3A, 2.4MHz, Low-Voltage, I²C Programmable Buck Regulator

Features

- 2.5 to 5.5V Input Voltage Range
- 0.6 to 3.345V Programmable Option for Vout
  - 6.25mV steps below 1.39375V
  - 15mV steps above 1.44V
- 3.0A Output Current
- ±1% Accuracy at T_A = +25°C
  - ±2.5% over line/load/temp/setting
- Fast Transient Response
- Dynamic Voltage Scaling (DVS) with 8 ramp rates
- 90% Peak Efficiency at Vout = 1.136V
- 2.4MHz with Auto-Skip at light loads
  - Programmable forced-PWM mode
- 46uA typ. No-Load Supply Current in Skip Mode
- Tiny External Components
  - L = 330 or 470nH (2012 or 2016 metric size)
  - Cin = 10µF (0402), Cout = 2x22µF (2x0402)
- Soft Start, Over-Current, Short-Circuit, Under/Over-V_{IN}, and Thermal Shutdown Protections
- 1MHz I²C Interface
- -40°C to 85°C Operating Temperature Range
- 15-bump Pb-free WLCSP (0.4mm pitch)
  - 1.340 x 2.045mm (0.6mm height)
  - Pin/Register Compatible with FAN53527

Brief Description

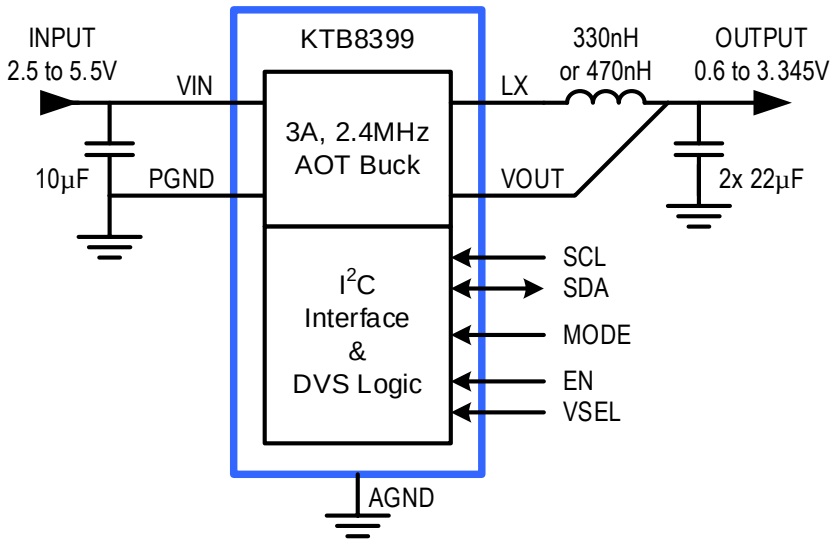
The KTB8399 is a precision adaptive-on-time (AOT) buck switching regulator with class-leading accuracy, transient response, efficiency, and solution size optimized for mobile and non-mobile application. It is I²C programmable for output voltages in the 0.6V to 3.345V range. It features soft-start and DVS with multiple programmable ramp rates. Versions with various default settings can be ordered. The features and performance make the KTB8399 suitable for a variety of applications including CPU/GPU core, DSP and baseband, DDR memory, VIO, and sensor/analog power.

The KTB8399 is available in RoHS and Green compliant 15-bump 1.340mm x 2.045mm x 0.6mm wafer-level chip-scale package (WLCSP).

Applications

- CPU, GPU, AP, DSP, FPGA, I/O, XCVR Power
- HDD, LPDDR3, LPDDR4, LPDDR5 Memory Power
- Tablets, Netbooks, Ultra-Books
- Smartphones, Mobile Internet Devices, IoT
- DSC, Drones, Gaming Consoles, Accessories

Typical Application Schematic

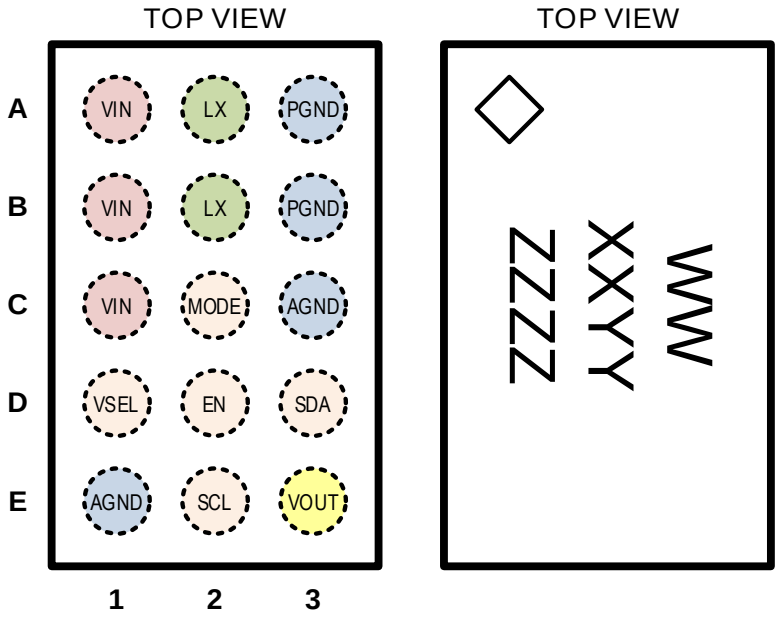


Pin Descriptions

Pin #	Name	Function
A1, B1, C1	VIN	Voltage Input for buck regulator and IC power
A3, B3	PGND	Power Ground for buck regulator
C3, E1	AGND	Analog Ground for IC power
A2, B2	LX	Inductor Connection for buck regulator
E3	VOUT	Output Voltage sense input
D2	EN	Chip Enable logic input
E2	SCL	I^2C Clock digital input
D3	SDA	I^2C Data digital I/O
C2	MODE	Operation mode select MODE = 0(LOW): The IC follows register bits MODE0 and MODE1 for Auto-Skip or Forced-PWM operation MODE = 1(HIGH): The IC enters Forced-PWM mode regardless of register bit MODE0 and MODE1 status. Do not let this pin float.
D1	VSEL	DVS Voltage Select and Auto-Skip vs. Forced-PWM Mode Select logic input

Pinout Diagram

WLCSP-15



15-Bump 1.340mm x 2.045mm x 0.62mm
WLCSP Package, 0.4mm pitch

Top Mark
WW = Device Code,
XX = Date Code, YY = Assembly Code,
ZZZZ = Serial Number

Absolute Maximum Ratings[1]

(T_A = 25°C unless otherwise noted)

Symbol	Description	Value	Units
V_{IN}	VIN to AGND	-0.3 to 6	V
V_{PGND}	PGND to AGND	-0.3 to 0.3	V
V_{LX}[2]	LX to PGND	-0.3 to (V_{IN}+0.3)	V
V_{OUT}	VOUT to AGND	-0.3 to (V_{IN}+0.3)	V
V_{IO}	SCL, SDA, VSEL to AGND	-0.3 to 6	V
	EN, MODE to AGND	-0.3 to V_{IN}	V
I_{LX}	LX Continuous Current	3.2	A_{RMS}
	LX Peak Current (1ms maximum)	9.6	A_{PEAK}
T_J	Operating Junction Temperature Range	-40 to 150	°C
T_S	Storage Temperature Range	-55 to 150	°C
T_{LEAD}	Maximum Soldering Temperature (at leads, 10 sec)	260	°C

ESD Ratings[3]

Symbol	Description	Value	Units
V_{ESD_HBM}	JEDEC JS-001-2017 ESD Human Body Model (all pins)	±2	kV

Thermal Capabilities[4]

Symbol	Description	Value	Units
Θ_{JA}	Thermal Resistance – Junction to Ambient	77	°C/W
P_D	Maximum Power Dissipation at T_A = 25°C	1.62	W
$\Delta P_D/\Delta_T$	Derating Factor Above T_A = 25°C	-13	mW/°C

Ordering Information

Part Number[5]	Marking[6]	Default Output Voltage & Mode[7]		7-bit I²C Slave Address	Package
		(VSEL = 1)	(VSEL = 0)		
KTB8399MEDAA-TR	*RKXXYYZZZZ*	*1.081V Auto-Skip*	*1.125V Auto-Skip*	*1100 000=0x60h*	*WLCSP15*
KTB8399NEDAA-TR	RTXXYYZZZZ	1.3V Forced-PWM	1.3V Auto-Skip	1100 000=0x60h	WLCSP15
KTB8399PEDAA-TR	RUXXYYZZZZ	2.05V Forced-PWM	2.05V Auto-Skip	1100 000=0x60h	WLCSP15
KTB8399QEDAA-TR	*RVXXYYZZZZ*	*1.8V Forced-PWM*	*1.8V Auto-Skip*	*1100 000=0x60h*	*WLCSP15*
KTB8399REDAA-TR	*RWXXYYZZZZ*	*1.128V Forced-PWM*	*1.128V Auto-Skip*	*1100 001=0x61h*	*WLCSP15*
KTB8399TEDAA-TR	*SAXXYYZZZZ*	*0.8V Auto-Skip*	*1.15V Forced-PWM*	*1100 000=0x60h*	*WLCSP15*
KTB8399UEDAA-TR	*SBXXYYZZZZ*	*0.8V Auto-Skip*	*1.15V Forced-PWM*	*1100 001=0x61h*	*WLCSP15*
KTB8399VEDAA-TR	*SCXXYYZZZZ*	*0.6V Forced-PWM*	*0.6V Auto-Skip*	*1100 000=0x60h*	*WLCSP15*
KTB8399XEDAA-TR	SOXXYYZZZZ	1.913V Auto Skip	1.825V Auto-Skip	1100 000=0x60h	WLCSP15
KTB8399YEDAA-TR	UOXXYYZZZZ	1.913V Auto Skip	1.825V Auto-Skip	1100 001=0x61h	WLCSP15

1. Stresses above those listed in Absolute Maximum Ratings may cause permanent damage to the device. Functional operation at conditions other than the operating conditions specified is not implied. Only one Absolute Maximum rating should be applied at any one time.
2. Absolute Maximum Rating for VLX is a DC voltage rating. During normal switching operation, short-duration voltage spikes beyond the DC rating are expected and normal for all DC-DC switching regulators.
3. ESD Ratings conform to JEDEC industry standards. Some pins may actually have higher performance.
4. Junction to Ambient thermal resistance is highly dependent on PCB layout. Values are based on thermal properties of the device when soldered to an EV board.
5. For part numbers in Italics, please contact your local sales representative for availability
6. "WW" is the device ID, "XX" is the date code, "YY" is the assembly code, and "ZZZZ" is the serial number.
7. Contact a Kinetic Technologies representative regarding versions with other default settings.

