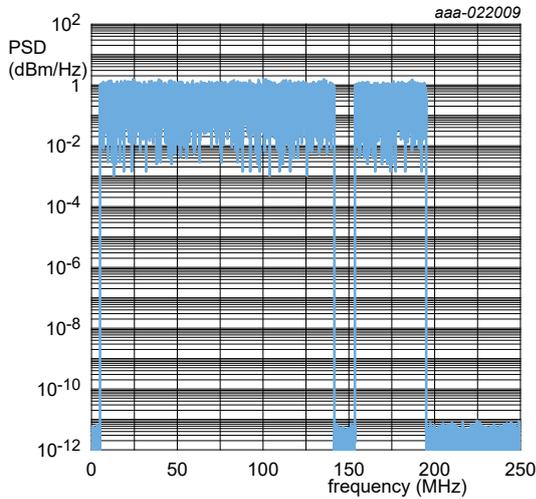
$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^\circ\text{C}$;
 P_L per tone = 65 dBmV, P_i per tone = 28 dBmV; 1 MHz tone
 space; Fix P_i at 65 dBmV, P_o / tone = $>$ lower gain setting
 (1) $f = 5\text{ MHz}$
 (2) $f = 100\text{ MHz}$
 (3) $f = 204\text{ MHz}$

Figure 20. Third order intermodulation distortion as a function of gain setting; typical values



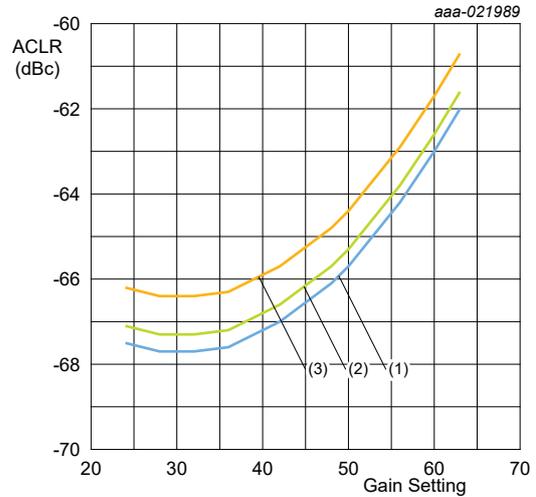
$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^\circ\text{C}$;
 P_L per tone = 65 dBmV, P_i per tone = 28 dBmV; 1 MHz tone
 space; Fix P_i at 65 dBmV, P_o / tone = $>$ lower gain setting
 (1) $T_{amb} = -40\text{ }^\circ\text{C}$
 (2) $T_{amb} = 25\text{ }^\circ\text{C}$
 (3) $T_{amb} = 85\text{ }^\circ\text{C}$

Figure 21. Third order intermodulation distortion as a function of temperature; typical values



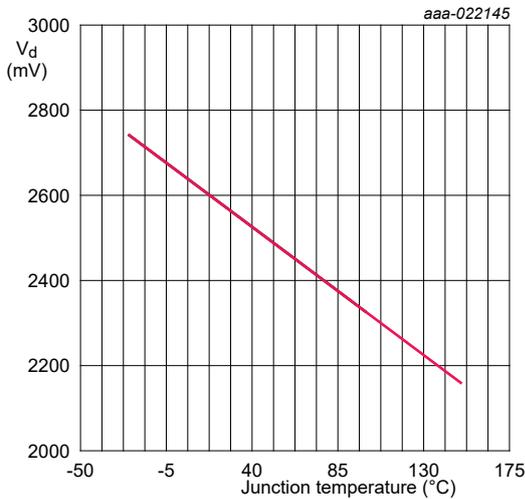
OFDM input signal for ACLR measurement
 total bandwidth = 192 MHz
 exclusion band = 147.5 MHz
 all with a bandwidth of 9.6 MHz.
 Peak-to-average (PAR) ratio of the signal = 13 dB

Figure 22. Power Spectral Density of input signal used for ACLR measurements



$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^{\circ}\text{C}$
 Up to decimal gain setting = 60; $P_i = 31\text{ dBmV}$
 At decimal gain settings = 61, 62, and 63; $P_o = 68\text{ dBmV}$
 Input signal applied as listed in [Figure 22](#)
 (1) $T_{amb} = -40\text{ }^{\circ}\text{C}$
 (2) $T_{amb} = 25\text{ }^{\circ}\text{C}$
 (3) $T_{amb} = 85\text{ }^{\circ}\text{C}$