Electrical Characteristics[8]

Unless otherwise noted, the *Min* and *Max* specs are applied over the full operation temperature range of -40°C to +85°C and V_{IN} = 2.5V to 5.5V. *Typ* values are specified at +25°C with V_{IN} = 3.6V.

Supply Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
V_{IN}	Input Supply Operating Range		2.5		5.5	V
V_{UVLO}	Under-Voltage Lockout Threshold	V_{IN} rising V_{IN} hysteresis	2.15	2.3 200	2.49	V mV
V_{OVP}	Over-Voltage Protection Threshold	V_{IN} rising	5.55	5.75	5.95	V
		V_{IN} hysteresis		200		mV
I_{IN}	No-Load Supply Current	EN = 1, V_{IN} = 3.6V, V_{OUT} ≤2.5V, Skip		46	80	µA
		EN = 1, V_{IN} = 3.6V, Forced-PWM		20		mA
I_{SHDN}	Shutdown Supply Current	EN = 0, T_A = 25°C		0.2	1	µA
		EN = 1, ENn[0]=0, T_A = 25°C		0.2	1.5	µA

Logic Pin Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
MODE						
V_{IH}	Input Logic High		0.6V_{IN}			V
V_{IL}	Input Logic Low				0.3V_{IN}	V
EN, VSEL						
V_{IH}	Input Logic High		1.15			V
V_{IL}	Input Logic Low				0.4	V
I_{I_LK}	Input Logic Leakage	T_A = 25°C, V_I = 0V or V_{IN}	-1	±0.01	1	µA
R_{I_PD}	Input Logic Pull-Down (EN, VSEL)	only connected when $V_I \le V_{IL}$ (disconnected when $V_I \ge V_{IH}$)		250		kΩ

Thermal Shutdown Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
T_{J_SHDN}	IC Junction Thermal Shutdown[9]	T_J rising		150		°C
		Hysteresis		20		°C

(continued next page)

8. Device is guaranteed to meet performance specifications over the -40°C to +85°C operating temperature range by design, characterization and correlation with statistical process controls.

9. Guaranteed by design, characterization and statistical process control methods; not production tested.

Electrical Characteristics (continued)[8]

Unless otherwise noted, the *Min* and *Max* specs are applied over the full operation temperature range of -40°C to +85°C and V_{IN} = 2.5V to 5.5V. *Typ* values are specified at +25°C with V_{IN} = 3.6V.

I²C-Compatible Interface Specifications (SCL, SDA), see Figure 1

Symbol	Description	Conditions	Min	Typ	Max	Units
V_{IH}	Input Logic High Threshold		1.1			V
V_{IL}	Input Logic Low Threshold				0.4	V
V_{OL}	SDA Output Logic Low	I_{SDA} = 3mA			0.4	V
t_1	SCL clock period[9]		2.5			µs
t_2	Data in setup time to SCL high[9]		100			ns
t_3	Data out stable after SCL low[9]		0			ns
t_4	SDA low setup time to SCL low (Start)[9]		100			ns
t_5	SDA high hold time after SCL high (Stop)[9]		100			ns

(continued next page)

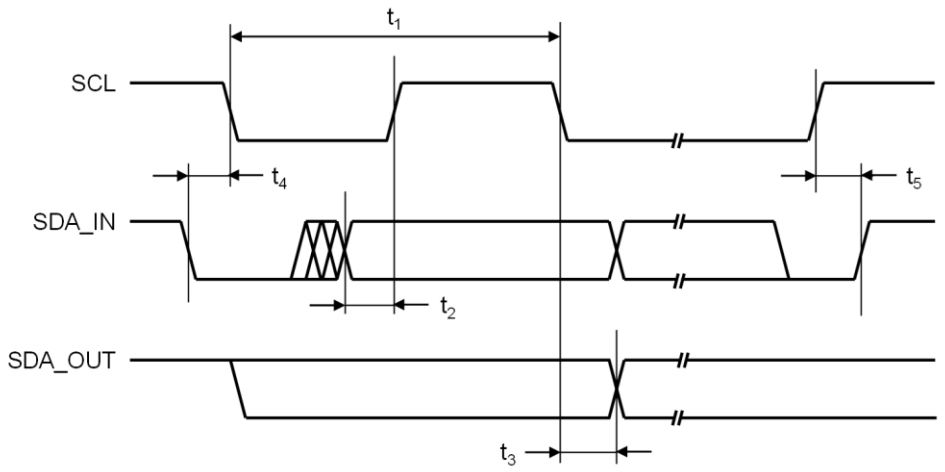


Figure 1. I²C Compatible Interface Timing

Electrical Characteristics (continued)[8]

Unless otherwise noted, the *Min* and *Max* specs are applied over the full operation temperature range of -40°C to +85°C and V_{IN} = 2.5V to 5.5V. *Typ* values are specified at +25°C with V_{IN} = 3.6V.

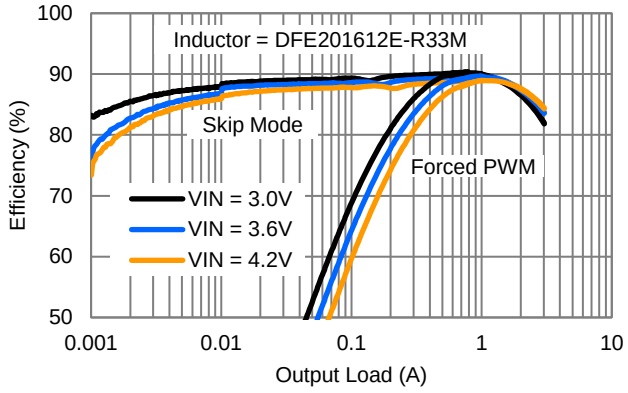
Buck Regulator Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
V_{OUT}	Output Voltage Setting Range	Low Range (G=1), RANGE[0]=0	0.600		1.39375	V
		High Range (G=2.4), RANGE[0]=1	1.440		3.345	V
V_{OUT_STEP}	Output Voltage Setting Step Size	Low Range (G=1), RANGE[0]=0		6.25		mV
		High Range (G=2.4), RANGE[0]=1		15		mV
V_{OUT_ACC}	Output Voltage DC Accuracy	T_A = 25°C, V_{OUT} = 1.136V, FPWM	-1	±0.2	1	%
		FPWM over line/load/temp/setting	-2.5	±1	2.5	%
		Auto-Skip over line/setting		±1		%
V_{OUT_LOAD}	Output Voltage Load Regulation[9]	I_{LOAD} = 1A to 3A, FPWM		-0.01		%/A
V_{OUT_LINE}	Output Voltage Line Regulation[9]	V_{IN} = 2.5V to 5.5V, I_{LOAD} = 1.5A		±0.01		%/V
V_{OUT_TRAN}	Output Voltage Transient Response[9]	I_{LOAD} = 10mA↔1.5A, $t_r = t_f$ = 200ns, V_{OUT} = 1.136V		±50		mV
I_{OUT_MAX}	Maximum Output Current		3			A
I_{LX_PEAK}	LX Peak Current Limit		4	4.6	5.5	A
I_{LX_VALLEY}	LX Valley Current Limit		3	4.3	5.5	A
I_{LX_LEAK}	LX Leakage Current	V_{LX} = 0V or 5.5V, T_A = 25°C		0.1	1	µA
R_{DSON_MS}	Main Switch On-Resistance	MOSFET + metal + bumps		50		mΩ
R_{DSON_SR}	Synch. Rectifier On-Resistance	MOSFET + metal + bumps		30		mΩ
R_{LX_DIS}	LX Active Discharge Resistance	EN = 0 or ENn[0] = 0, DIS[0] = 1		133		Ω
t_{SS_DELAY}	Soft-Start Ramp Delay[9]	Hardware enable, EN=0→1		170		µs
		Software enable, ENn[0] = 0→1				
dV/dt_{SS}	Soft-Start Ramp Rates	V_{IN} = 3.6V, RANGE[0]=0		18		mV/µs
		V_{IN} = 3.6V, RANGE[0]=1		38		
t_{DVS_DELAY}	DVS Ramp Delay[9]	Hardware DVS, VSEL = 0↔1		5		µs
		Software DVS, by I^2C command		<10		µs
dV/dt_{DVS}	DVS Ramp Rates	8 programmable rates	3.125		25	mV/µs
V_{OUT_POK}	V_{OUT} Power OK Threshold	Percentage of V_{OUT} setting, falling		88		%
		Percentage of V_{OUT} setting, rising		92		%

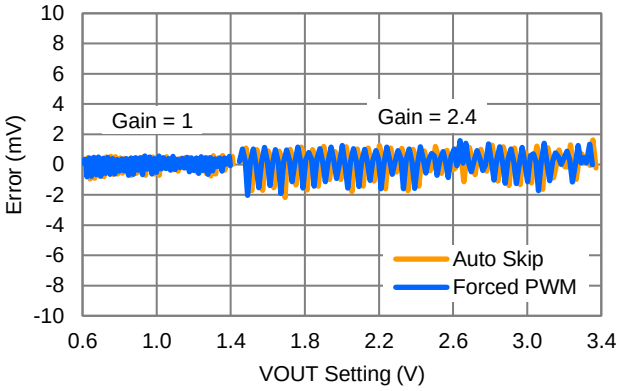
Typical Characteristics

Unless otherwise noted, V_{IN} = V_{EN} = 3.6V, L = 330nH, C_{IN} = 10µF, C_{OUT} = 2x 22µF, and T_A = 25°C.

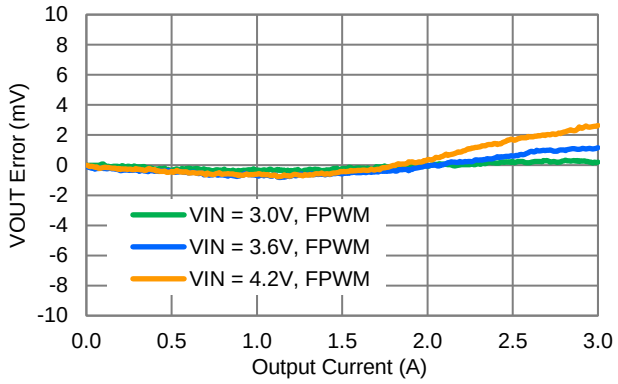
Efficiency vs. Load at V_{OUT} = 1.136V



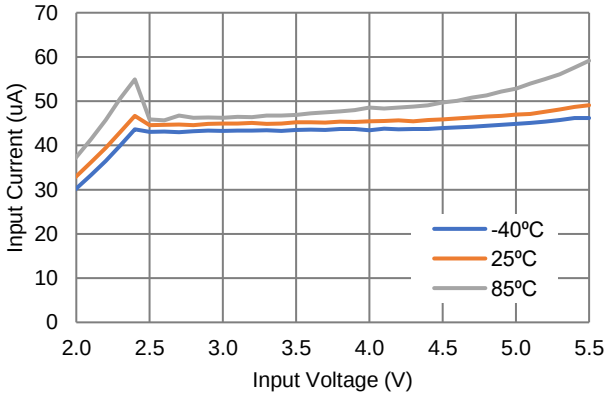
VOUT Setting Accuracy
(I_{OUT} = 10mA)



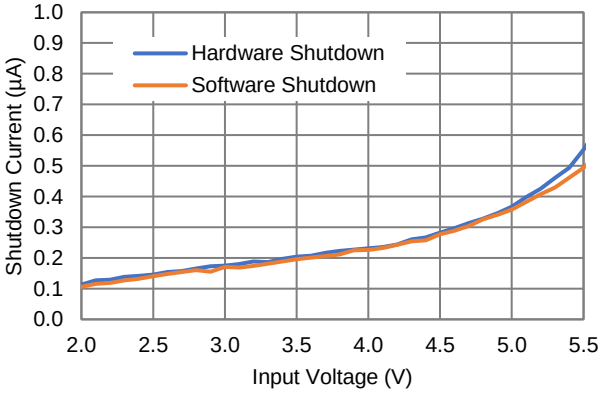
V_{OUT} Accuracy vs. Load (FPWM)
(V_{OUT} = 1.136V)



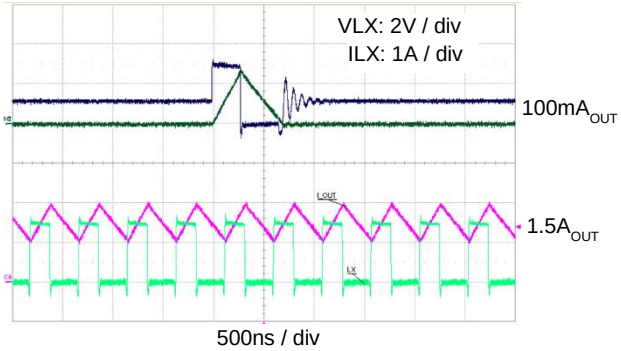
Supply Current vs. V_{IN}
(V_{OUT} = 1.136V, No Load, Mode = Auto)



Shutdown Supply Current vs. V_{IN}



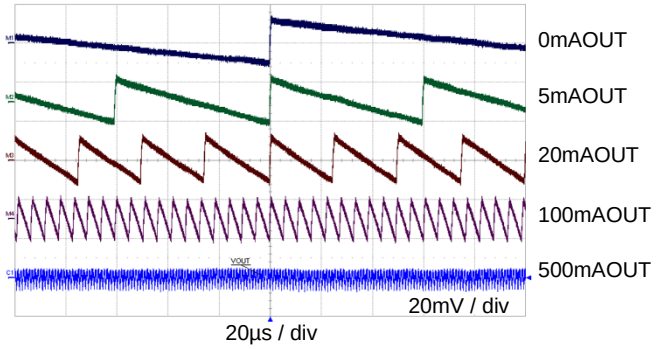
V_{LX} and I_{LX} Switching Waveforms
(V_{OUT} = 1.136V, Mode = Auto)



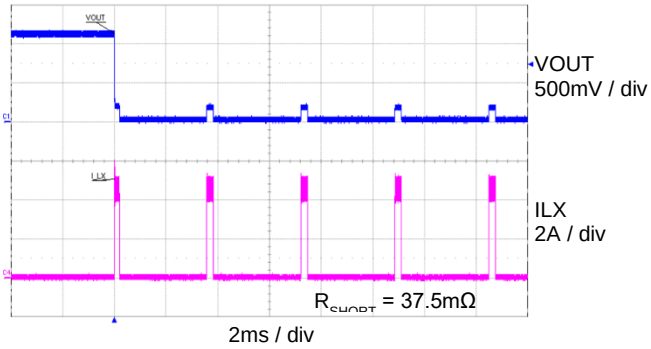
Typical Characteristics (continued)

Unless otherwise noted, V_{IN} = V_{EN} = 3.6V, L = 330nH, C_{IN} = 10µF, C_{OUT} = 2x 22µF, and T_A = 25°C.

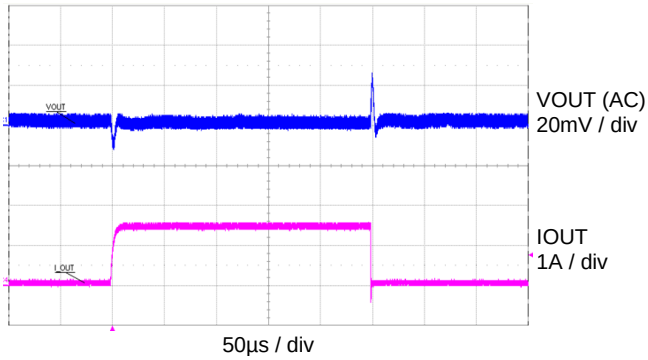
V_{OUT} Ripple Waveforms
(V_{OUT} = 1.136V, Mode = Auto)



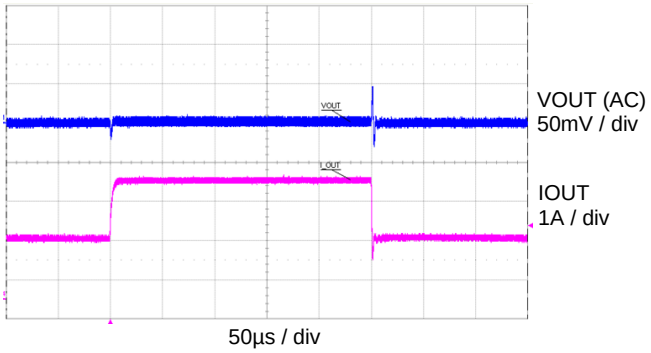
Short-Circuit Protection



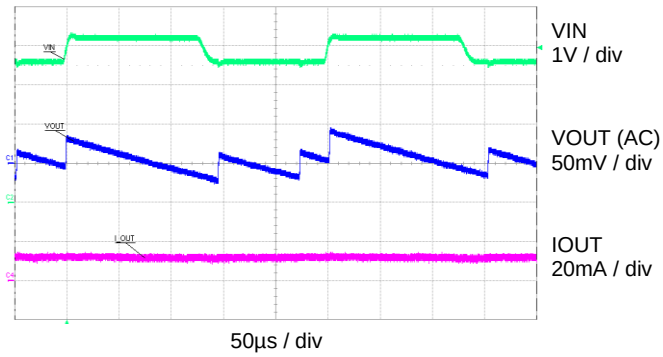
Load Transient Response
(V_{OUT} = 1.136V, Load = 10mA-1.5A, Mode = FPWM)



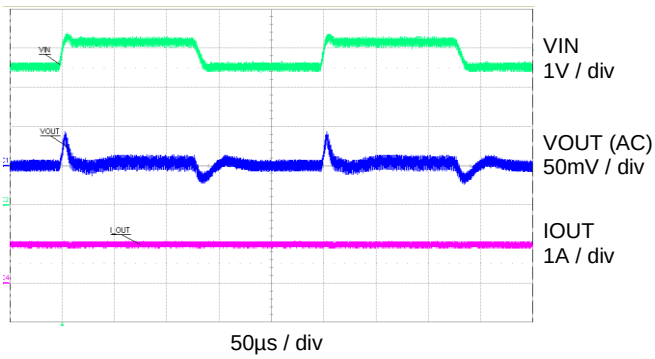
Load Transient Response
(V_{OUT} = 1.136V, Load = 1.5A-3A, Mode = Auto)



Line Transient Response
(V_{OUT} = 1.136V, Mode = Auto)



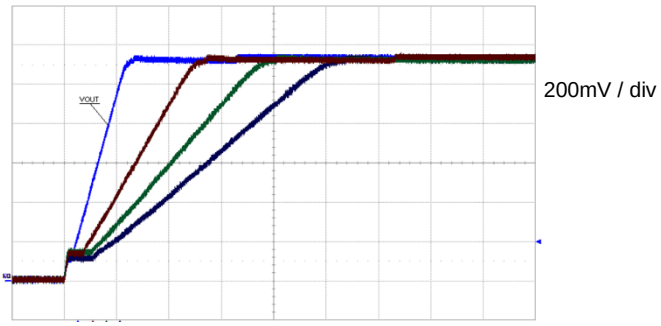
Line Transient Response
(V_{OUT} = 1.136V, Mode = FPWM)



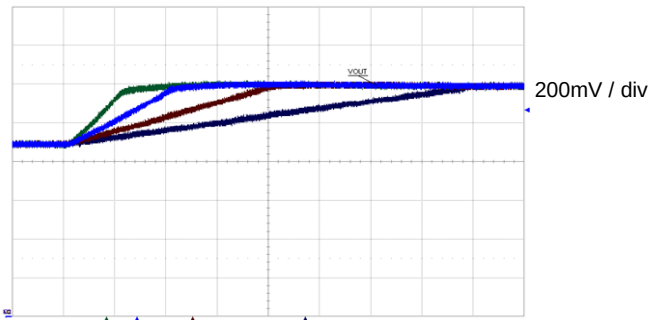
Typical Characteristics (continued)

Unless otherwise noted, V_{IN} = V_{EN} = 3.6V, L = 330nH, C_{IN} = 10µF, C_{OUT} = 2x 22µF, and T_A = 25°C.