Figure 23. Adjacent channel leakage ratio as function of temperature; typical values

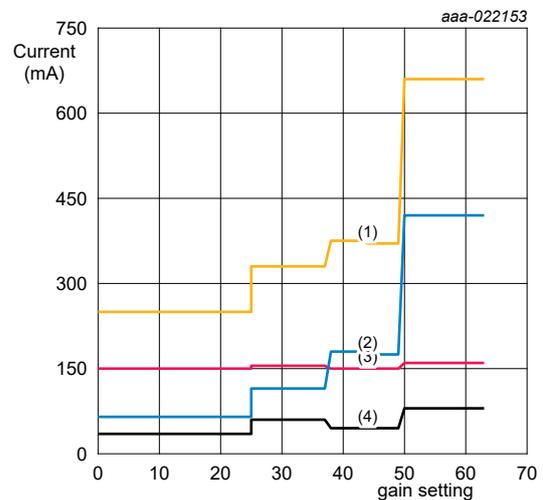


$V_{CC} = 5\text{ V}$; decimal current setting = 3
 TMP_SENS; DC current = 1 mA
 decimal gain setting = 63

Note:

For more thermal details, refer to the BGA3131 Thermal management guidelines AN11753 at www.nxp.com.

Figure 24. Thermal diode voltage as function of junction temperature; typical values



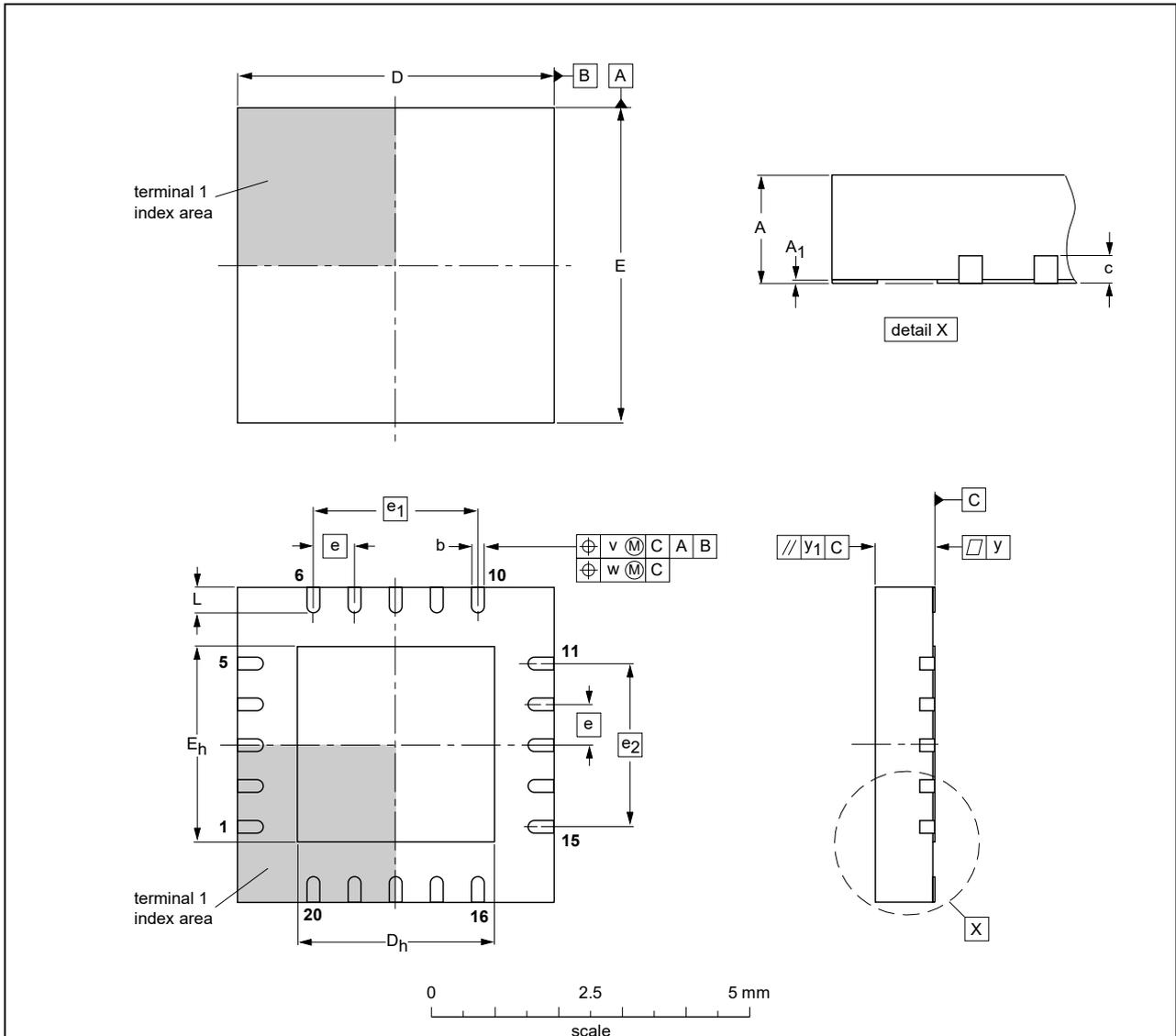
$V_{CC} = 5\text{ V}$; decimal current setting = 3; $T_{amb} = 25\text{ }^{\circ}\text{C}$
 (1) total current; $I_{CC1} + I_{CC2} + I_{CT}$
 (2) ICT; output balun center-tap current
 (3) I_{CC1} ; current through Pin 17 (V_{CC1})
 (4) I_{CC2} ; current through Pin 16 (V_{CC2})