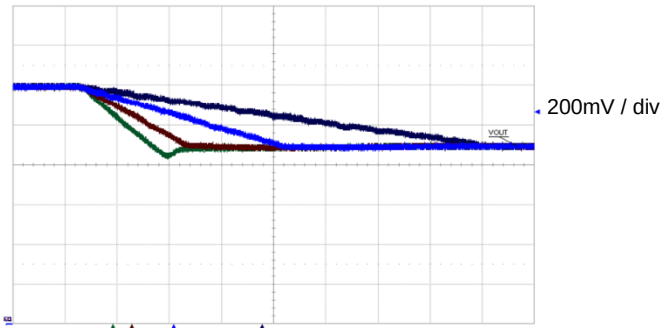
Buck Enable Soft-Start Rates
(V_{OUT} = 1.136V)



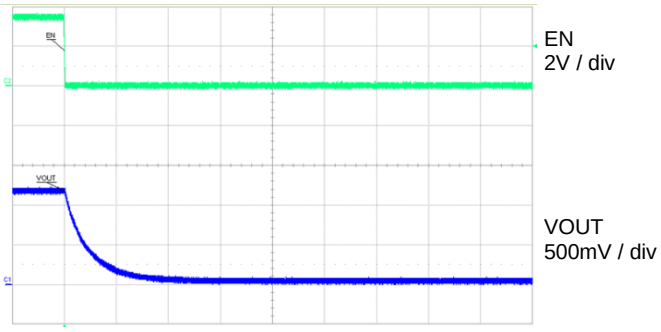
DVS Positive Ramp-Up Rates
(V_{OUT} = 0.9-1.2V)



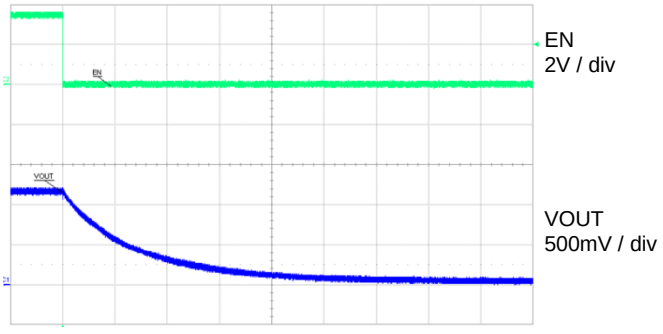
DVS Negative Ramp-Down Rates
(V_{OUT} = 1.2-0.9V)



Buck Shutdown
(V_{OUT} = 1.136V, Active Discharge Enabled)



Buck Shutdown
(V_{OUT} = 1.136V, Active Discharge Disabled)



Functional Description

The KTB8399 is a highly efficient, high-performance, small buck regulator that operates from an input voltage of 2.5V to 5.5V and can output up to 3A. It integrates the main switch, synchronous rectifier switch, PWM control circuitry, V_{OUT} setting DAC, various protection features, and an I²C serial interface to configure the output voltage, dynamic voltage scaling (DVS), control modes, and interrupts.

Control Scheme

The KTB8399 uses a proprietary adaptive on-time (AOT) PWM control scheme to maintain a nearly constant switching frequency as input voltage and output voltage vary. Compared to typical current-mode PWM schemes, the AOT control scheme provides quick response to line and load transients with excellent stability and wide bandwidth, thereby minimizing output voltage droop and soar for dynamic loads, even with minimal output capacitance. The adaptive on-time approximates fixed-frequency switching without using a fixed clock oscillator, which eliminates the need to wait for the next clock before responding to a load transient.

The KTB8399 feedback loop also adds a proprietary, internally-compensated, integrating error amplifier to remove the output voltage offset normally associated with other AOT, constant on-time (COT), and hysteretic architectures.

Shutdown Mode

When the EN pin is low, the KTB8399 buck is in shutdown mode and draws very little current. In shutdown, the I²C is still active, allowing read and write commands, as long as V_{IN} is above V_{UVLO}. Similarly, register contents are retained during shutdown (and even when V_{IN} is a little below V_{UVLO}), as long as V_{IN} is above V_{POR}=1.8V.

Hardware Enable

The KTB8399 buck regulator is turned on and off via hardware enable using the EN pin or via software enable using I²C commands. The default register settings allow simple hardware control. For hardware enable, drive the EN pin high. For hardware disable, drive the EN pin low.

Software Enable

For software enable/disable control, first write 0 into the EN0 bit (B7) of the VSEL0 configuration register (0x00), and/or write 0 into the EN1 bit (B7) of the VSEL1 configuration register (0x01). The VSEL pin determines which register is used. Then bring the EN pin high. After that, write 1 into the corresponding EN0 or EN1 bit to enable the buck, and write 0 to disable the buck.

Soft-Start

The KTB8399 buck contains soft-start circuitry to ramp up V_{OUT} slowly in order to reduce inrush current at V_{IN} and prevent the inductor current from reaching the peak current limit (I_{LX_PEAK}) during startup. The inductor's average current during soft-start is given by:

$$I_{L\_SS} = C_{OUT} \times (dV/dt_{SS}) + I_{LOAD\_SS}$$

where dV/dt_{SS} = 18mV/µs is the soft-start ramp rate and I_{LOAD_SS} is the load current during soft-start. Choose a C_{OUT} that keeps the inductor current average below 3A, or even lower.

Setting the Output Voltage

The KTB8399 has two independent output voltage ranges. The low range is V_{OUT} = 0.6V to 1.39375V in 6.25mV steps. The high range is V_{OUT} = 1.44V to 3.345V in 15mV steps. The range is I²C programmable using the RANGE bit (B3) in the CONTROL configuration register (0x02). The default range setting is factory trimmed to match the default output voltage setting – see the *Ordering Information* section.

After the correct range is selected, the KTB8399 register map contains two voltage setting registers to facilitate hardware control of DVS using the VSEL pin. The first output voltage setting is I²C programmable using the VOUT_VSEL0[6:0] bits (B6:0) in the VSEL0 configuration register (0x00). The second output voltage setting is I²C programmable using the VOUT_VSEL1[6:0] bits (B6:0) in the VSEL1 configuration register (0x01). The output voltage setting is given by:

$$V_{OUT} = G \times (600mV + 6.25mV \times VOUT\_VSELn)$$

where G = 1.0 when the RANGE bit is 0, and G = 2.4 when the RANGE bit is 1. VOUT_VSELn is the decimal equivalent of the binary bits of VOUT_VSEL0[6:0] or VOUT_VSEL1[6:0]. The default voltage settings are factory trimmed, and several versions are available – see the *Ordering Information* section.

Dynamic Voltage Scaling (DVS)

Dynamic Voltage Scaling (DVS) is used to slew the output voltage between two voltage settings contained in the VSEL0 and VSEL1 registers – see *Setting the Output Voltage* section. The VSEL pin selects which register is used. Or, as an alternative method, I^2C commands can be used by simply connecting the VSEL pin to ground and dynamically writing a new setting into the VSEL0 register.

In either method, all DVS output voltage transitions are slew-rate-controlled by an on-chip up/down counter and DAC. The DVS slew rate is I^2C programmable using the SLEW[2:0] bits (B6:4) in the CONTROL configuration register (0x02). DVS slewing is only possible within a single output voltage range; slewing between a voltage in the low range and a voltage in the high range is ***not*** possible.

Forced-PWM vs. Auto-Skip Modes

The KTB8399 has two ways to control light-load switching behavior – Forced-PWM mode and Auto-Skip mode. In Forced-PWM, the switching frequency remains nearly constant. This mode is helpful for applications that are noise sensitive.

In Auto-Skip mode, the KTB8399 transitions automatically between PWM switching at heavy loads and Skip/PFM switching at light loads. Auto-Skip mode is helpful for applications that need high efficiency at light loads. While skipping, single pulses are evenly spaced, resulting in the lowest output ripple and noise when compared to competing “pulse-grouping” or “burst mode” devices. Furthermore, the PFM frequency during single-pulse skipping remains above the audio frequency band (20Hz to 20kHz) down to very light loads – see the *Typical Operating Characteristics* section.

Auto-Skip mode can be disabled(set Forced-PWM mode) by programming the MODE[1:0] bits (B1:0) in the CONTROL configuration register (0x02) in combination with the state of the VSEL pin or simply by setting the MODE pin high. See Table 1. The default mode settings are factory trimmed, and several versions are available – see the *Ordering Information* section.