Figure 25. DC-current as function of decimal gain setting; typical values

16 Package outline

HVQFN20: plastic thermal enhanced very thin quad flat package; no leads;
20 terminals; body 5 x 5 x 0.85 mm

SOT662-1



DIMENSIONS (mm are the original dimensions)

UNIT	A ⁽¹⁾ max.	A ₁	b	c	D ⁽¹⁾	D _h	E ⁽¹⁾	E _h	e	e ₁	e ₂	L	v	w	y	y ₁
mm	1	0.05 0.00	0.38 0.23	0.2	5.1 4.9	3.25 2.95	5.1 4.9	3.25 2.95	0.65	2.6	2.6	0.75 0.50	0.1	0.05	0.05	0.1

Note

1. Plastic or metal protrusions of 0.075 mm maximum per side are not included.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	JEITA			
SOT662-1	---	MO-220	---			-01-08-08- 02-10-22

Figure 26. Package outline SOT662-1 (HVQFN20)

16.1 Footprint and solder information

NXP recommends by default to apply the soldering and footprint guidelines as are released in POD SOT2013-2.

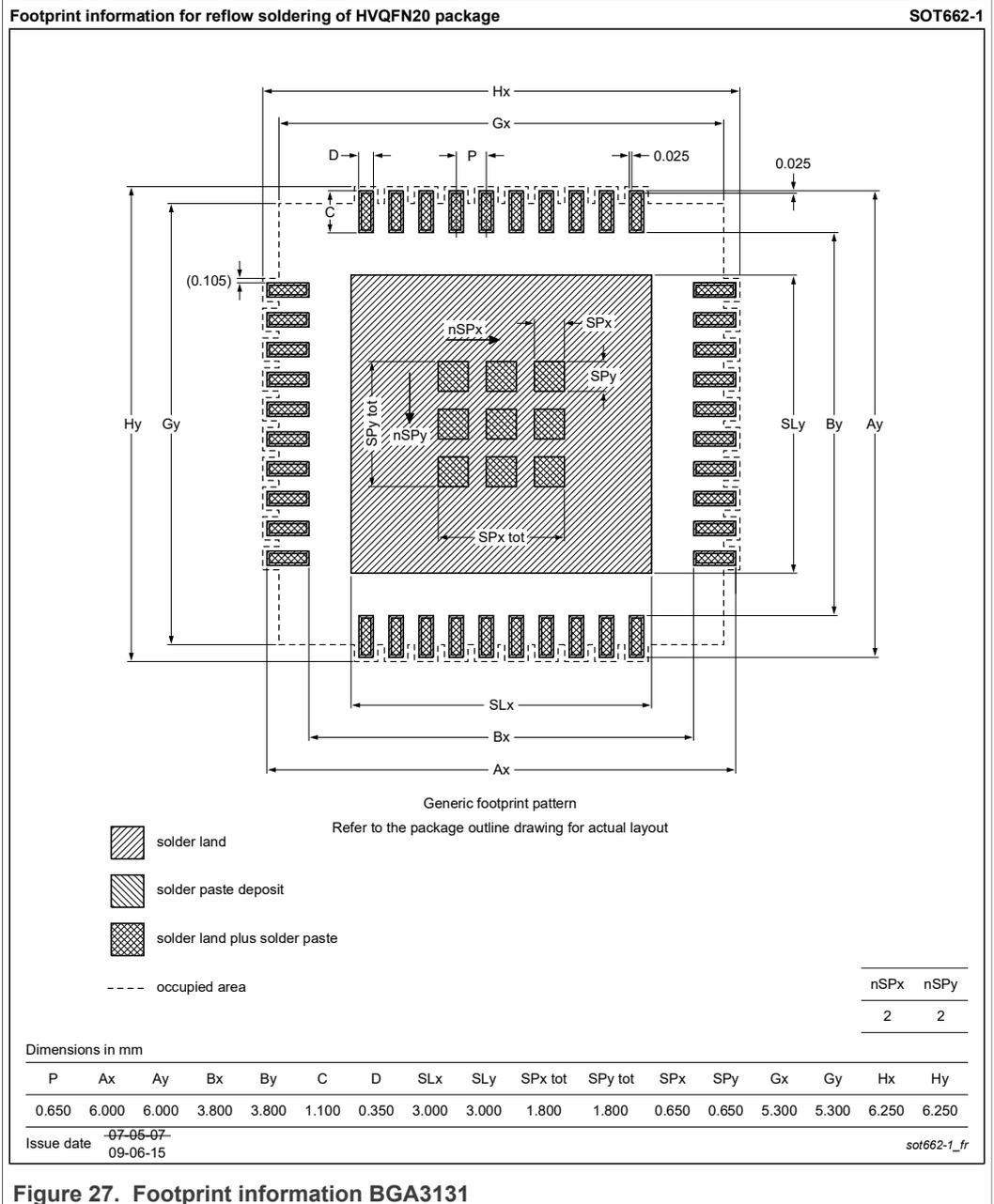


Figure 27. Footprint information BGA3131

17 Handling information

17.1 Moisture sensitivity

Table 18. Moisture sensitivity level

Test methodology	Class
JESD-22-A113	MSL1

17.2 ESD information

CAUTION



This device is sensitive to ElectroStatic Discharge (ESD). Observe precautions for handling electrostatic sensitive devices. Such precautions are described in the *ANSI/ESD S20.20*, *IEC/ST 61340-5*, *JESD625-A* or equivalent standards.

18 Abbreviations

Table 19. Abbreviations

Acronym	Description
ACLR	adjacent channel leakage ratio
CATV	community antenna television
CW	continuous wave
DOCSIS	Data Over Cable Service Interface Specification
ESD	electroStatic discharge
HVQFN	heat sink very thin quad flat pack no leads
OFDM	orthogonal frequency division multiplexing
PAPR	peak-to-average power ratio
SMA	subminiature version A
RoHS	restriction of hazardous substances
SMD	surface-mounted device
TX	transmission
VGA	variable gain amplifier
VoIP	voice over Internet protocol

19 Revision history

Table 20. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