Table 1. Hardware/Software Enable and Operation Mode

Control Pins			ENx Bits		Mode Bits [1:0]	Operation	V_{OUT}
EN	VSEL	MODE	EN0	EN1			
0	X	X	X	X	XX	Shutdown	N/A
1	X	X	0	0	XX	Shutdown	N/A
1	0	0	1	X	X0	Auto-Skip	VOUT_VSEL0
1	0	0	1	X	X1	Forced-PWM	VOUT_VSEL0
1	1	0	X	1	0X	Auto-Skip	VOUT_VSEL1
1	1	0	X	1	1X	Forced-PWM	VOUT_VSEL1
1	0	X	0	X	XX	Shutdown	N/A
1	0	1	1	X	XX	Forced-PWM	VOUT_VSEL0
1	1	X	X	0	XX	Shutdown	N/A
1	1	1	X	1	XX	Forced-PWM	VOUT_VSEL1

Active Discharge

When the KTB8399 buck is disabled, an active discharge feature connects an on-chip resistor (R_{LX_DIS}) between the LX and PGND pins. This resistor discharges the output capacitor through the inductor. By default, this feature is enabled; however, it can be disabled for applications that require a high impedance at the output during shutdown. Enable and disable the active discharge feature via I^2C commands by writing the DIS bit (B7) in the CONTROL configuration register (0x02).

POR and Software Register Reset

The KTB8399 does NOT contain non-volatile memory for the register settings. When V_{IN} rises above V_{POR}=1.8V, either at initial power up or after a temporary V_{IN} droop below V_{POR}, a Power-On Reset (POR) circuit resets all registers to their factory default settings. Thereafter, as long as V_{IN} remains above V_{POR}, the I^2C registers contents are retained, regardless if the buck is enabled or disabled. However, to reliably read or write to the I^2C registers, V_{IN} should be above V_{UVLO}.

To reset the registers manually via software, use I^2C to write a 1 to the RESET bit (B2) in the CONTROL configuration register (0x02). This resets nearly all registers to their default settings. There are a few exceptions, as makes logical sense; refer to the CONTROL register description.

Internal Status Monitor

The KTB8399 contains a MONITOR status register (0x05), which can be read to check the present status of the IC. The register has individual bits for the status of V_{OUT} power-OK, V_{IN} under-voltage lockout, V_{IN} over-voltage protection, V_{OUT} positive slew, V_{OUT} negative slew, software reset event latch, over-temperature thermal shutdown, and buck enable. Refer to the MONITOR register description for more details.

Input Under-Voltage Lockout (UVLO)

When the input voltage (V_{IN}) is below the under-voltage lockout threshold (V_{UVLO}), the buck is disabled. The I^2C registers and all logic pins remain functional during UVLO, so long as V_{IN} remains above V_{POR} = 1.8V. Exiting UVLO does ***not*** reset any registers. When V_{IN} rises above V_{UVLO}, either at initial power up or after a temporary V_{IN} droop below V_{UVLO}, and if the buck is enabled, the programmed soft-start ramp begins.

The UVLO status is reflected in the MONITOR register. UVLO events do ***not*** reset the registers to their defaults. V_{IN} must fall below V_{POR} to reset the registers.

Input Over-Voltage Protection (OVP)

When the input voltage (V_{IN}) is above the over-voltage protection threshold (V_{OVP}), the buck is disabled. The I^2C registers and all logic pins remain functional during OVP. When V_{IN} returns below V_{OVP}, and if the buck is enabled, the programmed soft-start ramp begins.

Just like UVLO, the OVP status is reflected in the MONITOR register.

Inductor Over-Current Protection (OCP)

Inductor peak current limit (I_{LX_PEAK}) and valley current limit (I_{LX_VALLEY}) protect the buck and inductor during over-current faults. The current limits control the buck's switching on a cycle-by-cycle basis and have a higher priority than the voltage regulation threshold.

During sustained over-current faults, the output voltage typically droops below the regulation threshold. The POK_STAT bit (B7) in the MONITOR register (0x05) indicates when V_{OUT} is okay (***above*** the V_{OUT_POK} threshold).

Output Short-Circuit Protection (SCP) and Hiccup Mode

During a short-circuit event at the buck's output, the inductor experiences a very low discharge voltage during the switching cycle's off-time (t_{OFF}, when the synchronous rectifier switch is on). In this case, the inductor current ramps down very slowly. In order to prevent inductor current runaway, the valley current limit (I_{LX_VALLEY}) extends t_{OFF}, keeping the inductor current well controlled.

If an over-current fault or short-circuit event persists for more than 250µs, the buck enters hiccup mode and pause all switching. After about 5ms, the buck attempts to soft-start. If the fault persists, the buck once again enters hiccup mode and periodically re-attempts soft-start until the fault is removed. The low duty-factor during hiccup mode prevents the IC from getting hot.

Thermal Shutdown Over-Temperature (OT)

Over-temperature (OT) protection occurs if the die junction temperature exceeds the thermal shutdown threshold (T_{J_SHDN}). During thermal shutdown, the buck pauses all switching until the die temperature cools. Once cooled, the buck re-starts with the programmed soft-start ramp.

The OT status is reflected in the MONITOR register.

Trim Options

The KTB8399 is factory trimmed using one-time programmable (OTP) registers. Standard versions are available for various default output voltage settings and modes – see the *Ordering Information* section. Contact a Kinetic Technologies representative regarding versions with other default settings or I²C slave addresses.

I²C Interface Description

I²C Serial Data Bus

The KTB8399 supports the I²C bus protocol. A device that sends data onto the bus is defined as a transmitter, and a device receiving data as a receiver. The device that controls the bus is called a master, whereas the devices controlled by the master are known as slaves. A master device must generate the serial clock (SCL), control bus access and generate START and STOP conditions to control the bus. The KTB8399 operates as a slave on the I²C bus. Within the bus specifications, a standard mode (100kHz maximum clock rate) and a fast mode (400kHz maximum clock rate) are defined. The KTB8399 works in both modes. Connections to the bus are made through the open-drain I/O lines SDA and SCL.

The following bus protocol has been defined in Figure 2:

- Data transfer may be initiated only when the bus is not busy.
- During data transfer, the data line must remain stable whenever the clock line is HIGH. Changes in the data line while the clock line is high are interpreted as control signals.

Accordingly, the following bus conditions have been defined:

Bus Not Busy
Both data and clock lines remain HIGH.

Start Data Transfer
A change in the state of the data line, from HIGH to LOW, while the clock is HIGH, defines a START condition.

Stop Data Transfer
A change in the state of the data line, from LOW to HIGH, while the clock line is HIGH, defines the STOP condition.

Data Valid
The state of the data line represents valid data when, after a START condition, the data line is stable for the duration of the HIGH period of the clock signal. The data on the line must be changed during the LOW period of the clock signal. There is one clock pulse per bit of data.

Each data transfer is initiated with a START condition and terminated with a STOP condition. The number of data bytes transferred between START and STOP conditions are not limited, and are determined by the master device. The information is transferred byte-wise and each receiver acknowledges with a ninth bit.

Acknowledge
Each receiving device, when addressed, is obliged to generate an acknowledge after the reception of each byte. The master device must generate an extra clock pulse that is associated with this acknowledge bit.

A device that acknowledges must pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is stable LOW during the HIGH period of the acknowledge-related clock pulse. Setup and hold times must also be taken into account.

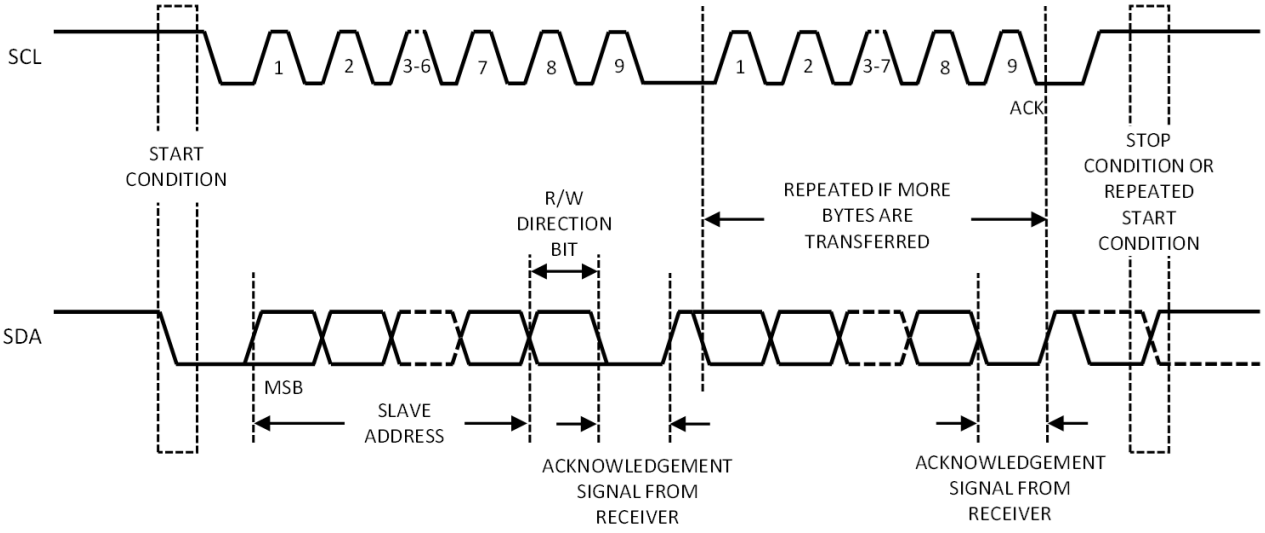


Figure 2. Data Transfer on I²C Serial Bus

The KTB8399 7-bit slave device address is listed in the Ordering Information table in page 3.

There are two kinds of I²C data transfer cycles: write cycle and read cycle.

I²C Write Cycle

For I²C write cycle, data is transferred from a master to a slave. The first byte transmitted is the 7-bit slave address plus one bit of '0' for write. Next follows a number of data bytes. The slave returns an acknowledge bit after each received byte. Data is transferred with the most significant bit (MSB) first. Figure 3 shows the sequence of the I²C write cycle.

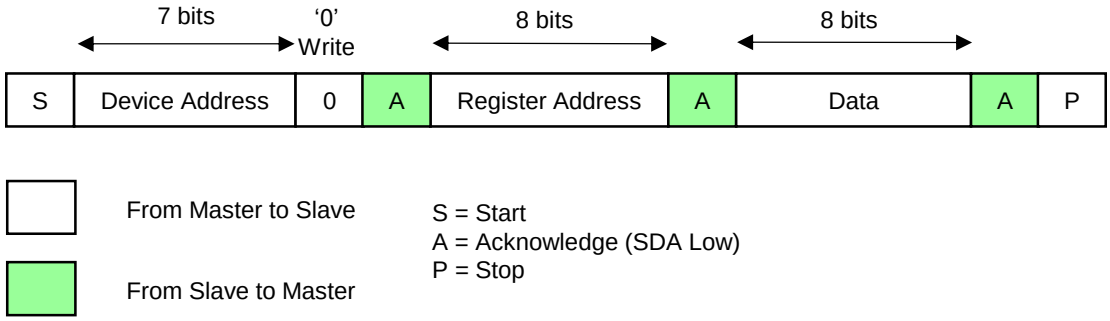


Figure 3. I²C Write Cycle

I²C Write Cycle Steps:

- Master generates start condition.
- Master sends 7-bit slave address and 1-bit data direction '0' for write.
- Slave sends acknowledge if the slave address is matched.
- Master sends 8-bit register address.
- Slave sends acknowledge.
- Master sends 8-bit data for that addressed register.
- Slave sends acknowledge.
- If master sends more data bytes, the register address will be incremented by one after each acknowledge.
- Master generate stop condition to finish the write cycle.

I²C Read Cycle

For I²C read cycle, data is transferred from a slave to a master. But to start the read cycle, master needs to write the register address first to define which register data to read. Figure 4 shows the steps of the I²C read cycle.

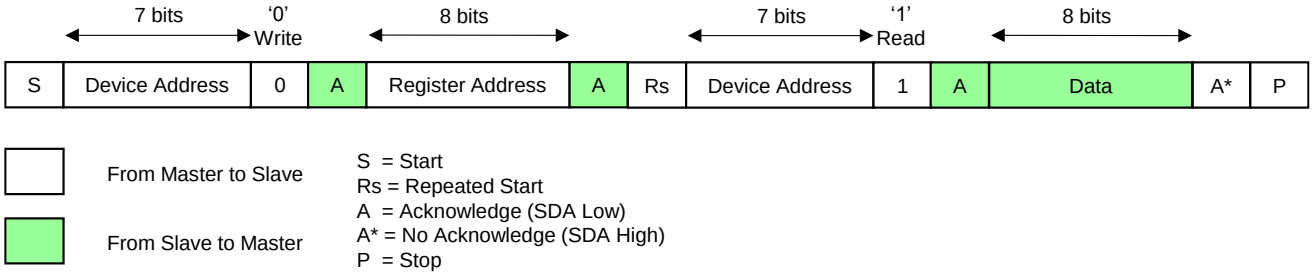


Figure 4. I²C Read Cycle

I²C Read Cycle Steps:

- Master generates start condition.
- Master sends 7-bit slave address and 1-bit data direction '0' for write.
- Slave sends acknowledge if the slave address is matched.
- Master sends 8-bit register address.
- Slave sends acknowledge.
- Master generates repeated start condition.
- Master sends 7-bit slave address and 1-bit data direction '1' for read.
- Slave sends acknowledge if the slave address is matched.
- Slave sends the data byte of that addressed register.
- If master sends acknowledge, the register address will be incremented by one after each acknowledge and the slave will continue to send the data for the updated addressed register.
- If master sends no acknowledge, the slave will stop sending the data.
- Master generate stop condition to finish the read cycle.

I²C Registers

I²C Slave Address

Options[10]	7-Bit Address	Write Address	Read Address	Bits 7 6 5 4 3 2 1 0
KTB8399M	0x60	0xC0	0xC1	1 1 0 0 0 0 0 *R/$\overline{W}$*

I²C Register Map

Hex Address	Name	Type	Access	Default Reset	B7	B6	B5	B4	B3	B2	B1	B0
0x00	VSEL0	Config	R/W	1xxx xxxx	EN0	VOUT_VSEL0[6:0]						
0x01	VSEL1	Config	R/W	1xxx xxxx	EN1	VOUT_VSEL1[6:0]						
0x02	CONTROL	Config	R/W	1000 x0xx	DIS	SLEW[2:0]			RANGE	RESET	MODE[1:0]	
0x03	ID1	Data	R	1010 0010	VENDOR[2:0]			DIE_ID[4:0]				
0x04	ID2	Status	R	0000 xxxx	RSVD	RSVD	RSVD	RSVD	DIE_REV[3:0]			
0x05	MONITOR	Status	R	0000 0000	POK_STAT	UVLO_STAT	OVP_STAT	POS_STAT	NEG_STAT	RESET_STAT	OT_STAT	EN_STAT

Register contents are reset in hardware to their default values by V_{IN} power-on reset. Additionally, most registers can be reset in software by writing 1 to the RESET bit (B2) in the CONTROL register (0x02). Default Reset bits marked with lower-case "x" in the register map and register details tables depend upon the ordered part# suffix and die revision; please see the *Ordering Information* section. Upper-case "X" used elsewhere in the tables designates "don't care".

VSEL0 Configuration Register

Register Address 0x00

Bit	Name	Access	Default Reset	Description
7	EN0	R/W	1	Software Buck Enable. When EN pin is low, the regulator is off. When EN pin is high and VSEL pin is low, the EN0 bit takes precedent.
6:0	VOUT_VSEL0[6:0]	R/W	xxx xxxx	Sets the nominal V_{OUT} regulation voltage when VSEL pin is low. When RANGE bit is low, the V_{OUT} range is from 0.600 to 1.39375V. When RANGE bit is high, the V_{OUT} range is from 1.440 to 3.345V. $$V_{OUT} = G \times (600mV + 6.25mV \times VOUT\_VSEL0)$$ where G = 1.0 when RANGE bit is low, and G = 2.4 when RANGE bit is high.

EN pin	VSEL pin	EN0 bit	Regulator
0	X	X	off
1	0	0	off
1	0	1	on

10. For Alternate 1/2/3 Slave Addresses, please contact a Kinetic Technologies representative.

VSEL1 Configuration Register

Register Address 0x01

Bit	Name	Access	Default Reset	Description
7	EN1	R/W	1	Software Buck Enable. When EN pin is low, the regulator is off. When EN pin is high and VSEL pin is high, the EN1 bit takes precedent. EN pin / VSEL pin / EN1 bit / Regulator 0 / X / X / off 1 / 1 / 0 / off 1 / 1 / 1 / on
6:0	VOUT_VSEL1[6:0]	R/W	xxx xxxx	Sets the nominal V_{OUT} regulation voltage when VSEL pin is high. When RANGE bit is low, the V_{OUT} range is from 0.600 to 1.39375V. When RANGE bit is high, the V_{OUT} range is from 1.440 to 3.345V. $V_{OUT} = G \times (600mV + 6.25mV \times VOUT\_VSEL1)$ where G = 1.0 when RANGE bit is low, and G = 2.4 when RANGE bit is high.

CONTROL Configuration Register

Register Address 0x02

Bit	Name	Access	Default Reset	Description
7	DIS	R/W	1	Active Discharge of LX when regulator is disabled. Discharging LX will discharge C_{OUT} through the inductor. 0 = V_{OUT} is high impedance when disabled. 1 = V_{OUT} is discharged through an internal pull-down resistor when disabled.
6:4	SLEW[2:0]	R/W	000	Slew Rate Control for DVS ramp-up/down rates. Works in conjunction with the RANGE bit.

RANGE bit	SLEW bits	DVS	f_{STEP}
0	000	3.125mV/µs	500kHz
0	001	3.125mV/µs	500kHz
0	010	6.25mV/µs	1MHz
0	011	6.25mV/µs	1MHz
0	100	12.5mV/µs	2MHz
0	101	12.5mV/µs	2MHz
0	110	25mV/µs	4MHz
0	111	25mV/µs	4MHz
1	000	3.75mV/µs	250kHz
1	001	3.75mV/µs	250kHz
1	010	7.5mV/µs	500kHz
1	011	7.5mV/µs	500kHz
1	100	15mV/µs	1MHz
1	101	15mV/µs	1MHz
1	110	25mV/µs	2MHz
1	111	25mV/µs	2MHz

<table>
<tr><th>Bit</th><th>Name</th><th>Access</th><th>Default Reset</th><th>Description</th></tr>
<tr><td>3</td><td>RANGE</td><td>R/W</td><td>x</td><td>Sets the V_{OUT} range. Also influences the DVS ramp-up/down slew rate (see SLEW[2:0] above).
<table>
<tr><th>RANGE bit</th><th>G</th><th>$V_{OUT(min)}$</th><th>$V_{OUT(max)}$</th><th>$V_{OUT(step)}$</th></tr>
<tr><td>0</td><td>1.0</td><td>0.600V</td><td>1.39375V</td><td>6.25mV</td></tr>
<tr><td>1</td><td>2.4</td><td>1.440V</td><td>3.345V</td><td>15mV</td></tr>
</table>
$$V_{OUT} = G \times (600mV + 6.25mV \times VOUT\_VSELn)$$
where G = 1.0 when RANGE bit is low, and G = 2.4 when RANGE bit is high.
Note: There is no slew-rate control when toggling the RANGE bit; therefore, it is recommended to disable V_{OUT} before changing the RANGE setting.</td></tr>
<tr><td>2</td><td>RESET</td><td>R/W</td><td>0</td><td>Software Reset to default register settings. Writing 1 resets nearly all the registers. The RESET bit always reads back as 0. Before self-clearing, the software reset event is latched into the RESET_STAT bit (B2) in the MONITOR register (0x05) and also into the RESET_INT bit (B2) in the INTLATCH register (0x07) if unmasked.
Items that are NOT reset by the RESET bit are:
1. RESET_STAT bit in the MONITOR register.
2. RESET_INT bit in the INTLATCH register</td></tr>
<tr><td>1:0</td><td>MODE[1:0]</td><td>R/W</td><td>xx</td><td>Auto-Skip vs. Forced-PWM mode control. Works in conjunction with the VSEL and MODE pin. Auto-Skip automatically uses PWM in continuous conduction at heavy loads and PFM pulse-skipping in discontinuous conduction at light loads. Refer to Table 1.</td></tr>
</table>