BGA3131 v.3	20210430	Product data sheet	CIN 2021040571	BGA3131 v.2
modification	• updated the Revision history with the CIN number			
BGA3131 v.3	20210301	Product data sheet	CIN	BGA3131 v.2

Table 20. Revision history...continued

Document ID	Release date	Data sheet status	Change notice	Supersedes
modification				
				<ul style="list-style-type: none"> aligned the description above the Quick reference tables and Characteristic tables corrected conditions on IMD3, ACLR, second, and third harmonics added conditions to some graphics aligned the information in the Features and benefits with the Characteristic values for α_{2H}, α_{3H}, and IMD3 changed the T_{case} overall in the data sheet to T_{amb} changed value of C8 in the application schematic to 56 nF added orderable part number added marking code changed decoupling capacitor C8 (at V_{CC2}) in the application schematic from 10 nF to 56 nF. The value is changed to improve transient voltage behavior by using ferrite beads on V_{CC} lines aligned the line number in the graphics with the number in the description and change the colors of the lines to the new NXP format
BGA3131 v.2	20201229	Product data sheet	-	BGA3131 v.1
modification				
				<ul style="list-style-type: none"> overrides existing Product data sheet rev. 1 of 13 May 2016
BGA3131 v.1	20160513	Product data sheet		BGA3131 v.2
modification				
				<ul style="list-style-type: none"> changed Preliminary Rev.2 to Product data sheet Rev.1 corrected the data on α_{2H}, α_{3H}, IMD3
BGA3131 v.2		Preliminary data sheet		BGA3131 v.1
Modification: From objective to Preliminary data sheet				
BGA3131 v.1	<td>	Objective data sheet	-	-

20 Legal information

20.1 Data sheet status

Document status ^{[1][2]}	Product status ^[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

[3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL <http://www.nxp.com>.

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