ID1 Data Register

Register Address 0x03

Bit	Name	Access	Default Reset	Description
7:5	VENDOR[2:0]	R	101	Vendor Identification 101 = Kinetic Technologies
4:0	DIE_ID[4:0]	R	xxxx	Die Type Identification

ID2 Status Register

Register Address 0x04

Bit	Name	Access	Default Reset	Description
7	RSVD	R	0	Reserved. Always reads back as 0.
6	RSVD	R	0	Reserved. Always reads back as 0.
5	RSVD	R	0	Reserved. Always reads back as 0.
4	RSVD	R	0	Reserved. Always reads back as 0.
3:0	DIE_REV[3:0]	R	xxxx	Die Revision Identification

MONITOR Status Register

Register Address 0x05

Bit	Name	Access	Default Reset	Description
7	POK_STAT	R	0	V_{OUT} Power OK Status 0 = V_{OUT} is below POK comparator threshold 1 = V_{OUT} is above POK comparator threshold During normal operation, the POK_STAT bit is 1. During disabled, soft-start or overload conditions, the POK_STAT bit is 0.
6	UVLO_STAT	R	0	V_{IN} Under-Voltage Status 0 = V_{IN} is above the UVLO comparator threshold 1 = V_{IN} is below the UVLO comparator threshold During normal operation or when disabled, the UVLO_STAT bit is 0. During low input voltage conditions, the buck is disabled by the UVLO comparator and the UVLO_STAT bit is set to 1 (so long as V_{IN} remains above V_{POR}=1.8V).
5	OVP_STAT	R	0	V_{IN} Over-Voltage Status 0 = V_{IN} is below the OVP comparator threshold 1 = V_{IN} is above the OVP comparator threshold During normal operation or when disabled, the OVP_STAT bit is 0. During high input voltage conditions, the buck is disabled by the OVP comparator and the OVP_STAT bit is set to 1.
4	POS_STAT	R	0	V_{OUT} Positive Slew Status 0 = V_{OUT} is at its set value 1 = V_{OUT} is slewing in the positive direction towards it set value During normal operation, the POS_STAT bit is 0. During soft-start or positive DVS slew transitions, the POS_STAT bit is 1.
3	NEG_STAT	R	0	V_{OUT} Negative Slew Status 0 = V_{OUT} is at its set value 1 = V_{OUT} is slewing in the negative direction towards it set value During normal operation, the NEG_STAT bit is 0. During negative DVS slew transitions, the NEG_STAT bit is 1.
2	RESET_STAT	R/C	0	Software Reset Status Latch 0 = software reset was not performed since this bit was cleared 1 = software reset was performed since this bit was cleared The RESET_STAT bit is set to 1 when a software reset is written to the RESET bit (B2) in the CONTROL register. The RESET_STAT bit is ***NOT*** reset by the software reset. Instead, it is cleared (reset to 0) after the RESET_STAT bit is read or whenever chip power is removed (V_{IN}<V_{POR}).
1	OT_STAT	R	0	Over-Temperature Status 0 = the die is not in thermal shutdown 1 = the die is in thermal shutdown During normal operation, the OT_STAT bit is 0. During thermal shutdown, the OT_STAT bit is 1.
0	EN_STAT	R	0	Enable Status 0 = the buck is disabled by hardware or software 1 = the buck is enabled by hardware or software During normal operation, the EN_STAT bit is 1. During hardware or software disabled conditions, the EN_STAT bit is 0.

Applications Information

Recommended Inductors

The KTB8399 is trimmed for inductors with nominal inductance of 330nH or 470nH. Select an inductor with a saturation current rating that is higher than the KTB8399 peak current limit. Also, choose an inductor with sufficient temperature-rise current rating to satisfy the RMS load-current of the application. Consider the inductor's resistance (both DCR and ACR at 2.4MHz), since these will affect the efficiency. (Generally, the ACR vs. frequency characteristic is available upon request from the inductor supplier.) Larger physical case-sizes, good winding designs, and better magnetic materials can increase efficiency. Typically, metric 2012 and 2016 case-sizes are suitable. Table 1 is a list of recommended inductors from two leading suppliers.

Table 1. Recommended Inductors

Maker	Part #	L (typ)	DCR (typ)	I_{SAT} (min)	$I_{\Delta T+40C}$ (min)	Size (typ/typ/max)
Murata	DFE201210U-R33M DFE201210U-R47M	330nH 470nH	25mΩ 34mΩ	5.2A 4.4A	3.4A 3.0A	2.0 x 1.2 x 1.0mm
Murata	DFE201610E-R33M DFE201610E-R47M	330nH 470nH	21mΩ 26mΩ	5.5A 4.8A	4.0A 3.6A	2.0 x 1.6 x 1.0mm
Murata	DFE201612E-R33M DFE201612E-R47M	330nH 470nH	15mΩ 20mΩ	6.3A 5.5A	4.8A 4.5A	2.0 x 1.6 x 1.2mm
Samsung E-M	CIGT201208EHR47MNE	470nH	31mΩ	4.1A	3.7A	2.0 x 1.25 x 0.8mm
Samsung E-M	CIGT201210UHR33MNE CIGT201210UHR47MNE	330nH 470nH	21mΩ 25mΩ	5.4A 4.9A	4.0A 3.6A	2.0 x 1.25 x 1.0mm
Samsung E-M	CIGT201608EHR47MNE	470nH	24mΩ	4.3A	4.3A	2.0 x 1.6 x 0.8mm
Samsung E-M	CIGW201610GHR33MLE CIGW201610GLR47MLE CIGT201610EHR47MNE	330nH 470nH 470nH	18mΩ 21mΩ 18mΩ	5.5A 5.0A 5.5A	4.0A 4.7A 4.8A	2.0 x 1.6 x 1.0mm

Recommended Capacitors

Ceramic input and output capacitors with X5R or X6S are recommended due to their low ESR, low ESL, low temperature coefficients, and small physical sizes. Consider the voltage rating, size, and DC bias derating characteristic of the capacitor. Table 2 is a list of recommended capacitors from two leading suppliers.

Table 2. Recommended Capacitors

Maker	Part #	C (typ)	$C_{EFF(3V)}$	V_{RATING}	T_{CHAR}	Size
Murata	GRM155R60J106ME GRM155R60J156ME GRM155R60G226ME GRM155R60J226ME	10µF 15µF 22µF 22µF	3.5µF 5µF 7µF 7µF	6.3V 6.3V 4.0V 6.3V	X5R	(0402) 1.0 x 0.5mm
Murata	GRM188R60G226ME GRM188R60J226ME	22µF 22µF	9µF 9µF	4.0V 6.3V	X5R	(0603) 1.6 x 0.8mm
Samsung E-M	CL05A106MQ5 CL05A156MQ5	10µF 15µF	4µF 6µF	6.3V 6.3V	X5R	(0402) 1.0 x 0.5mm
Samsung E-M	CL10A226MQ8 CL10A476MQ8	22µF 47µF	11µF 14µF	6.3V 6.3V	X5R	(0603) 1.6 x 0.8mm

Input Capacitor

Choose an input capacitor with voltage rating of 6.3V or more, 10uF total nominal capacitance or more, and 1005M (0402) case-size or larger. Larger values and larger case-size provide more effective capacitance when considering the DC bias derating characteristic of the capacitor. If the application's input voltage is supplied through a connector or a cable, add additional bypass/bulk capacitance where V_{IN} first arrives to the PCB.

Output Capacitors

Choose output capacitors with voltage rating of 4.0V or more, 20µF total nominal capacitance or more, and 1005M (0402) case-size or larger. Consider the V_{OUT} setting of the regulator and how case size has a significant impact

on DC bias derating. At high V_{OUT} settings, more total nominal capacitance is needed to achieve the same effective capacitance compared to lower V_{OUT} settings.

For the very best possible load transient response, use multiple capacitors in parallel to achieve sufficient total effective output capacitance:

$$C_{OUT_{EFFECTIVE}} \geq \frac{L \times I_{STEP}}{33m\Omega \times (V_{IN} - V_{OUT})}$$

where I_{STEP} is the largest load transient step in the application. Please note that the above formula is already guard-banded by a margin of 2x to accommodate capacitor and inductor tolerances and the variability of a transient arrival time with respect to the switching cycle of the regulator.

If needed, the total effective output capacitance can be distributed by placing additional capacitors remotely at the point of load. In applications where transient performance is less critical, especially when V_{IN} minus V_{OUT} is small, it is acceptable to reduce the total effective output capacitance to save board space and cost at the expense of load transient droop and soar.

As a design example, consider a system with V_{IN} = 3.2V (min), V_{OUT} = 1.8V, and I_{STEP} = 2A (max):

$$C_{OUT_{EFFECTIVE}} \geq \frac{330nH \times 2A}{33m\Omega \times (3.2V - 1.8V)} \cong 14\mu F$$

In this example, choose output capacitors with total effective capacitance of 14µF or more at a DC bias of 1.8V. A single 15µF capacitor will not be enough when considering its DC bias characteristic, per Figure 5. At 1.8V bias, it retains only about 8µF; therefore, for best transient response, use two of these capacitors in parallel for a total effective capacitance of 16µF.

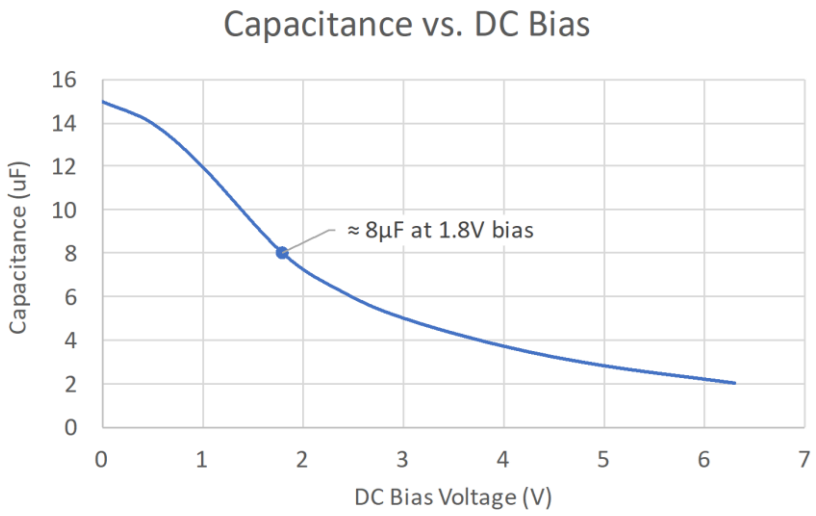


Figure 5. Typical DC bias derating characteristic for example 15µF ceramic capacitor.

When operating near dropout with high V_{OUT} and low V_{IN} (for example 3.3V_{OUT} and 3.6V_{IN}), add additional bulk capacitance to C_{OUT} to reduce V_{OUT} droop and ringing during load transient events that is inherent in buck regulators operating at high duty-cycle.

Recommended PCB Layout

Refer to Figure 6 for two example PCB layouts optimized for small footprint, low EMI, and good performance. One example fits in a narrow rectangular area, while the other example is nearly square. Both examples follow the below PCB layout recommendations:

1. Connect the input capacitor C_{IN} as close as possible to the VIN and PGND pins using top-side thick metal traces.
2. Connect the ground terminals of output capacitors C_{OUT} as close as possible to the ground terminal of C_{IN} and the PGND pins using top-side metal.
3. Connect the local top-side PGND island to the PCB ground plane using multiple parallel vias.
4. Do not connect the AGND pins directly to the top-side PGND. Instead, connect the AGND pins to the PCB ground plane using their own vias.
5. Connect the inductor to the LX pins with a wide trace. For smallest C_{IN}, the LX trace will not fit between the top-side landing pads of a 1005M (0402) size capacitor. Therefore, the LX trace is normally routed on PCB layer 2 or layer 3 using multiple parallel vias.
    a. The added inductance of vias in series with the actual inductor have minimal effect on performance. However, it is important to use enough parallel vias to handle the peak inductor current in the application.
    b. If using a larger C_{IN} of 1608M (0603) size or more, it is possible to squeeze the LX route on the top-side between the C_{IN} landing pads. If pinched, make sure to widen the LX trace as much as possible before and after the pinched area.
6. Connect the V_{OUT} terminals of the inductor to the output capacitors with a wide and short trace.
7. Route the V_{OUT} sense trace from C_{OUT} to the VOUT pin with care to keep it away from noisy traces, especially the LX trace. Additionally, use ground fill to shield noise from coupling into the V_{OUT} sense.
8. Depending upon PCB design rules, it may be possible to place filled micro-vias directly under WLCSP bumps. If not, route short traces to nearby vias.

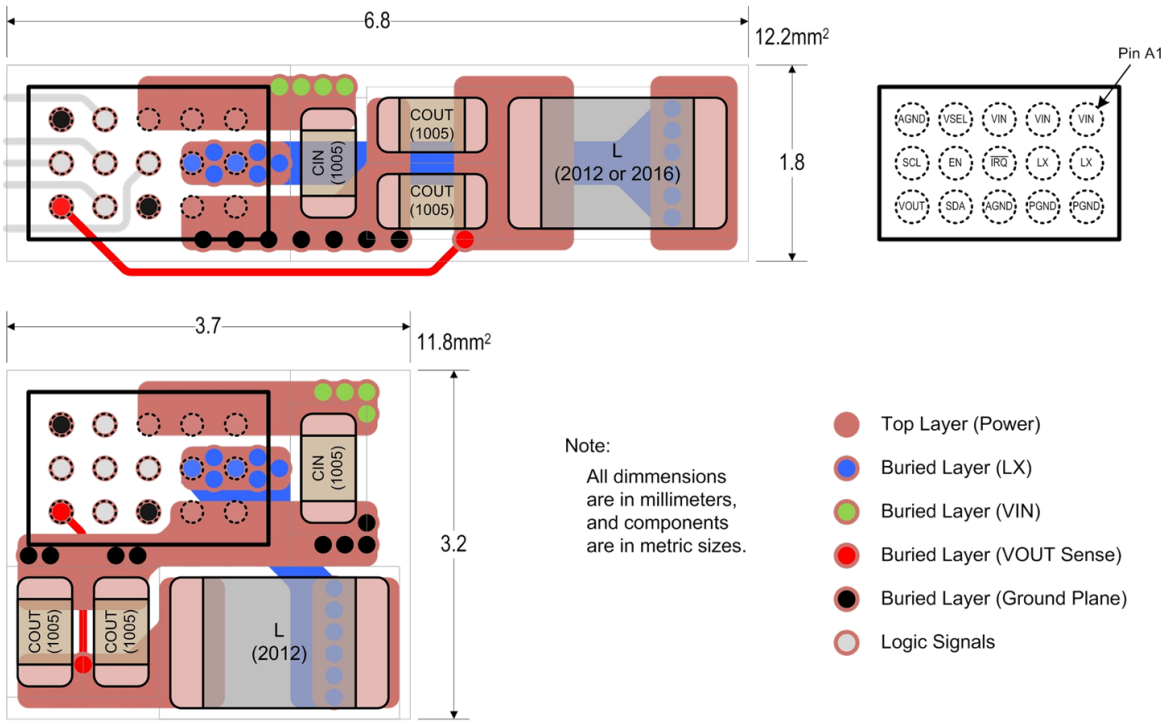
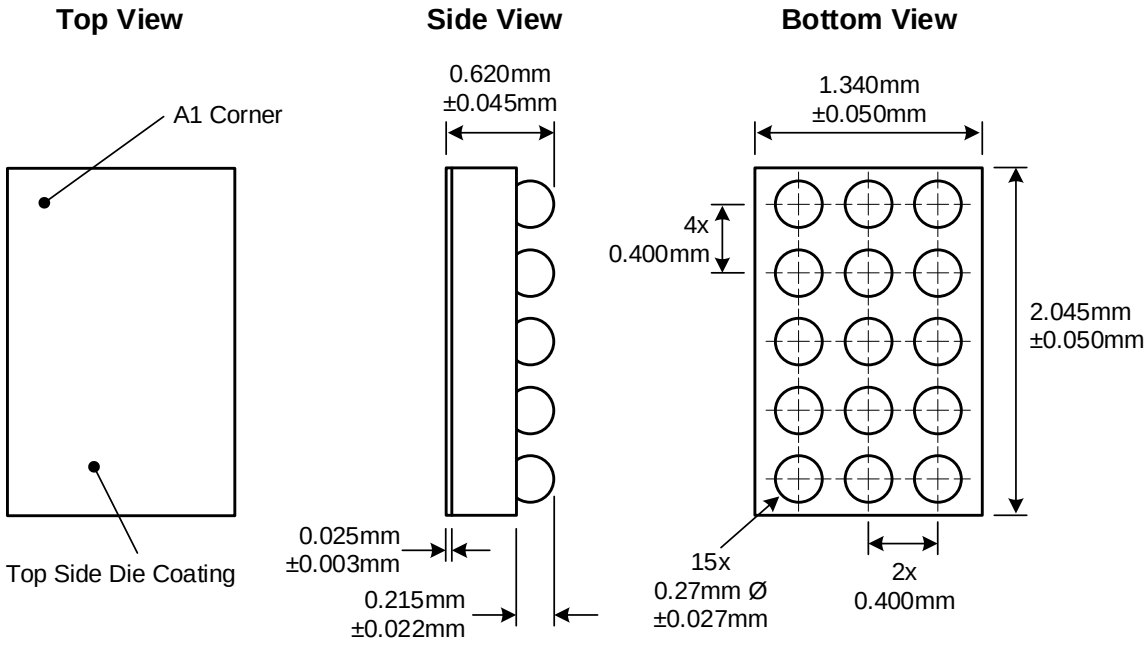


Figure 6. Two Recommended PCB Layouts

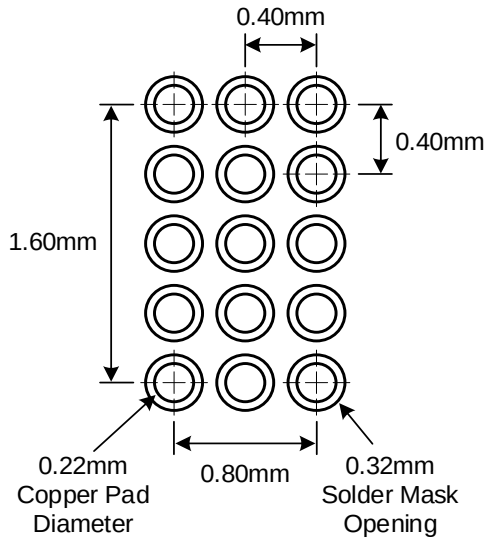
Packaging Information

WLCSP35-15 (1.340mm x 2.045mm x 0.620mm)



Recommended Footprint

(NSMD Pad Type)



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